

<b>Device</b>	<b>TC23x</b>
<b>Marking/Step</b>	<b>ES-AC, AC</b>
<b>Package</b>	<b>see Data Sheet</b>

## 10312AERRA

This Errata Sheet describes the deviations from the current user documentation.

**Table 1 Current Documentation<sup>1)</sup>**

TC21x/TC22x/TC23x User's Manual	V1.1	2014-12
TC233/TC234/TC237 AC-Step Data Sheet	V1.0	2017-03
TriCore TC1.6P & TC1.6E Core Architecture, Instruction Set	V1.0D10, V1.0D15	2012-02, 2013-07
OCDS User's Manual <sup>2)</sup>	V2.9.1	2014-11-24

- 1) Newer versions replace older versions, unless specifically noted otherwise.
- 2) Distribution under NDA, only relevant for tool development not for application development.

Make sure you always use the corresponding documentation for this device (User's Manual, Data Sheet, Documentation Addendum (if applicable), TriCore Architecture Manual, Errata Sheet) available in category 'Documents' at [www.infineon.com/AURIX](http://www.infineon.com/AURIX) and [www.myInfineon.com](http://www.myInfineon.com).

### Conventions used in this document

Each erratum identifier follows the pattern **Module\_Arch.TypeNumber**:

- **Module**: subsystem, peripheral, or function affected by the erratum
- **Arch**: microcontroller architecture where the erratum was initially detected
  - **AI**: Architecture Independent
  - **TC**: TriCore

- **Type:** category of deviation
  - **[none]:** Functional Deviation
  - **P:** Parametric Deviation
  - **H:** Application Hint
  - **D:** Documentation Update
- **Number:** ascending sequential number within the three previous fields. As this sequence is used over several derivatives, including already solved deviations, gaps inside this enumeration can occur.

## Notes

1. This Errata Sheet applies to all temperature and frequency versions and to all memory size variants, unless explicitly noted otherwise. For a derivative synopsis, see the latest Data Sheet/User's Manual.  
This Errata Sheet covers several device versions. If an issue is related to a particular module, and this module is not specified for a specific device version, this issue does not apply to this device version.  
E.g. issues with identifier "ETH" only apply to devices with a specific extended feature set.
2. Devices marked with EES or ES are engineering samples which may not be completely tested in all functional and electrical characteristics, therefore they should be used for evaluation only.  
The specific test conditions for EES and ES are documented in a separate Status Sheet.
3. This device is equipped with TriCore "TC1.6E" core(s). Some of the errata have workarounds which are possibly supported by the tool vendors. Some corresponding compiler switches need possibly to be set. Please see the respective documentation of your compiler.  
For effects of issues related to the on-chip debug system, see also the documentation of the debug tool vendor.

# 1 History List / Change Summary

**Table 2 History List**

Version	Date	Remark
1.0	2016-10-21	<ul style="list-style-type: none"> <li>• New/updated text modules see columns “Change” in tables 4..6 of errata sheet V1.0.</li> <li>• The following text modules have been removed, as no device variant exists of this design step (AC) that includes EMEM, ETH, FFT modules:               <ul style="list-style-type: none"> <li>– EMEM_TC.H002 (EMEM will raise ECC errors when not properly initialized)</li> <li>– ETH_AI.003 (Overflow Status bits of Missed Frame and Buffer Overflow counters get cleared without a Read operation)</li> <li>– ETH_AI.H001 (Sequence for Switching between MII and RMII Modes)</li> <li>– ETH_TC.004 (DMA Access to Reserved/Protected Resources: FPI Error Response not correctly evaluated)</li> <li>– ETH_TC.H002 (Minimum operation frequency for Ethernet MAC)</li> <li>– ETH_TC.H003 (Interrupt Generation by Wake-up or Magic Packet Frames)</li> <li>– FFT_TC.001 (FFT Access with disabled FFT Module)</li> <li>– FFT_TC.002 (FFT Kernel Reset Function)</li> <li>– FFT_TC.003 (No Error reported upon Write to FFT Registers in User Mode)</li> </ul> </li> <li>• The following text module has been removed, as no ADAS variant exists of this design step (AC)               <ul style="list-style-type: none"> <li>– SCU_TC.H012 (Overlay Feature for ADAS Variants)</li> </ul> </li> </ul>

**Table 2 History List (cont'd)**

<b>Version</b>	<b>Date</b>	<b>Remark</b>
1.1	2017-03-31	<ul style="list-style-type: none"> <li>• New/updated text modules see columns “Change” in tables 4..6 of errata sheet V1.1.</li> <li>• Removed reference to “GTM-IP 210 Errata Sheet” in Table 1 - all GTM errata relevant for this design step are considered in this TC23x errata sheet</li> <li>• The following text modules have been included in Table 3 (Errata fixed in this step):               <ul style="list-style-type: none"> <li>– ADC_TC.P007 (Additional Parameter for Data Sheet: Wakeup Time <math>t_{WU}</math>): see specification of <math>t_{WU}</math> in TC23x AC-Step Data Sheet</li> <li>– IO_TC.P003 (Calculating the 1.3 V Current Consumption for TC23x): see corresponding section in TC23x AC-Step Data Sheet</li> <li>– PADS_TC.P007 (Connection of Ball U17 in LFBGA-292 Package): see chapter “Package and Pinning Definitions” in TC23x AC-Step Data Sheet</li> </ul> </li> </ul>
1.2	2017-11-03	<ul style="list-style-type: none"> <li>• Update: new/updated text modules see columns “Change” in tables 4..6 of errata sheet V1.2.</li> </ul>
1.3	2018-06-11	<ul style="list-style-type: none"> <li>• New/updated text modules see columns “Change” in tables 4..6 of errata sheet V1.3</li> <li>• Replaced:               <ul style="list-style-type: none"> <li>– DMA_TC.029 (DMA Double Buffering Overflow),</li> <li>– DMA_TC.047 (DMA Double Buffering Buffer Switch),</li> <li>– DMA_TC.057 (Double Buffering Overflow Causes Other Channel Corruption)</li> <li>– &gt;&gt; replaced by DMA_TC.061 (DMA Double Buffering Operations)</li> </ul> </li> </ul>

**Table 2 History List (cont'd)**

Version	Date	Remark
... 1.3 (cont'd)	...	... <ul style="list-style-type: none"> <li>Removed:               <ul style="list-style-type: none"> <li>GTM_TC.010 (Effects of GTM Resets) - TC23x..TC21x do not have GTM SRAM</li> </ul> </li> </ul>
1.4	2019-08-30	<ul style="list-style-type: none"> <li>Update: new/updated text modules see columns "Change" in tables 4..6 of errata sheet V1.4</li> </ul>
1.5	2020-11-06	<ul style="list-style-type: none"> <li>Update: new/updated text modules see columns "Change" in tables 4..6 of errata sheet V1.5</li> </ul>
1.6	2022-07-04	<ul style="list-style-type: none"> <li>Update: new/updated text modules see columns "Change" in tables 4..6</li> </ul>

**Table 3 Errata fixed in this step**

Errata	Short Description	Change
ADC_TC.P007	Additional Parameter for Data Sheet: Wakeup Time $t_{WU}$	Fixed
CCU_TC.002	Clock Monitors - Target Monitoring Frequency Selection	Fixed
CCU_TC.003	Maximum Amplitude for Frequency Modulation	Fixed
FLASH_TC.044	Repetitive Erase Suspend Requests on Data Flash	Fixed
I0_TC.P003	Calculating the 1.3 V Current Consumption for TC23x	Fixed
MTU_TC.016	Wrong Address(es) Tracked in Registers ETRRx of TC1.6E CPU0 PSPR and DSPR	Fixed
MultiCAN_AI.047	Transmit Frame Corruption after Protocol Exception (CAN FD only)	Fixed

**Table 3 Errata fixed in this step (cont'd)**

<b>Errata</b>	<b>Short Description</b>	<b>Change</b>
MultiCAN_TC.043	CAN FD: Idle Condition	Fixed (node 0,1 only)
MultiCAN_TC.044	CAN FD: Missing Hardsync	Fixed (node 0,1 only)
PADS_TC.P007	Connection of Ball U17 in LFBGA-292 Package	Fixed
RESET_TC.007	Unexpected SMU Reset during SSW execution if no HARR requested	Fixed
SMU_TC.005	Unexpected/Incorrect Reset caused by SMU Alarms	Fixed

*Note: Changes to the previous errata sheet version are particularly marked in column "Change" in the following tables.*

**Table 4 Functional Deviations**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<a href="#">ADC_AI.016</a>	<a href="#">No Channel Interrupt in Fast Compare Mode with GLOBRES</a>		<a href="#">29</a>
<a href="#">ADC_TC.068</a>	<a href="#">Effect of VAGND Cross Coupling on Conversion Result</a>		<a href="#">29</a>
<a href="#">ASCLIN_TC.004</a>	<a href="#">SLSO in SPI mode still active after module disable</a>		<a href="#">32</a>
<a href="#">ASCLIN_TC.005</a>	<a href="#">Unjustified collision detection error in half-duplex SPI mode</a>		<a href="#">32</a>
<a href="#">ASCLIN_TC.006</a>	<a href="#">Unjustified response timeout in LIN slave mode</a>		<a href="#">33</a>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>ASCLIN_TC.007</b>	<b>Break Detected in LIN Frames in Soft Suspend mode</b>		<b>33</b>
<b>ASCLIN_TC.008</b>	<b>Response timeout in LIN Mode in case of header only</b>		<b>33</b>
<b>ASCLIN_TC.009</b>	<b>RFL flag set in Buffer Mode when Receive FIFO Inlet is disabled</b>		<b>34</b>
<b>ASCLIN_TC.010</b>	<b>Flush of TXFIFO leads to frame transmission</b>		<b>34</b>
<b>ASCLIN_TC.012</b>	<b>Recover sequence after timeout in LIN master mode</b>	<b>New</b>	<b>35</b>
<b>BROM_TC.008</b>	<b>Sporadic Power-on Reset after Wake-up from Standby Mode</b>		<b>35</b>
<b>BROM_TC.015</b>	<b>DSPR Data Integrity after Wake-up from Standby Mode</b>		<b>36</b>
<b>CPU_TC.123</b>	<b>Data Corruption possible when CPU GPR accesses made via SRI slave with CPU running</b>		<b>37</b>
<b>CPU_TC.127</b>	<b>Pending Interrupt Priority Number PIPN in Register ICR</b>		<b>38</b>
<b>CPU_TC.132</b>	<b>Unexpected PSW values used upon Fast Interrupt entry</b>		<b>39</b>
<b>DAP_TC.002</b>	<b>DAP client_blockread has Performance issue in Specific Operation Modes</b>		<b>40</b>
<b>DAP_TC.003</b>	<b>DAP CRC32 definition and algorithm</b>		<b>41</b>
<b>DAP_TC.004</b>	<b>DAP client_blockwrite telegram with CRC6 and CRC32 protection options</b>		<b>42</b>
<b>DAP_TC.005</b>	<b>DAP client_read: dirty bit feature of Cerberus' Triggered Transfer Mode</b>		<b>43</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>DAP_TC.006</b>	<b>CRC6 error in telegram following a get_CRCdown telegram prevents reset of CRC32 calculator</b>		<b>43</b>
<b>DAP_TC.007</b>	<b>Incomplete client_blockread telegram in DXCM mode when using the “read CRCup” option</b>		<b>44</b>
<b>DAP_TC.009</b>	<b>CRC6 error in client_blockwrite telegram</b>		<b>44</b>
<b>DMA_TC.015</b>	<b>DMA Double Buffering: No Timestamp Support</b>		<b>45</b>
<b>DMA_TC.016</b>	<b>Byte and Half-word Write Accesses to specific Registers not supported</b>		<b>45</b>
<b>DMA_TC.017</b>	<b>Pattern Detection Double Interrupt Trigger when INTCT = 11<sub>B</sub></b>		<b>46</b>
<b>DMA_TC.018</b>	<b>FPI timeout can cause pipelined register reads to break</b>		<b>46</b>
<b>DMA_TC.019</b>	<b>CBS Accesses with Large SPB:SRI Clock Ratios Configured</b>		<b>47</b>
<b>DMA_TC.020</b>	<b>DMA Conditional Linked List: Circular Buffer Enabled</b>		<b>47</b>
<b>DMA_TC.021</b>	<b>Combined Software/Hardware Controlled Mode Spurious Errors</b>		<b>48</b>
<b>DMA_TC.022</b>	<b>Conditional Linked List: Bus Error</b>		<b>48</b>
<b>DMA_TC.024</b>	<b>Suspend Request coincident with Channel Activation</b>		<b>49</b>
<b>DMA_TC.025</b>	<b>Conditional Linked List: new non-CLL mode TCS load can corrupt SDCRC RAM write</b>		<b>49</b>
<b>DMA_TC.026</b>	<b>Linked List: Failed TCS load can trigger wrap interrupt</b>		<b>50</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>DMA_TC.028</b>	<b>Transaction Request Lost (TRL) Interrupt Service Request Behaviour</b>		<b>50</b>
<b>DMA_TC.031</b>	<b>CHCSR.ICH can be incorrectly set after pattern match</b>		<b>50</b>
<b>DMA_TC.034</b>	<b>DMA Timestamp and Destination Circular Buffer</b>		<b>51</b>
<b>DMA_TC.035</b>	<b>Last DMA Transaction in a Linked List triggers a DMA Daisy Chain</b>		<b>52</b>
<b>DMA_TC.036</b>	<b>Linked List: SADR/DADR can be overwritten when loading a non-LL TCS</b>		<b>53</b>
<b>DMA_TC.037</b>	<b>Conditional Linked List: Bit TSR.CH not cleared for a CLL transaction upon pattern match</b>		<b>53</b>
<b>DMA_TC.038</b>	<b>Linked List: SIT interrupt when SIT bit set in newly loaded TCS</b>		<b>54</b>
<b>DMA_TC.039</b>	<b>Read Data CRC</b>		<b>54</b>
<b>DMA_TC.040</b>	<b>DMA Linked Lists: Intermittent Clearing of Hardware Transaction Request Enable with mixed mode Transaction Control Sets</b>		<b>55</b>
<b>DMA_TC.041</b>	<b>DMA Circular Buffer Wrap Interrupt</b>		<b>55</b>
<b>DMA_TC.042</b>	<b>DMA Interrupt from Channel reported before Completion of DMA Transaction</b>		<b>56</b>
<b>DMA_TC.043</b>	<b>DMA Write Move Data Corruption for non 32-byte Aligned Cacheable Source Address</b>		<b>57</b>
<b>DMA_TC.044</b>	<b>Clock Switch after SPB Error Reported results in Spurious SRI Error</b>		<b>58</b>
<b>DMA_TC.045</b>	<b>DMA Reconfigures DMA Channels Lockup</b>		<b>58</b>
<b>DMA_TC.046</b>	<b>Shadow Operation Read Only Mode</b>		<b>58</b>
<b>DMA_TC.049</b>	<b>Bus Error Reported During LL TCS Load</b>		<b>59</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>DMA_TC.050</b>	<b>Clearing CHCSR.FROZEN during Double Buffering</b>		<b>59</b>
<b>DMA_TC.052</b>	<b>SER and DER During Linked List Operations</b>		<b>60</b>
<b>DMA_TC.053</b>	<b>TS16_ERR Type of Error Reporting Unreliable</b>		<b>61</b>
<b>DMA_TC.054</b>	<b>DMA Channel Halt Acknowledge Unreliable</b>		<b>61</b>
<b>DMA_TC.055</b>	<b>ICU to DMA Interface in Sleep Mode</b>		<b>62</b>
<b>DMA_TC.056</b>	<b>TSR and SUSENR Access Protection Unreliable</b>		<b>62</b>
<b>DMA_TC.058</b>	<b>Linked List Load Transaction Control Set (TCS) Integrity Error</b>		<b>64</b>
<b>DMA_TC.061</b>	<b>DMA Double Buffering Operations</b>		<b>65</b>
<b>DMA_TC.062</b>	<b>Termination of DMA Transaction for Pattern Match</b>		<b>67</b>
<b>DMA_TC.063</b>	<b>DMA Timestamp Destination Address</b>		<b>67</b>
<b>DMA_TC.064</b>	<b>DMA Daisy Chain Request</b>		<b>68</b>
<b>DMA_TC.065</b>	<b>DMA Move Concurrent Bus Accesses</b>		<b>69</b>
<b>DMA_TC.066</b>	<b>DMA Double Buffering Operations - Update Address Pointer</b>		<b>69</b>
<b>DTS_TC.001</b>	<b>Temperature Sensor Formula</b>		<b>70</b>
<b>FLASH_TC.052</b>	<b>Use of Write Page Once command</b>		<b>70</b>
<b>FlexRay_AI.087</b>	<b>After reception of a valid sync frame followed by a valid non-sync frame in the same static slot the received sync frame may be ignored</b>		<b>71</b>
<b>FlexRay_AI.088</b>	<b>A sequence of received WUS may generate redundant SIR.WUPA/B events</b>		<b>72</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>FlexRay_AI.089</b>	<b>Rate correction set to zero in case of SyncCalcResult=MISSING_TERM</b>		<b>72</b>
<b>FlexRay_AI.090</b>	<b>Flag SFS.MRCS is set erroneously although at least one valid sync frame pair is received</b>		<b>73</b>
<b>FlexRay_AI.091</b>	<b>Incorrect rate and/or offset correction value if second Secondary Time Reference Point (STRP) coincides with the action point after detection of a valid frame</b>		<b>74</b>
<b>FlexRay_AI.092</b>	<b>Initial rate correction value of an integrating node is zero if pMicroInitialOffsetA,B = 0x00</b>		<b>75</b>
<b>FlexRay_AI.093</b>	<b>Acceptance of startup frames received after reception of more than gSyncNodeMax sync frames</b>		<b>75</b>
<b>FlexRay_AI.094</b>	<b>Sync frame overflow flag EIR.SFO may be set if slot counter is greater than 1024</b>		<b>76</b>
<b>FlexRay_AI.095</b>	<b>Register RCV displays wrong value</b>		<b>77</b>
<b>FlexRay_AI.096</b>	<b>Noise following a dynamic frame that delays idle detection may fail to stop slot</b>		<b>77</b>
<b>FlexRay_AI.097</b>	<b>Loop back mode operates only at 10 MBit/s</b>		<b>78</b>
<b>FlexRay_AI.099</b>	<b>Erroneous cycle offset during startup after abort of startup or normal operation</b>		<b>79</b>
<b>FlexRay_AI.100</b>	<b>First WUS following received valid WUP may be ignored</b>		<b>80</b>
<b>FlexRay_AI.101</b>	<b>READY command accepted in READY state</b>		<b>80</b>
<b>FlexRay_AI.102</b>	<b>Slot Status vPOC!SlotMode is reset immediately when entering HALT state</b>		<b>81</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>FlexRay_AI.103</b>	<b>Received messages not stored in Message RAM when in Loop Back Mode</b>		<b>82</b>
<b>FlexRay_AI.104</b>	<b>Missing startup frame in cycle 0 at coldstart after FREEZE or READY command</b>		<b>82</b>
<b>FlexRay_AI.105</b>	<b>RAM select signals of IBF1/IBF2 and OBF1/OBF2 in RAM test mode</b>		<b>83</b>
<b>FlexRay_AI.106</b>	<b>Data transfer overrun for message transfers Message RAM to Output Buffer (OBF) or from Input Buffer (IBF) to Message RAM</b>		<b>84</b>
<b>GTM_AI.132</b>	<b>GTM_TOP level: AEI write to BRIDGE_MODE register can result in blocking of AEI configuration interface</b>		<b>87</b>
<b>GTM_AI.141</b>	<b>TIM: Incorrect data captured to GPR registers and routed via ARU when EGPRi_SEL,GPRi_SEL= 100 in TIM channel mode TIEM, TPWM, TIPM, TPIM, TGPS</b>		<b>88</b>
<b>GTM_AI.142</b>	<b>TIM: Incorrect data captured to GPR registers and routed via ARU when EGPRi_SEL,GPRi_SEL= 100 in TIM channel mode TBCM</b>		<b>89</b>
<b>GTM_AI.143</b>	<b>GTM_TOP level: AEI pipelined write to GTM_BRIDGE_MODE register directly after setting aei_reset='0' can result in blocking of AEI configuration interface</b>	<b>Update</b>	<b>90</b>
<b>GTM_AI.144</b>	<b>TIM: TIM interrupts as trigger source from TIM to TOM/ATOM not functional</b>		<b>90</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>GTM_AI.153</b>	<b>TIM: Incorrect data captured to CNTS register when TIM channel operates in mode TPWM or TPIM and CNTS_SEL = 1 and selected CMU_CLK ≠ sys_clk</b>		<b>91</b>
<b>GTM_AI.154</b>	<b>TOM: Incorrect duty cycle in PCM mode (bit reversed mode)</b>		<b>92</b>
<b>GTM_AI.157</b>	<b>CMU: Incorrect AEI status by writing 1 to bit 24 of register CMU_CLK_6/7_CTRL</b>		<b>92</b>
<b>GTM_AI.163</b>	<b>TIM: timeout signaled when TDU unit is reenabled</b>		<b>93</b>
<b>GTM_AI.164</b>	<b>TIM: capturing of data into TIM[i]_CH[x]_CNTS with setting CNTS_SEL=1 not functional in TPWM and TPIM mode</b>		<b>94</b>
<b>GTM_AI.181</b>	<b>TIM: Incorrect signal level bit ECNT[0] in mode TIEM, TPWM, TIPM, TPIM, TGPS</b>		<b>94</b>
<b>GTM_AI.202</b>	<b>(A)TOM: no CCU1 interrupt in case of CM1=0 or 1 and RST_CCU0=1</b>		<b>95</b>
<b>GTM_AI.205</b>	<b>TIM: unexpected CNTS register update in TPWM OSM mode</b>		<b>96</b>
<b>GTM_AI.209</b>	<b>TOM/ATOM: no update of CM0/CM1/CLK_SRC via trigger signal from preceding instance if selected CMU_CLKx is not SYS_CLK</b>		<b>96</b>
<b>GTM_AI.260</b>	<b>TOM/ATOM: Async. update in SOMP mode with CM1=0 and selected CMU clock unequal sys_clk not functional</b>		<b>97</b>
<b>GTM_AI.270</b>	<b>(A)TOM: output signal is postponed one period for the values CM0=1 and CM1&gt;CM0 if CN0 is reset by the trigger of a preceding channel (RST_CCU0=1)</b>		<b>98</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>GTM_AI.298</b>	<b>TOM/ATOM: wrong output behaviour in SOMP oneshot mode when oneshot pulse is triggered by TIM_EXT_CAPTURE(x)</b>		<b>99</b>
<b>GTM_AI.299</b>	<b>TOM/ATOM: wrong output behaviour in SOMP oneshot mode when oneshot pulse is triggered by trig_[x-1]</b>		<b>100</b>
<b>GTM_AI.336</b>	<b>GTM Bus Bridge: Incorrect AEI access execution in case the previous AEI access was aborted with the access timeout abort function</b>		<b>100</b>
<b>GTM_AI.340</b>	<b>TOM/ATOM: Generation of TRIG_CCU0/TRIG_CCU1 trigger signals skipped in initial phase of A/TOM SOMP one-shot mode</b>		<b>101</b>
<b>GTM_AI.341</b>	<b>TOM/ATOM: False generation of TRIG_CCU1 trigger signal in SOMP one-shot mode with OSM_TRIG=1 when CM1 is set to value 1</b>		<b>103</b>
<b>GTM_AI.347</b>	<b>TOM/ATOM: Reset of (A)TOM[i]_CH[x]_CN0 with TIM_EXT_CAPTURE are not correctly synchronized to selected CMU_CLK/CMU_FXCLK</b>		<b>105</b>
<b>GTM_AI.361</b>	<b>IRQ: Missing pulse in single-pulse interrupt mode on simultaneous interrupt and clear event</b>	<b>New</b>	<b>106</b>
<b>GTM_AI.380</b>	<b>(A)TOM: potentially wrong output signal in case of RST_CCU0=1 and CM0=1 on triggered channel in SOMP mode</b>	<b>New</b>	<b>107</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>GTM_AI.408</b>	<b>(A)TOM-RTL: Missing edge on output signal (A)TOM_OUT when CN0 is reset with force update event</b>	<b>New</b>	<b>108</b>
<b>GTM_AI.411</b>	<b>A change of the BRIDGE_MODE register might be delayed indefinitely</b>	<b>New</b>	<b>110</b>
<b>GTM_AI.419</b>	<b>TIM: Potentially wrong capture values</b>	<b>New</b>	<b>111</b>
<b>GTM_AI.429</b>	<b>TIM: Missing glitch detection interrupt event</b>	<b>New</b>	<b>113</b>
<b>GTM_AI.430</b>	<b>TIM: Unexpected increment of filter counter</b>	<b>New</b>	<b>114</b>
<b>GTM_AI.431</b>	<b>TIM: Glitch detection interrupt event of filter is not a single cycle pulse</b>	<b>New</b>	<b>115</b>
<b>GTM_AI.462</b>	<b>(A)TOM: Missing CCU0TC_IRQ interrupt signal</b>	<b>New</b>	<b>116</b>
<b>GTM_TC.009</b>	<b>TBU signals not wired to debug logic</b>		<b>117</b>
<b>GTM_TC.012</b>	<b>Read Access Control by Register ODA</b>		<b>117</b>
<b>IOM_TC.002</b>	<b>Missed or spurious IOM events when pulse length exceeds Event Window counter range</b>		<b>118</b>
<b>IOM_TC.003</b>	<b>Unexpected Event upon Kernel Reset</b>		<b>119</b>
<b>IOM_TC.004</b>	<b>Write to IOM register space when IOM_CLC.RMC &gt; 1</b>		<b>119</b>
<b>MTU_TC.005</b>	<b>Access to MCx_ECCD and MCx_ETRRi while MBIST disabled</b>		<b>120</b>
<b>MTU_TC.007</b>	<b>Error Overflow Indication ECCD.EOV</b>		<b>121</b>
<b>MTU_TC.011</b>	<b>MBIST Bitmap not working for w0 - r1</b>		<b>122</b>
<b>MTU_TC.012</b>	<b>Security of CPU Cache Memories During Runtime is Limited</b>		<b>122</b>
<b>MultiCAN_TC.043</b>	<b>CAN FD: Idle Condition</b>		<b>123</b>

**Table 4 Functional Deviations (cont'd)**

<b>Functional Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<a href="#">MultiCAN_TC.044</a>	<a href="#">CAN FD: Missing Hardsync</a>		<a href="#">124</a>
<a href="#">MultiCAN_TC.045</a>	<a href="#">Simultaneous communication of ISO 11898-1:2015 CAN FD and Non-ISO CAN FD nodes</a>		<a href="#">126</a>
<a href="#">OCDS_TC.038</a>	<a href="#">Disconnecting a debugger without device reset (“hot detach”) may require reading of OCS registers</a>		<a href="#">127</a>
<a href="#">OCDS_TC.042</a>	<a href="#">OTGS capture registers can miss single clock cycle triggers</a>		<a href="#">127</a>
<a href="#">OCDS_TC.043</a>	<a href="#">Read-Modify-Write Bus Transactions to Cerberus Registers</a>		<a href="#">128</a>
<a href="#">PINNING_TC.001</a>	<a href="#">Port functions and pinning tables for TC234* devices without ETH module in TQFP-144 package - Documentation update</a>		<a href="#">128</a>
<a href="#">PLL_ERAY_TC.001</a>	<a href="#">PLL_ERAY Initialization after Cold Power-up or Wake-up from Standby mode</a>		<a href="#">130</a>
<a href="#">PLL_TC.005</a>	<a href="#">PLL Initialization after Cold Power-up or Wake-up from Standby mode</a>	<a href="#">Update</a>	<a href="#">130</a>
<a href="#">PLL_TC.007</a>	<a href="#">PLL Loss of lock when oscillator shaper is used</a>		<a href="#">132</a>
<a href="#">PMC_TC.002</a>	<a href="#">Switch Capacitor Regulator Mode, Frequency Spreading - Documentation Update to Register EVRSDCTRL1</a>		<a href="#">132</a>
<a href="#">PMC_TC.003</a>	<a href="#">Usecase limitation of LDO mode with on chip pass device for SAL devices</a>		<a href="#">134</a>
<a href="#">QSPI_TC.006</a>	<a href="#">Baud rate error detection in slave mode (error indication in current frame)</a>		<a href="#">135</a>
<a href="#">QSPI_TC.017</a>	<a href="#">Slave: Reset when receiving an unexpected number of bits</a>		<a href="#">135</a>

**Table 4 Functional Deviations (cont'd)**

Functional Deviation	Short Description	Change	Page
RESET_TC.005	Indication of Power Fail Events in SCU_RSTSTAT		136
SCU_TC.034	TESTMODE pin shall be held at static high level during LBIST	New	136
SMU_TC.006	OCDS Trigger Bus OTGB during Application Reset		136
SMU_TC.007	Size and Position of Field ACNT in Register SMU_AFCNT		137
SMU_TC.008	Behavior of Action Counter ACNT		138
SMU_TC.010	Transfer to SMU_AD register not triggered correctly		138
SMU_TC.012	Unexpected alarms when registers FSP or RTC are written		139
SRI_TC.003	XBAR_PRIOL/H Register Layout and Reset Values		140

**Table 5 Deviations from Electrical- and Timing Specification**

AC/DC/ADC Deviation	Short Description	Change	Page
ADC_TC.P010	Increased Gain Error ( $EA_{GAIN}$ ) for $T_J < 0^\circ\text{C}$		143
IDD_TC.H001	IPC Limits used in Production Test for IDD Max Power Pattern		143
IEVRSB_TC.P001	Test Condition for $I_{EVRSB}$ (sum of all currents in standby mode) - Data Sheet correction		144
PADS_TC.H004	PN-Junction Characteristics for Pad Type S		144

**Table 5 Deviations from Electrical- and Timing Specification (cont'd)**

<b>AC/DC/ADC Deviation</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>RTH_TC.H001</b>	<b>Thermal characteristics of the package - Footnote update for LF-BGA-292-6 package</b>		<b>144</b>
<b>VDDPPA_TC.H001</b>	<b>Voltage to ensure defined pad states - Footnote update</b>		<b>145</b>

**Table 6 Application Hints**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>ADC_AI.H003</b>	<b>Injected conversion may be performed with sample time of aborted conversion</b>		<b>146</b>
<b>ADC_TC.H011</b>	<b>Bit DCMSB in register GLOBCFG</b>		<b>147</b>
<b>ADC_TC.H014</b>	<b>VADC Start-up Calibration</b>		<b>147</b>
<b>ADC_TC.H015</b>	<b>Conversion Time with Broken Wire Detection</b>		<b>148</b>
<b>ADC_TC.H020</b>	<b>Minimum/Maximum Detection Compares 12 Bits Only</b>		<b>149</b>
<b>ADC_TC.H022</b>	<b>Sample Time Control - Formula</b>		<b>150</b>
<b>ADC_TC.H024</b>	<b>Documentation: Filter control only in registers GxRRC7/GxRRC15</b>		<b>151</b>
<b>ADC_TC.H031</b>	<b>High precision bandgap voltage - documentation update</b>		<b>151</b>
<b>ADC_TC.H038</b>	<b>Multiplexer Diagnostics Connection - Documentation update</b>		<b>152</b>
<b>ADC_TC.H041</b>	<b>Offset address of register GxTRCTR - Correction to table "Registers Overview" in User's Manual</b>	<b>New</b>	<b>152</b>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<a href="#">ADC_TC.H042</a>	<a href="#">Precharging of capacitor CAINSW - Documentation update</a>	New	<a href="#">153</a>
<a href="#">ASCLIN_TC.H001</a>	<a href="#">Bit field FRAMECON.IDLE in LIN slave tasks</a>		<a href="#">153</a>
<a href="#">ASCLIN_TC.H003</a>	<a href="#">Behavior of LIN Autobaud Detection Error Flag</a>		<a href="#">154</a>
<a href="#">ASCLIN_TC.H004</a>	<a href="#">Changing the Transmit FIFO Inlet Width / Receive FIFO Outlet Width</a>		<a href="#">154</a>
<a href="#">ASCLIN_TC.H005</a>	<a href="#">Collision detection error reported twice in LIN slave mode</a>		<a href="#">155</a>
<a href="#">ASCLIN_TC.H006</a>	<a href="#">Sample point position when using three samples per bit - Documentation update</a>	New	<a href="#">156</a>
<a href="#">ASCLIN_TC.H007</a>	<a href="#">Handling TxFIFO and RxFIFO interrupts in single move mode – Documentation update</a>	New	<a href="#">157</a>
<a href="#">ASCLIN_TC.H008</a>	<a href="#">SPI master timing – Additional information to Data Sheet characteristics</a>	New	<a href="#">158</a>
<a href="#">BCU_TC.H001</a>	<a href="#">HSM Transaction Information not captured</a>		<a href="#">159</a>
<a href="#">BROM_TC.H003</a>	<a href="#">Information related to Register FLASH0_PROCOND</a>		<a href="#">159</a>
<a href="#">BROM_TC.H009</a>	<a href="#">Re-Enabling Lockstep via BMHD</a>		<a href="#">159</a>
<a href="#">BROM_TC.H010</a>	<a href="#">Interpretation of value UNIQUE_CHIP_ID_32BIT</a>		<a href="#">160</a>
<a href="#">BROM_TC.H019</a>	<a href="#">CRC32 ethernet polynomial - Footnote correction</a>	New	<a href="#">160</a>
<a href="#">BUS_TC.H001</a>	<a href="#">CPU access latency for TC21x/TC22x/TC23x - Documentation update</a>	New	<a href="#">161</a>

**Table 6 Application Hints (cont'd)**

Hint	Short Description	Change	Page
<a href="#">BUS_TC.H002</a>	<a href="#">Reset value for register XBAR_IDINTEN - Documentation update</a>	New	<a href="#">161</a>
<a href="#">CCU6_AI.H001</a>	<a href="#">Update of Register MCMOUT</a>		<a href="#">162</a>
<a href="#">CCU6_AI.H002</a>	<a href="#">Description of Bit RWHE in Register ISR</a>		<a href="#">162</a>
<a href="#">CCU6_AI.H003</a>	<a href="#">Bit TRPCTR.TRPM2 in Manual Mode - Documentation Update</a>		<a href="#">163</a>
<a href="#">CCU_TC.H001</a>	<a href="#">Clock Monitor Check Limit Values</a>		<a href="#">163</a>
<a href="#">CCU_TC.H002</a>	<a href="#">Oscillator Gain Selection via OSCCON.GAINSEL</a>		<a href="#">164</a>
<a href="#">CCU_TC.H005</a>	<a href="#">References to <math>f_{PLL2}</math>, <math>f_{PLL2\_ERAY}</math> and K3 Divider in User's Manual</a>		<a href="#">164</a>
<a href="#">CCU_TC.H006</a>	<a href="#">Clock Monitor Support - Documentation Update</a>		<a href="#">165</a>
<a href="#">CCU_TC.H007</a>	<a href="#">Oscillator Watchdog Trigger Conditions for ALM3[0]</a>		<a href="#">165</a>
<a href="#">CCU_TC.H010</a>	<a href="#">Oscillator Mode control in register OSCCON - Documentation Update</a>		<a href="#">166</a>
<a href="#">CPU_TC.H006</a>	<a href="#">Store Buffering in TC1.6/P/E Processors</a>		<a href="#">166</a>
<a href="#">CPU_TC.H008</a>	<a href="#">Instruction Memory Range Limitations</a>		<a href="#">169</a>
<a href="#">CPU_TC.H009</a>	<a href="#">Details on CPU Clock Control</a>		<a href="#">169</a>
<a href="#">CPU_TC.H012</a>	<a href="#">Behavior of bit-wise operations on certain peripheral register bits which need to be written back with the same value</a>		<a href="#">170</a>
<a href="#">CPU_TC.H014</a>	<a href="#">ACCEN* Protection for Write Access to Safety Protection Registers - Documentation Update</a>	Update	<a href="#">172</a>
<a href="#">CPU_TC.H015</a>	<a href="#">Register Access Modes for Safety Protection Registers - Documentation Update</a>		<a href="#">172</a>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<a href="#">CPU_TC.H017</a>	<a href="#">MSUB.Q does not match MUL.Q+SUB - Documentation Update</a>		<a href="#">172</a>
<a href="#">DAP_TC.H002</a>	<a href="#">DAP client_blockread in Combination with TGIP and all Parcels with CRC6</a>		<a href="#">174</a>
<a href="#">DAP_TC.H003</a>	<a href="#">Not acknowledged DAP telegrams in noisy environments</a>		<a href="#">174</a>
<a href="#">DMA_TC.H002</a>	<a href="#">Bit CHCSRz.BUFFER can be toggled when not in Double Buffer Mode</a>		<a href="#">175</a>
<a href="#">DMA_TC.H004</a>	<a href="#">Transaction Request Lost upon software trigger with pattern match</a>		<a href="#">175</a>
<a href="#">DMA_TC.H005</a>	<a href="#">Linked List Transfer leading to loading of non-Linked List TCS causes corruption</a>		<a href="#">176</a>
<a href="#">DMA_TC.H006</a>	<a href="#">Clearing of HTRE when DMA channel is configured for Single Mode</a>		<a href="#">176</a>
<a href="#">DMA_TC.H007</a>	<a href="#">Selecting the Priority for DMA Channels</a>		<a href="#">177</a>
<a href="#">DMA_TC.H008</a>	<a href="#">Transaction Request State</a>		<a href="#">178</a>
<a href="#">DMA_TC.H009</a>	<a href="#">Resetting Bits ICH and IPM in register CHCSRz</a>		<a href="#">178</a>
<a href="#">DMA_TC.H010</a>	<a href="#">Calculation of DMA Address Checksum for DMA read moves to Cacheable Addresses</a>		<a href="#">179</a>
<a href="#">DMA_TC.H011</a>	<a href="#">DMA_ADICRz.SHCT - Reserved Values</a>		<a href="#">179</a>
<a href="#">DMA_TC.H012</a>	<a href="#">TCS Update in Halt State</a>		<a href="#">180</a>
<a href="#">DMA_TC.H013</a>	<a href="#">MExSR.WS and MExSR.RS Status Bits</a>		<a href="#">180</a>
<a href="#">DMA_TC.H016</a>	<a href="#">DMARAM ECC Error Disable</a>		<a href="#">181</a>
<a href="#">DMA_TC.H017</a>	<a href="#">DMA Channel Request Control - Documentation Update</a>		<a href="#">181</a>
<a href="#">DTS_TC.H001</a>	<a href="#">Update of Bit DTSSTAT.BUSY</a>		<a href="#">181</a>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<a href="#">ENDINIT_TC.H001</a>	<a href="#">Endinit Protection for Registers KRST0, KRST1, KRSTCLR</a>		<a href="#">182</a>
<a href="#">FLASH_TC.H007</a>	<a href="#">Advice for using Suspend and Resume</a>		<a href="#">182</a>
<a href="#">FLASH_TC.H008</a>	<a href="#">Understanding Flash Retention/Endurance Figures in the Data Sheet</a>		<a href="#">184</a>
<a href="#">FLASH_TC.H022</a>	<a href="#">Flash Wait State configuration</a>	New	<a href="#">185</a>
<a href="#">FlexRay_AI.H004</a>	<a href="#">Only the first message can be received in External Loop Back mode</a>		<a href="#">185</a>
<a href="#">FlexRay_AI.H005</a>	<a href="#">Initialization of internal RAMs requires one eray_bclk cycle more</a>		<a href="#">186</a>
<a href="#">FlexRay_AI.H006</a>	<a href="#">Transmission in ATM/Loopback mode</a>		<a href="#">186</a>
<a href="#">FlexRay_AI.H007</a>	<a href="#">Reporting of coding errors via TEST1.CERA/B</a>		<a href="#">186</a>
<a href="#">FlexRay_AI.H009</a>	<a href="#">Return from test mode operation</a>		<a href="#">187</a>
<a href="#">FlexRay_AI.H011</a>	<a href="#">Behavior of interrupt flags in FlexRay™ Protocol Controller (E-Ray)</a>		<a href="#">187</a>
<a href="#">FlexRay_TC.H002</a>	<a href="#">Initialization of E-Ray RAMs</a>	Update	<a href="#">188</a>
<a href="#">FlexRay_TC.H004</a>	<a href="#">Bit WRECC in register TEST2 has no function</a>	New	<a href="#">190</a>
<a href="#">FPI_TC.H002</a>	<a href="#">Write Access to Register ACCEN1</a>		<a href="#">190</a>
<a href="#">GPT12_TC.H001</a>	<a href="#">Timer T5 Run Bit T5R - Documentation Correction</a>		<a href="#">191</a>
<a href="#">GPT12_TC.H002</a>	<a href="#">Bits TxUD and TxUDE in incremental interface mode - Additional information</a>	New	<a href="#">191</a>
<a href="#">GTM_TC.H004</a>	<a href="#">Correction to Bit Fields GTM_TIMi_IN_SRC.VAL_x</a>		<a href="#">192</a>
<a href="#">GTM_TC.H005</a>	<a href="#">External Capture in TIM Pulse Integration Mode (TPIM)</a>		<a href="#">192</a>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>GTM_TC.H007</b>	<b>GTM to CAN Timer Triggers</b>		<b>193</b>
<b>GTM_TC.H009</b>	<b>TIM0 Channel x Input Selection - Mapping for QFP-80 and QFP-100 Packages</b>		<b>194</b>
<b>GTM_TC.H011</b>	<b>First CM0 updates in case of SR0=1 and (A)TOM used as Triggered Channel</b>		<b>197</b>
<b>GTM_TC.H014</b>	<b>Synchronous Bridge Mode Restrictions</b>		<b>197</b>
<b>GTM_TC.H015</b>	<b>Register TIMi_CHx_CTRL - Correction to Register Image</b>		<b>198</b>
<b>GTM_TC.H020</b>	<b>GTM can cause unintended bus errors after enabling when SPB or GTM frequency is very low</b>		<b>198</b>
<b>GTM_TC.H025</b>	<b>Field TOCTRL in register GTM_TIM0_CHx_CTRL - Documentation correction</b>	<b>New</b>	<b>199</b>
<b>INT_TC.H004</b>	<b>Corrections to the Interrupt Router Documentation</b>		<b>199</b>
<b>IOM_TC.H001</b>	<b>How to clear the IOM_LAMEWCm register</b>		<b>200</b>
<b>IOM_TC.H002</b>	<b>IOM Clock Control</b>		<b>200</b>
<b>IOM_TC.H003</b>	<b>Configuration of LAMCFG.IVW and LAMEWS.THR</b>		<b>202</b>
<b>IOM_TC.H004</b>	<b>Behavior of LAMEWCn.CNT when LAMEWSn.THR is 0</b>		<b>203</b>
<b>IOM_TC.H006</b>	<b>ACCEN* Protection for Write Access to IOM Registers</b>		<b>203</b>
<b>IOM_TC.H007</b>	<b>Write Access to FPESR</b>		<b>204</b>
<b>LBIST_TC.H004</b>	<b>Update reset behavior of LBISTCTRL2 register - Additional information</b>	<b>New</b>	<b>204</b>
<b>LMU_TC.H002</b>	<b>On-the-fly BBB:SRI clock ratio switching</b>		<b>205</b>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>LMU_TC.H003</b>	<b>Function of Bit MEMCON.PMIC (Protection Bit for Memory Integrity Control Bit)</b>		<b>205</b>
<b>MTU_TC.H003</b>	<b>AURIX™ Memory Tests using the MTU</b>		<b>206</b>
<b>MTU_TC.H004</b>	<b>Handling the Error Tracking Registers ETRR</b>		<b>207</b>
<b>MTU_TC.H005</b>	<b>Handling SRAM Alarms</b>		<b>208</b>
<b>MTU_TC.H006</b>	<b>Alarm Propagation to SMU via Error Flags in MCx_ECCD</b>		<b>209</b>
<b>MTU_TC.H008</b>	<b>Memory Controllers for DSPR</b>		<b>210</b>
<b>MTU_TC.H009</b>	<b>Reset Value for Register ECCD</b>		<b>211</b>
<b>MTU_TC.H010</b>	<b>Register MCONTROL - Bit Field Res4</b>		<b>212</b>
<b>MTU_TC.H011</b>	<b>Access Protection for Memory Control Registers</b>		<b>212</b>
<b>MTU_TC.H012</b>	<b>Kernel Reset triggers Reset of MBIST Registers</b>		<b>212</b>
<b>MTU_TC.H014</b>	<b>Access to SRAM while MTU operations are underway</b>		<b>213</b>
<b>MultiCAN_AI.H005</b>	<b>TxD Pulse upon short disable request</b>		<b>214</b>
<b>MultiCAN_AI.H006</b>	<b>Time stamp influenced by resynchronization</b>		<b>214</b>
<b>MultiCAN_AI.H007</b>	<b>Alert Interrupt Behavior in case of Bus-Off</b>		<b>215</b>
<b>MultiCAN_TC.H003</b>	<b>Message may be discarded before transmission in STT mode</b>		<b>215</b>
<b>MultiCAN_TC.H004</b>	<b>Double remote request</b>		<b>216</b>
<b>MultiCAN_TC.H007</b>	<b>Oscillating CAN Bus may Disable the CAN Interface</b>		<b>216</b>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<a href="#">MultiCAN_TC.H008</a>	<a href="#">Changes due to CAN FD protocol ISO 11898-1:2015</a>		<a href="#">217</a>
<a href="#">MultiCAN_TC.H009</a>	<a href="#">Limitation on Secondary Sample Point (SSP) Position (ISO CAN FD nodes only)</a>		<a href="#">221</a>
<a href="#">MultiCAN_TC.H010</a>	<a href="#">Limitation on maximum SJW Range for CAN FD Data Phase (ISO CAN FD nodes only)</a>		<a href="#">222</a>
<a href="#">MultiCAN_TC.H011</a>	<a href="#">Transmitter Delay Compensation Behaviour (CAN FD only)</a>		<a href="#">223</a>
<a href="#">MultiCAN_TC.H012</a>	<a href="#">Delayed time triggered transmission of frames</a>		<a href="#">224</a>
<a href="#">OCDS_TC.H010</a>	<a href="#">JTAG requires two initial clock cycles after PORST</a>		<a href="#">224</a>
<a href="#">OCDS_TC.H012</a>	<a href="#">Minimum Hold Time for Inputs OCDS_TGix</a>		<a href="#">225</a>
<a href="#">OCDS_TC.H019</a>	<a href="#">System or Application Reset while OCDS and lockstep monitoring are enabled</a>	New	<a href="#">225</a>
<a href="#">PACKAGE_TC.H008</a>	<a href="#">Exposed pad dimensions and package outlines for QFP packages - Updates to TC23x Data Sheet</a>	New	<a href="#">226</a>
<a href="#">PLL_ERAY_TC.H002</a>	<a href="#">Correction in Figure "PLL_ERAY Block Diagram"</a>		<a href="#">227</a>
<a href="#">PMC_TC.H001</a>	<a href="#">Check for permanent Overvoltage during Power-up</a>		<a href="#">228</a>
<a href="#">PMC_TC.H004</a>	<a href="#">Selecting the WUT Clock Divider</a>		<a href="#">228</a>
<a href="#">PMS_TC.H002</a>	<a href="#">Sensitivity to supply voltage ripple during start-up</a>		<a href="#">229</a>
<a href="#">PMS_TC.H008</a>	<a href="#">Interaction of interrupt and power management system - Additional information</a>	New	<a href="#">231</a>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
<b>PMU_TC.H002</b>	<b>Impact of Application Reset on register FLASH0_FCON</b>		<b>233</b>
<b>PORTS_TC.H006</b>	<b>Using P33.8 while SMU is disabled</b>		<b>234</b>
<b>PORTS_TC.H016</b>	<b>Oscillating signal may enable DXCPL and reconfigure the functionality of the port pins P14.0 and P14.1</b>	<b>New</b>	<b>235</b>
<b>QSPI_TC.H005</b>	<b>Stopping Transmission in Continuous Mode</b>		<b>235</b>
<b>QSPI_TC.H006</b>	<b>Corrections to Figures “QSPI - Frequency Domains” and “Phase Duration Control, Overview”</b>		<b>236</b>
<b>QSPI_TC.H007</b>	<b>RXFIFO Overflow Bit Behavior in Slave Mode</b>		<b>237</b>
<b>QSPI_TC.H008</b>	<b>Details of the Baud Rate and Phase Duration Control - Documentation update</b>		<b>237</b>
<b>QSPI_TC.H009</b>	<b>Dummy frame required after changing SCLK polarity and phase in three wire mode</b>	<b>New</b>	<b>238</b>
<b>RESET_TC.H002</b>	<b>Unexpected SMU Reset Indication in SCU_RSTSTAT</b>		<b>238</b>
<b>RESET_TC.H003</b>	<b>Usage of the Prolongation Feature for ESR0 as Reset Indicator Output</b>		<b>239</b>
<b>RESET_TC.H004</b>	<b>Effect of Power-on and System Reset on DSPR</b>		<b>240</b>
<b>SCU_TC.H009</b>	<b>LBIST Influence on Pad Behavior</b>		<b>240</b>
<b>SCU_TC.H010</b>	<b>LBIST Signature Depends on Debug Interface Configuration</b>		<b>241</b>
<b>SCU_TC.H013</b>	<b>Correction to Register References in Chapter “Watchdog Timers”</b>		<b>241</b>

**Table 6 Application Hints (cont'd)**

<b>Hint</b>	<b>Short Description</b>	<b>Change</b>	<b>Page</b>
SCU_TC.H014	Reset Value of Bit Field IOCR.PC1 - Control for Pin ESR1		242
SENT_TC.H003	First Write Access to Registers FDR and TPD after ENDINIT Status Change		243
SENT_TC.H004	Short Serial Message - Figure Correction		244
SENT_TC.H005	Interface Connections of the SENT Module - Documentation Correction		244
SMU_TC.H001	Write all bit fields of SMU_PCTL with one write access		245
SMU_TC.H005	Correction to Figure "SMU Register Map"		245
SMU_TC.H006	Description of Bit EFRST in Register SMU_AGC		246
SMU_TC.H007	SPB Bus Control Unit (SBCU) Alarm Signalling to SMU		246
SMU_TC.H009	Alarm Table Corrections		247
SMU_TC.H010	Clearing individual SMU flags: use only 32-bit writes		252
SMU_TC.H013	Increased Fault Detection for SMU Bus Interface (SMU_CLC Register)		253
SMU_TC.H014	Unintended short pulse on FSP pins in Time switching or Dual-rail mode		254
SRI_TC.H001	Using LDMST and SWAPMSK.W instructions on SRI mapped Peripheral Registers (range 0xF800 0000-0xFFFF FFFF)		254
STM_TC.H001	Effect of kernel reset on interrupt outputs STMIR0/1		255
STM_TC.H002	Access Protection for STM Control Registers		255

**Table 6** Application Hints (cont'd)

Hint	Short Description	Change	Page
STM_TC.H003	Suspend control for STMx - Documentation Update		256
STM_TC.H004	Access to STM registers while STMDIV = 0		257

## 2 Functional Deviations

### **ADC\_AI.016 No Channel Interrupt in Fast Compare Mode with GLOBRES**

In fast compare mode, the compare value is taken from bitfield RESULT of the selected result register and the result of the comparison is stored in the respective bit FCR.

A channel event can be generated when the input becomes higher or lower than the compare value.

In case the global result register GLOBRES is selected, the comparison is executed correctly, the target bit is stored correctly, source events and result events are generated, but a channel event is not generated.

#### **Workaround**

If channel events are required, choose a local result register GxRESy for the operation of the fast compare channel.

### **ADC\_TC.068 Effect of VAGND Cross Coupling on Conversion Result**

Due the implementation of the clock dividers as fractional dividers, a statistical phase shift of one  $f_{VADC}$  clock can occur between the operation of different converter groups. If the last  $f_{VADC}$  clock of the sample phase of a converter group Gx coincides with the first  $f_{VADC}$  clock of a conversion step of (one or more) other converter groups Gy, the Total Unadjusted Error (TUE) of the conversion result of Gx is increased due to cross coupling via VAGND.

For TC26x, TC23x, TC22x, and TC21x, the TUE is increased up to  $\pm 25 \text{ LSB}_{12}$

#### **Workarounds - Introduction**

Workaround 1..3 may be used with any device step.

Workaround 4 can only be used with TC21x, TC22x, TC23x  $\geq$  step AB.

### Workaround 1

Synchronize the trigger events of different converter groups as follows:

- Operate the arbiters and the analog parts of the VADC at the same clock frequency, i.e. select the divider factors DIVA and DIVD in register GLOBCFG such that  $f_{ADCD} = f_{ADCI}$  for all converter groups:
  - Note: As  $f_{ADCD} = f_{VADC}/4$  with the maximum divider (DIVD = 3), this implies that  $f_{VADC} = f_{SPB}$  must be limited to 80 MHz to achieve  $f_{ADCD} = f_{ADCI}$  with the error limits specified for  $f_{ADCI} = 20$  MHz in the Data Sheet.
- Enlarge the length of an arbitration round to a minimum of 16 arbitration slots (i.e. bit field GxARBCFG.ARBRND  $\geq 2$  for any x).
- Select the conversion time (including sample time) of the longest conversion of any group Gx to be shorter than two arbitration rounds. This ensures that all converters are idle when the arbiters have determined the next conversion request.
- Synchronize the digital and the analog clock by switching off/on the Module Disable Request bit, i.e. set CLC.DISR = 1<sub>B</sub> and then CLC.DISR = 0<sub>B</sub>.
- Initiate the start-up calibration by setting bit GLOBCFG.SUCAL = 1<sub>B</sub> (mandatory after switching off/on VADC clocks via CLC.DISR).

### Workaround 2

Ensure that conversions never overlap for any two converter groups Gx and Gy. This may be achieved under software control, or by exclusively using the VADC background request source.

For this workaround, no restrictions apply on clock and arbitration round settings.

### Workaround 3

Use the converters within a synchronization group in master/slave configuration, such that they are synchronized for parallel sampling, triggered by one common master. In this case, the cross coupling effect will not occur as long as only one synchronization group is performing conversions.

For devices that support more than one synchronization group, operate the synchronization groups in an interleaving manner.

For this workaround, no restrictions apply on clock and arbitration round settings.

#### Workaround 4

To avoid the cross coupling effect, this device step (see “Workarounds - Introduction” above) supports selection of signal CCU6061\_TRIG1 to synchronize the start of the converter groups to a raster of  $1/f_{\text{ADCI}}$  (e.g.  $5/f_{\text{SPB}} = 50 \text{ ns}$  @  $f_{\text{SPB}} = 100 \text{ MHz}$  and  $f_{\text{ADCI}} = 20 \text{ MHz}$ , or  $4/f_{\text{SPB}} = 64 \text{ ns}$  @  $f_{\text{SPB}} = 62.5 \text{ MHz}$  and  $f_{\text{ADCI}} = 12.5 \text{ MHz}$ ). The resulting jitter (delay from trigger to start of conversion) is thus limited to max.  $1/f_{\text{ADCI}}$ .

For this workaround, either CCU60\_T13 or CCU61\_T13 is configured (reserved) to provide the synchronization signal. The selection is performed via bit field TRIG1SEL in register CCU60\_MOSEL:

- TRIG1SEL =  $000_{\text{B}}$ : signal CCU60\_COUT63 from CCU60\_T13 is selected
- TRIG1SEL =  $001_{\text{B}}$ : signal CCU61\_COUT63 from CCU61\_T13 is selected

The synchronization signal is enabled inside the VADC module by setting bit GLOBCFG.DCMSB =  $1_{\text{B}}$ . The default function of this bit (DCMSB =  $0_{\text{B}}$ : one clock cycle for MSB conversion step) is hardwired and thus stays unaffected.

The following examples describe the initialization of CCU60 or CCU61, respectively, to provide a 20 MHz synchronization signal @  $f_{\text{SPB}} = 100 \text{ MHz}$ :

#### Example for CCU60 initialization

```
CCU60_CLC = 0x0;           // enable CCU60 kernel
CCU60_T13PR = 0x4;        // 4+1 clock periods with ..
CCU60_CC63SR = 0x1;       // duty cycle 40 ns low / 10 ns high
CCU60_PSLR |= 0x0080;     // passive state level of COUT63 = 1
CCU60_MODCTR |= 0x8000;   // ECT130 = 1 enables T13 output
                           // (CC63ST -> COUT63)
CCU60_TCTR4 |= 0x4200;    // set bit T13STR and T13RS ..
                           // to enable shadow transfer and start T13
CCU60_MOSEL &= 0x1C7;     // CCU6061_TRIG1 is CCU60_COUT63
```

#### Example for CCU61 initialization

```
CCU61_CLC = 0x0;           // enable CCU61 kernel
```

*Note: In case an application only uses kernel CCU61, ensure that kernel CCU60 is also clocked until register CCU60\_MOSEL is configured.*

```
CCU60_CLC = 0x0;           // ensure CCU60 kernel is clocked
                          // until CCU60_MOSEL is configured
CCU61_T13PR = 0x4;        // 4+1 clock periods with ..
CCU61_CC63SR = 0x1;       // duty cycle 40 ns low / 10 ns high
CCU61_PSLR |= 0x0080;     // passive state level of COUT63 = 1
CCU61_MODCTR |= 0x8000;   // ECT130 = 1 enables T13 output
                          // (CC63ST -> COUT63)
CCU61_TCTR4 |= 0x4200;    //set bit T13STR and T13RS ..
                          // to enable shadow transfer and start T13
CCU60_MOSEL |= 0x8;       // CCU6061_TRIG1 is CCU61_COUT63
```

### **ASCLIN\_TC.004 SLSO in SPI mode still active after module disable**

It is expected that in SPI mode, after module disable, the Slave Select Output signal SLSO should be in idle state according to configuration of Slave Polarity in Synchronous mode (IOCR.SPOL).

However, in this design step, when the module is disabled, the Slave Select Output signal SLSO is always 0 (low) independent of IOCR.SPOL, i.e., it is still active even when IOCR.SPOL = 1<sub>B</sub>.

#### **Workaround**

Before disabling the ASCLIN module, set SLSO to the desired level in the corresponding Port control registers.

### **ASCLIN\_TC.005 Unjustified collision detection error in half-duplex SPI mode**

In Half Duplex SPI mode, when collision detection is enabled and the number of stop bits in SPI frame is configured as any value from 1 to 7 in FRAMECON.STOP, a Collision Error (FLAGS.CE) is triggered during the trailing phase (i.e., during stop bits), although RX and TX signal are identical.

### Workaround

In half-duplex SPI mode, set FRAMECON.STOP = 0 if trailing phase is irrelevant, or ignore/disable collision error if FRAMECON.STOP > 0.

### **ASCLIN\_TC.006 Unjustified response timeout in LIN slave mode**

When ASCLIN is configured as LIN slave and Response timeout is configured as DATCON.RM = 1<sub>B</sub>, Response timeout is triggered even when an incomplete LIN Header frame is received. The timeout counter runs further after Header timeout detection without reset and triggers Response Timeout when it reaches the Response Timeout Threshold value defined by DATCON.RESPONSE.

### Workaround

Ignore the Response Timeout which comes directly after a Header Timeout has occurred and before the next break is detected.

### **ASCLIN\_TC.007 Break Detected in LIN Frames in Soft Suspend mode**

When ASCLIN has entered Soft Suspend mode (OCS.SUS = 0x2), it still detects a Break Field in LIN frames and triggers an interrupt if enabled (FLAGSENABLE.BDE = 1<sub>B</sub>).

### Workaround

Ignore a detected break event when the module has been soft-suspended (e.g. set FLAGSENABLE.BDE = 0<sub>B</sub> when using soft suspend mode).

### **ASCLIN\_TC.008 Response timeout in LIN Mode in case of header only**

In LIN (Master/Slave) mode, when Header Only (DATCON.HO = 1<sub>B</sub>) is configured, Response timeout could occur even though no Response frame is expected.

### Workaround

To avoid the unwanted interrupt, disable the interrupt on Response Timeout by `FLAGSENABLE.RTE = 0B` whenever Header Only (`DATCON.HO = 1B`) is configured.

### **ASCLIN\_TC.009 RFL flag set in Buffer Mode when Receive FIFO Inlet is disabled**

When RXFIFO is configured in Buffer Mode (`RXFIFOCON.BUF = 1B`) and Receive FIFO Inlet is disabled (`RXFIFOCON.ENI = 0B`), the receive FIFO level flag is set (`FLAGS.RFL = 1B`) even though RXFIFO is not filled with new incoming data.

### Workaround

To avoid the unwanted Receive FIFO Level interrupt, disable it by setting `FLAGSENABLE.RFLE = 0B` whenever Receive FIFO Inlet is disabled (`RXFIFOCON.ENI = 0B`),

### **ASCLIN\_TC.010 Flush of TXFIFO leads to frame transmission**

When the TXFIFO is flushed (`TXFIFIOCON.FLUSH = 1B`), it triggers transmission of a frame in the following corner case:

- Starting condition:
  - TXFIFO is not empty and `TXFIFIOCON.ENO = 0B`
- Triggering condition:
  - Write to TXFIFIOCON with both `TXFIFIOCON.FLUSH = 1B` and `TXFIFIOCON.ENO = 1B`

### Workaround

Do not flush TXFIFO and change bit TXFIFIOCON.ENO from `0B` to `1B` in one single write to TXFIFIOCON if TXFIFO is not empty.

**ASCLIN\_TC.012 Recover sequence after timeout in LIN master mode**

Due to an internal state machine problem, unexpected behavior will occur in the scenario described below.

**Expected behavior**

In the LIN master mode, the LIN state machine is supposed to abort its current processing and return to idle state when a header or response timeout occurs.

**Observed behavior**

If a timeout error occurred, no further LIN frames are transmitted.

**Exception**

In rare cases, when the timeout error occurs in the LIN state machine at the same time as a soft suspend request to the ASCLIN module, the LIN state machine returns to idle state and the recover sequence is as described in the User's Manual.

**Workaround**

After a header or response timeout has been detected, set field FRAMECON.MODE to INIT (00<sub>B</sub>) and then to LIN mode (11<sub>B</sub>), as described in section "Abort sequence" in the ASCLIN chapter of the User's Manual.

**BROM\_TC.008 Sporadic Power-on Reset after Wake-up from Standby Mode**

On a wake-up from Standby mode, the Standby RAM redundancy installation procedure is executed. In case there is a sporadic Power-on reset in a time window between 600  $\mu$ s - 1 ms after Standby mode wake-up, it can happen that the application data stored in specific Standby RAM cells are overwritten.

*Note: This effect can occur only on devices where non-zero data are stored in CPU0 DSPR at locations D000 2000<sub>H</sub> to D000 203F<sub>H</sub> by the Startup Software (SSW) after cold power-on (see section "Preparation before to enter Stand-by mode" in the BootROM chapter of the User's Manual).*

*Only CPU0 DSPR Standby RAM is affected, EMEM in ADAS or ED devices is not affected.*

## Workarounds

1. Calculate CRC over critical Standby RAM data and store result before Standby mode entry. On a consequent wake-up, CRC of the critical data shall be carried out. The CRC is a general recommended measure for improved robustness of Standby RAM handling.  
Or / and
2. Keep a copy of the critical data at a second location in Standby RAM. On wake-up, compare data from both locations to ascertain their integrity.

## **BROM\_TC.015 DSPR Data Integrity after Wake-up from Standby Mode**

The CPU0 DSPR memory can be used as Standby RAM to save data required after resuming from standby mode.

Applications using DSPR powered in standby mode (to save data) may face data loss within DSPR after wake-up on TC23x product variants.

*Note: The following types of applications are NOT affected:*

- Applications which do not power DSPR during standby mode
- Applications initializing the entire used DSPR address range with 32-bit word accesses after PORST and after wake-up from standby mode
- Applications which do not use the standby feature at all.

## Detailed Problem Description

Aurix™ TC2xx devices use repair cells to substitute defective SRAM cells in the DSPR address range. Repair cells are not powered during standby mode and therefore will lose data.

After cold power-on reset, the start-up software (SSW) creates a list of repaired DSPR addresses which is used to restore data to repair cells after wake-up from standby mode (see chapter “RAM overwrite during start-up” in the BootROM chapter of the User’s Manual).

On affected devices, the list of repaired RAM addresses is not created correctly, with the following consequences:

- After wake-up from standby mode, read accesses to repaired addresses may deliver incorrect content, and in most cases there will be a bus error if the  $V_{DDP3}$  voltage domain is unpowered during standby mode
- Some RAM locations - incorrectly treated by the SSW as defective - may be overwritten, or the previous contents (before standby mode entry) may be lost, respectively.

## Scope

The failure rate depends on the application impact of up to 16 corrupted data words which are randomly distributed in the DSPR address range. About 1% of the devices shipped use between 1 and 16 repaired words in the DSPR address range and are therefore affected.

## Identification of affected devices

Affected applications can recognize the problem using the following sequence:

- Calculate a checksum on the DSPR address range (used by the application) and store it in DSPR before entering standby mode
- After wake-up, perform a plausibility check (re-calculate checksum and compare it with stored checksum value)
- If the plausibility check passes: the device is not affected by the problem.

## Workaround

None

## **CPU TC.123 Data Corruption possible when CPU GPR accesses made via SRI slave with CPU running**

Data corruption may occur when another master accesses a TriCore CPU's General Purpose Registers (GPRs) via its SRI slave port whilst the CPU is running (i.e. not Idle, Halted or Suspended). The TriCore GPRs are A0-A15 and D0-D15. The scenarios in which data corruption may occur are different for the TC1.6P and TC1.6E processors as described below.

TC1.6P - Data corruption may occur when one of the CPU GPRs is **written** via the SRI slave port whilst the CPU is running. Both AGPR and DGPR writes may be affected.

TC1.6E - Data corruption may occur when one of the CPU Address GPRs (A0-A15) is **read** via the SRI slave port whilst the CPU is running. However, data corruption can only occur when the slave AGPR read interacts with the execution of a specific form of store instruction. The store instructions affected by this issue are ST.A and ST.DA, where the address register to be stored is modified by the addressing mode of the store instruction. For example:

```
ST.A [+A0], A0
```

However, such store instructions are architecturally undefined and should not be being used. In the case of this errata all data written to memory by this store instruction may be corrupted.

### Workaround

Writes to a CPU's GPRs via its SRI slave port must never be performed whilst the CPU is running. If it is necessary for an external master to write to a CPU's GPR then that CPU must first be placed in Idle, Halt or Suspend mode.

If it is necessary for an external master to read a TC1.6E CPU's AGPR whilst that CPU is running then store instructions of the form above (where any source register is modified by the addressing mode of the store instruction) are not allowed.

### **CPU TC.127 Pending Interrupt Priority Number PIPN in Register ICR**

In the TriCore Architecture Manual, it is described for the Pending Interrupt Priority Number ICR.PIPN that it is reset to 0x0 in case there is no request pending.

However, the AURIX™ hardware implementation behaves differently, as the value of PIPN is not changed after the interrupt is serviced in case there is no further request pending.

## **CPU\_TC.132 Unexpected PSW values used upon Fast Interrupt entry**

Under certain conditions, unexpected PSW values may be used during the first instructions of an interrupt handler, if the interrupt has been taken as a fast interrupt. For a description of fast interrupts, see the “CPU Implementation-Specific Features” section of the relevant User’s Manual.

When the problem occurs, the first instructions of the interrupt handler may be executed using the PSW state from the end of the previous exception handler, rather than that which is being loaded by the fast interrupt entry sequence. The TC1.6E, TC1.6P and TC1.6.2P processors are all affected by this problem as follows:

- TC1.6E (in TC21x..TC27x): Only the first instruction of the ISR is affected.
- TC1.6P (in TC26x..TC29x), TC1.6.2P (in TC3xx): Up to 4 instructions at the start of the ISR may be affected. However, if the following precondition is not met, then there is no issue for these processor variants:
  - A11 must point to the first instruction of the fast interrupt handler at the end of the previous exception handler, i.e. the return value from the previous exception must be pointing to the very first instruction of the new interrupt handler. Note that this case should not occur normally, unless software updates the A11 register to a value corresponding to the start of an interrupt handler.

## **Workarounds**

### **Workaround 1**

When the PSW fields PSW.PRS, PSW.S, PSW.IO or PSW.GW need to be changed in an exception handler, the change should be wrapped in a function call.

```
_exception_handler:  
    CALL _common_handler  
    RFE
```

```
_common_handler:  
    MOV.U d0, #0x0380  
    MTCR #(PSW), d0    // PSW.IO updated to User-0 mode
```

...  
RET

Note that this workaround assumes `SYSCON.TS == SYSCON.IS` such that the workaround functions correctly for both traps and interrupts. If this is not the case it is possible for bus accesses to use an incorrect master Tag ID, potentially resulting in an access to be incorrectly allowed, or an unexpected alarm to be generated. In this case it should be ensured that for all interrupt handlers the potentially affected instructions do not produce bus accesses.

## Workaround 2

Do not use any instructions dependent upon PSW settings (e.g. BISR or ENABLE, dependent on PSW.IO) as the first instruction of an ISR in TC1.6E, or as one of the first 4 instructions in an ISR for TC1.6P or TC1.6.2P.

*Note: The workarounds need to be applied in TC1.6P and TC1.6.2P only in case software modifies the A11 register in an exception handler, as described in the preconditions above.*

## **DAP\_TC.002 DAP client\_blockread has Performance issue in Specific Operation Modes**

For achieving the highest block read bandwidth, the following word is already read chip internally while a word is transmitted on DAP. This read ahead is under certain conditions disabled in the case that the “All parcels with CRC6” bit is set in the telegram. In this case the distance between the reply parcels becomes significantly longer, due to the missing read ahead. This effect occurs also in Wide Mode.

The data values in the parcels are always correct, it is just a performance issue.

### Workaround

Don't use the “All parcels with CRC6” option, use “Read CRCup” instead.

This mode is anyway better in terms of performance for larger blocks (no CRC6 overhead for each parcel) and data protection (32 bit CRC). For a few words, the impact of this performance issue might be tolerable. For the first word a read ahead is not possible anyway.

**DAP\_TC.003 DAP CRC32 definition and algorithm**

The DAP CRC32 algorithm is different from the IEEE 802.3 Ethernet CRC.

**Workaround**

Use the following (VHDL) algorithm for each incoming data bit. The CRC32 value is initialized with all ones.

In Wide Mode the function is called for both DAP data bits in each DAP0 clock cycle.

```
subtype crc32_t is std_ulogic_vector(31 downto 0);
function calc_crc32_f(crc_now : crc32_t;
                    bit_new : std_ulogic)
    return crc32_t is
    variable crc : crc32_t;
begin
    crc(31 downto 1) := crc_now(30 downto 0);
    crc(0) := bit_new xor crc_now(31);
    crc(1) := bit_new xor crc_now(0) xor crc_now(31);
    crc(2) := bit_new xor crc_now(1) xor crc_now(31);
    crc(4) := bit_new xor crc_now(3) xor crc_now(31);
    crc(5) := bit_new xor crc_now(4) xor crc_now(31);
    crc(7) := bit_new xor crc_now(6) xor crc_now(31);
    crc(8) := bit_new xor crc_now(7) xor crc_now(31);
    crc(10) := bit_new xor crc_now(9) xor crc_now(31);
    crc(11) := bit_new xor crc_now(10) xor crc_now(31);
    crc(12) := bit_new xor crc_now(11) xor crc_now(31);
    crc(16) := bit_new xor crc_now(15) xor crc_now(31);
    crc(22) := bit_new xor crc_now(21) xor crc_now(31);
    crc(23) := bit_new xor crc_now(22) xor crc_now(31);
    crc(26) := bit_new xor crc_now(25) xor crc_now(31);
    return crc;
end calc_crc32_f;
```

## **DAP\_TC.004 DAP client\_blockwrite telegram with CRC6 and CRC32 protection options**

*Note: This problem is only relevant for tool development, not for application development.*

When issuing a DAP client\_blockwrite telegram from the tool to the device several CRC protection options are available, namely CRC6 and CRC32.

### **Expected Behavior**

- For CRC6 the expected behavior is:
  - (1) A CRC6 will be appended to the reply of only the last parcel of the telegram.
  - (2) An optional CRC6 can be appended to the devices “single startbit response” by setting DAPISC.RC6.
- For CRC32 the expected behavior is:
  - (3) The telegram can optionally send the CRCdown value as the last parcel.

### **Actual Implementation**

- For the actual implementation the CRC6 slightly differs as follows:
  - (1) The CRC6 of the last parcel will be erroneous if DAPISC.RC6 is set or if the CRCdown option is enabled.
  - (2) If DAPISC.RC6 = 1<sub>B</sub>, an unintentional CRC6 will be appended to the device response of parcels which are not the last parcel.
- For the actual implementation the CRC32 option slightly differs as follows:
  - (3) If also the CRC6option is set, the CRCdown option will not return the correct CRCdown value.

### **Workaround for (3)**

Workaround for (3) is not to use the CRCdown feature of the client\_blockwrite telegram, but to use the dedicated get\_CRCdown telegram.

### **DAP\_TC.005 DAP client\_read: dirty bit feature of Cerberus' Triggered Transfer Mode**

*Note: This problem is only relevant for tool development, not for application development.*

The DAP telegram client\_read reads a certain number of bits from an IOclient (e.g. Cerberus). The parameter k can be selected to be zero, which is supposed to activate reading of 32 bits plus dirty bit.

However, in the current implementation, the dirty bit feature does not work correctly.

It is recommended not to use this dirty bit feature, meaning the number k should not evaluate to "0".

### **DAP\_TC.006 CRC6 error in telegram following a get\_CRCdown telegram prevents reset of CRC32 calculator**

*Note: This problem is only relevant for tool development, not for application development.*

If a CRC6 error occurs in the telegram following a get\_CRCdown telegram the AURIX™ internal CRC32 calculator does not get reset, as is the expected behavior for get\_CRCdown.

This effect can lead to unexpected CRC32 values for the next get\_CRCdown telegram. This corresponds to the perception of the tool that there has been a CRC32 error, even if the data was transmitted correctly.

#### **Workaround 1**

**Accept extra traffic for a required retransmission:** In this case the tool could see a CRC32 error which is not based on a wrong transmission, but on the missing reset of the AURIX™ internal CRC32 calculator. This would trigger the retransmission of correctly sent data.

## Workaround 2

**Check for no-reply after a get\_CRCdown telegram:** If the tool does not receive an answer for the telegram following a get\_CRCdown, it needs to re-send the get\_CRCdown telegram and ignore the data.

### **DAP\_TC.007 Incomplete client\_blockread telegram in DXCM mode when using the “read CRCup” option**

In DXCM (DAP over CAN Messages) mode, the last parcel containing the CRC32 might be skipped in a client\_blockread telegram using the “read CRCup” option.

## Workaround

Do not use CRCup option with client\_blockread telegrams in DXCM mode. Instead the CRCup can be read by a dedicated getCRCup telegram.

### **DAP\_TC.009 CRC6 error in client\_blockwrite telegram**

*Note: This problem is only relevant for tool development, not for application development.*

If a CRC6 error happens in a client\_blockwrite telegram, the DAP module will not execute the write and the tool will run into timeout according to the DAP protocol.

But in this case a following client\_blockwrite (with start address) will be ignored by the DAP module.

## Workaround

If the tool is running into a timeout after a client\_blockwrite telegram it should transmit a dummy client\_blockread telegram (e.g. len=0, arbitrary address) which will clean up the DAP client\_blockwrite function.

### **DMA\_TC.015 DMA Double Buffering: No Timestamp Support**

When a DMA channel is configured for DMA Double Buffering, and flow control (or appendage of time stamp) is selected, i.e. `DMA_ADICRz.STAMP = 1B`, the Move Engine may lock up.

#### **Workaround**

When a DMA channel is configured for DMA Double Buffering then flow control (or appendage of time stamp) should not be selected, i.e. bit `DMA_ADICRz.STAMP` must be = `0B`.

### **DMA\_TC.016 Byte and Half-word Write Accesses to specific Registers not supported**

*Note: This erratum might affect the SFR C Header Definitions. In such cases, SFR usage in the software shall be analyzed within the applications for their correct handling.*

Byte and half-word write accesses via the SPB (System Peripheral Bus) to the Regfile and Request Control logic are not supported.

This affects the following registers:

- `DMA_OTSS` (OCDS Trigger Set Select)
- `DMA_ERRINTR` (Error Interrupt)
- `DMA_PRR0` (Pattern Read Register 0)
- `DMA_PRR1` (Pattern Read Register 1)
- `DMA_MODEy` (Hardware Resource Mode)
- `DMA_HRRz` (Hardware Resource Partition)
- `DMA_SUSENRz` (Channel Suspend Enable)
- `DMA_TSRz` (Transaction State)

#### **Workaround**

Make sure only 32-bit word data is written to the registers listed above by selecting the appropriate data types.

**DMA\_TC.017 Pattern Detection Double Interrupt Trigger when INTCT = 11<sub>B</sub>**

A DMA channel z is configured for pattern detection by programming the DMA\_CHCFGRz.PATSEL to reference a data value set in one of the pattern read registers DMA\_PRR0 or DMA\_PRR1. If DMA\_ADICRz.INTCT = 11<sub>B</sub> then DMA channel z will generate a channel interrupt trigger and set CHSRz.ICH each time TCOUNT is decremented.

If a pattern match is detected then a channel interrupt trigger will be correctly generated but a second channel interrupt trigger will be generated when TCOUNT decrements. The second interrupt trigger is a bug and should not occur.

If the DMA channel z interrupt trigger is directed via the Interrupt Router to generate a DMA hardware request to another DMA channel then the second interrupt trigger may result in a Transaction Request Lost event.

**Workaround**

Workaround is to ignore the generation of the Transaction Request Lost event:

- Either disable the generation of error interrupt service requests by setting ADICRz.ETRL = 0<sub>B</sub>
- Or if the error interrupt service request is enabled, check all error status bits. If only the TRL bit for DMA channel x (pattern detection channel) is set then clear TRL and continue normal DMA operation.

**DMA\_TC.018 FPI timeout can cause pipelined register reads to break**

Due to a problem in the FPI slave interface (SIF) to the System Peripheral Bus (SPB) in the DMA module, a register access which is pipelined behind an access which is timed-out may terminate early and return the wrong data to the bus.

The scenario for this problem to occur is as follows:

1. An FPI read transaction is performed which takes a long time in the data phase. Pipelined behind this is a register access to DMA or Cerberus.

2. The first transaction is timed out, and in the same cycle the register access is taken by the SIF.

### Workaround

Timeout indicates a severe problem, meaning that something took unexpectedly long. In the event of an FPI timeout on the SPB, an error routine should be run to determine the error, and perform a system reset.

### **DMA TC.019 CBS Accesses with Large SPB:SRI Clock Ratios Configured**

When operating in debug mode and a large SPB:SRI clock ratio is configured then Cerberus accesses to the SRI address space may be unreliable and result in the Cerberus hanging.

### Workaround

Limit the SPB:SRI clock ratio to 1:1, 2:1, 3:1 or 4:1, and do not perform Cerberus accesses to the SRI address space while switching the SPB:SRI clock ratio.

### **DMA TC.020 DMA Conditional Linked List: Circular Buffer Enabled**

When a DMA channel is configured for Conditional Linked List (i.e.  $ADICRz.SHCT = 1111_B$ ) and circular buffer operation (i.e.  $ADICRz.SCBE = 1_B$  OR  $ADICRx.DCBE = 1_B$ ) then if the source and destination addresses are not set to wrap boundaries then the behaviour will not be as intended, e.g. the wrap bits  $CHCSRz.WRPS$  and  $CHCESRz.WRPD$  may be spuriously set.

### Workaround

If a DMA channel is configured for Conditional Linked List and circular buffers are enabled then the user must set the source and destination addresses to wrap boundaries.

### **DMA TC.021 Combined Software/Hardware Controlled Mode Spurious Errors**

A DMA channel is configured for combined software/hardware controlled mode. If the Move Engine is servicing a DMA channel software request and a DMA channel hardware trigger is received then a Transaction Request Lost event is set. When the Move Engine completes the current DMA access the TSRz.CH bit is not cleared. The DMA channel will continue to request channel arbitration as the CH bit is set. If the DMA channel wins arbitration then the Move Engine will continue to service the DMA channel.

In summary, 2 DMA requests (software and hardware) have resulted in 2 X DMA transfers and 1 X Transaction Request Lost (i.e. 3 X DMA actions for 2 X DMA triggers) i.e. a spurious error is generated.

#### **Workaround**

If a DMA channel is configured for combined software/hardware mode then increased attention must be paid to de-conflict the triggering of DMA channels from the servicing of DMA requests. The workaround will remove the source of spurious errors.

### **DMA TC.022 Conditional Linked List: Bus Error**

When a DMA channel is configured for Conditional Linked List (i.e. ADICRz.SHCT = 1111<sub>B</sub>) then if a bus error is reported then:

- If there is a pattern match then the number of DMA moves subsequently executed may not be as intended.
- If there is an error during the loading of a new Transaction Control Set then the DMA channel does not clear the TSRz.CH bit and begins the next DMA transaction with an erroneous Transaction Control Set.

#### **Workaround**

If a DMA channel is configured for Conditional Linked List then the user must enable the error interrupt service request. On receiving notification of an error interrupt service request the user must read the Move Engine Error Status Registers to confirm that no bus errors were reported:

- If DMA\_ERRSRx.DER = 0<sub>B</sub> and DMA\_ERRSRx.SER = 0<sub>B</sub> then no bus errors reported.
- If a bus error is reported then check the last error channel DMA\_ERRSRx.LEC.
- If DMA\_ERRSRx.DLLER = 1<sub>B</sub> then there was an error during the loading of a new Transaction Control Set.

### **DMA\_TC.024 Suspend Request coincident with Channel Activation**

If DMA channel z is suspend enabled (SUSENRz.SUSEN = 1<sub>B</sub>) and the DMA receives a suspend request then if during the same clock cycle the DMA channel becomes active in a Move Engine, the following effects will occur:

- SUSACRz.SUSAC is set for a cycle and then cleared
- A DMA transfer is performed for DMA channel z
- SUSACRz.SUSAC is set again on completion of the DMA transfer and the DMA channel is finally suspended.

#### **Workaround**

When polling SUSACRz.SUSAC in software, additionally check whether DMA channel z is active in a Move Engine x by reading bit field MExSR.CH.

### **DMA\_TC.025 Conditional Linked List: new non-CLL mode TCS load can corrupt SDCRC RAM write**

When a Conditional Linked List (CLL) transaction is running and gets a CLL pattern match, this will stop the running transaction and cause a transaction control set (TCS) load.

In case the new TCS load is set up so that it is not in CLL mode, then the SDCRC value of the new TCS may get corrupted.

#### **Workaround**

Avoid selection of non-CLL mode in the TCS loaded after a CLL pattern match.

**DMA\_TC.026 Linked List: Failed TCS load can trigger wrap interrupt**

When a Transaction Control Set (TCS) linked list load is performed, and an error is received during the load process, this terminates the load. A DMA linked list error is indicated by the error status flag ERRSRx.DLLER.

If the DADR address left in the register matches the destination wrap boundary, this results in the issuing of a destination wrap interrupt in case the destination wrap interrupt enable is set. Hence a failed TCS load has triggered an interrupt.

*Note: This only happens for destination interrupts. Logic is already in place to exclude source interrupts.*

**Workaround**

An error interrupt for the DMA linked list error is triggered by the status flag ERRSRx.DLLER if enabled by EERx.ELER. Therefore the destination wrap buffer interrupt can be ignored in this case.

**DMA\_TC.028 Transaction Request Lost (TRL) Interrupt Service Request Behaviour**

The DMA channel TRL error interrupt service request is a DMA safety measure signalling a lost DMA request to the system. For each DMA channel TRL event, the DMA may trigger one or more error interrupt service requests.

The application software should include a DMA error handler to resolve all DMA errors including TRL.

**Workaround**

None.

**DMA\_TC.031 CHCSR.ICH can be incorrectly set after pattern match**

If a pattern match is seen during a transaction, the transaction is halted for the current active channel. The move engine zeroes its internal move counter, and holds the transfer count status MEx\_CHCSR\_TCOUNT at the last value.

However, the MEx\_CHCSR.ICH bit will still be set indicating a TCOUNT decrement.

### Workaround

As there is a pattern match, a DMA channel pattern match interrupt service request will be generated. The pattern match interrupt routine can service the interrupt and clear the status bits including ICH.

### DMA\_TC.034 DMA Timestamp and Destination Circular Buffer

The DMA must not write a DMA timestamp at an address that overwrites DMA move data stored at a DMA destination address. If the DMA channel is configured for linear DMA destination address generation (DMA channel ADICRz.DCBE = 0<sub>B</sub>), the DMA appends the DMA timestamp to the end of a DMA transaction (i.e. beyond the last DMA write move data).

If the DMA channel is configured for destination circular buffer (DMA channel ADICRz.DCBE = 1<sub>B</sub>), there are three use cases:

- **Use Case 1:** the size of the DMA transaction **equals** the size of the destination circular buffer. If the DMA writes the last DMA write move data at the last address in the destination circular buffer, the DMA correctly writes the DMA timestamp beyond the destination circular buffer.
- **Use Case 2:** the size of the DMA transaction is **less than** the size of the destination circular buffer. If the DMA writes the last DMA write move data NOT at the last address in the destination circular buffer, the DMA writes the DMA timestamp inside the destination circular buffer. Erroneously, the DMA may store the DMA timestamp at an address that overwrites DMA write move data.
- **Use Case 3:** the size of the DMA transaction is **greater than** the size of the destination circular buffer. After the DMA destination address has wrapped, the DMA will overwrite DMA write move data with fresh DMA write move data.

*Note: DMA Timestamp works as specified when using only source circular buffer.*

### Workaround 1

If a DMA channel is configured

- for destination circular buffering ( $ADICRz.DCBE = 1_B$ ) AND
- the appendage of a DMA timestamp ( $ADICRz.STAMP = 1_B$ ), AND
- the size of the DMA transaction (defined by  $CFCFGRz.TREL$ ) **equals** the size of the destination circular buffer (defined by  $ADICRz.CBLD$ ),

the DMA shall append the DMA timestamp beyond the destination circular buffer if

- For increment of DMA destination address ( $ADICRz.INCD = 1_B$ ), the initial DMA destination address is at the bottom of the destination circular buffer.
- For decrement of DMA destination address ( $ADICRz.INCD = 0_B$ ), the initial DMA destination address is at the top of the destination circular buffer.

### Workaround 2

If DMA channel z is configured

- for destination circular buffering ( $ADICRz.DCBE = 1_B$ ) AND
- increment of DMA destination address ( $ADICRz.INCD = 1_B$ ) AND
- the appendage of a DMA timestamp ( $ADICRz.STAMP = 1_B$ ) AND
- the size of the DMA transaction (defined by  $CFCFGRz.TREL$ ) is **less than** the size of the destination circular buffer (defined by  $ADICRz.CBLD$ ),

the DMA shall append the DMA timestamp to the DMA write move data if the following DMA channel parameters are configured:

- $ADICRz.DMF = 001_B$  (address offset is  $2 \times CHCFGRz.CHDW$ ) AND
- $CHCFGRz.CHDW = 010_B$  (32-bit data width for moves,  $SDTW$ ).

In all other DMA destination circular buffer use cases, the DMA channel shall be configured to disable the appendage of DMA timestamp ( $ADICRz.STAMP = 0_B$ ).

### **DMA\_TC.035 Last DMA Transaction in a Linked List triggers a DMA Daisy Chain**

DMA Channels can be daisy chained by setting the bit  $CHCFGRz.PRSEL = 1_B$ . When a higher priority DMA channel z completes a DMA transaction then it will

initiate a DMA transaction on the next lower priority DMA channel z-1 by setting the access pending bit  $TSRz-1.CH$ .

However, if the current transaction was the last one in a linked list, and  $PRSEL$  is set to daisy chain,  $TSRz-1.CH$  of the next lower channel z-1 is set just after the TCS (transaction control set) load, that is, before the last transaction of the linked list has even started. Therefore the last TCS is not executed by the Linked List.

### Workaround

Do not use Daisy Chain with Linked Lists (i.e. if  $ADICRz.SHCT[3:2] = 11_B$  then  $CHCFGRz.PRSEL = 0_B$ ).

If the use case needs to trigger a further TCS in the next lower DMA channel then the trigger should be routed via the Interrupt Router.

### **DMA TC.036 Linked List: SADR/DADR can be overwritten when loading a non-LL TCS**

If a Linked List (LL) loads in a non-LL Transaction Control Set (TCS) which has a shadow mode selected ( $ADICRz.SHCT = 0001_B$  or  $0010_B$  or  $0100_B$  or  $0101_B$ ), during the write-back it can overwrite the contents of SADR/DADR in the newly loaded TCS before the DMA transaction has been run.

### Workaround

Do not use shadow address modes with DMA Conditional Linked List.

*Note: The Application Note AP32245 "DMA Linked List" will highlight that shadow address modes are not required.*

### **DMA TC.037 Conditional Linked List: Bit $TSR.CH$ not cleared for a CLL transaction upon pattern match**

When a Conditional Linked List (CLL) pattern match is found, the transaction ends.  $TSR.CH$  should be cleared, and set later during write-back of the Transaction Control Set (TCS) if the newly loaded TCS is auto-starting (i.e.  $CHCSRz.SCH = 1_B$ ).

Due to an internal problem TSR.CH is not cleared in this case.

### Workaround

There is no workaround.

The assessment is that a DMA CLL transaction that does not get a match will transition to the next DMA transaction. The CH bit will be cleared.

### **DMA TC.038 Linked List: SIT interrupt when SIT bit set in newly loaded TCS**

The Set Interrupt Trigger (SIT) bit is a means of generating a DMA channel interrupt service request via software. It is a debug feature that allows to trigger the Interrupt Router, without configuring the DMA channel and executing a DMA transaction.

When a new Transaction Control Set (TCS) is loaded in linked list mode, and the SIT bit in the new TCS being loaded is set in the value written to register CHCSRz, a channel interrupt trigger will be activated.

Therefore, the SIT bit should always be set to 0<sub>B</sub> when using linked lists.

*Note: The latest versions of the documentation are/will be updated to reflect this.*

### **DMA TC.039 Read Data CRC**

The Read Data CRC (RDCRC) calculates an IEEE 802.3 ethernet CRC32 checksum as DMA moves read data through the DMA. The DMA implementation of the algorithm does not zero extend the read data for SDTB (8-bit) and SDTH (16-bit) accesses resulting in the calculation of a wrong checksum value

The RDCRC must only be used with STDW (32-bit), SDTD (64-bit), BTR2 (128-bit) and BTR4 (256-bit) access sizes. It must be noted that SDTD, BTR2 and BTR4 are only supported for SRI-source to SRI-destination transactions.

### **DMA\_TC.040 DMA Linked Lists: Intermittent Clearing of Hardware Transaction Request Enable with mixed mode Transaction Control Sets**

When a DMA channel is configured for linked list operation, if a Transaction Control Set (TCS) is configured for Continuous Mode (DMA\_CHCFGRz.CHMODE = 1<sub>B</sub>) and the next TCS is configured for Single Mode (DMA\_CHCFGRz.CHMODE = 0<sub>B</sub>) then DMA\_TSRz.HTRE may be intermittently cleared disabling the servicing of DMA hardware requests.

#### **Workaround**

If a DMA channel is configured for linked list operation then all application DMA transactions must be configured for Continuous Mode (DMA\_CHCFGRz.CHMODE = 1<sub>B</sub>). If there is a need for the application to clear the Hardware Transaction Request Enable (DMA\_TSRz.HTRE = 0<sub>B</sub>) then two additional dummy DMA transactions should be serviced by the DMA in the linked list:

- Dummy Transaction 1:  
the TCS is configured as a linked list TCS (DMA\_ADICRz.SHCT = 0xC, 0xD or 0xE) in Single Mode (DMA\_CHCFGRz.CHMODE = 0) and auto start (DMA\_CHCSRz.SCH = 1<sub>B</sub>). The TCS should configure a single DMA move to read a word from memory in order to write DMA\_TSRz.DCH = 1<sub>B</sub> and disable subsequent DMA hardware requests.
- Dummy Transaction 2:  
the TCS is configured for normal shadow control mode (DMA\_ADICRz.SHCT = 0000<sub>B</sub>) and Single Mode. A dummy DMA move is performed.

### **DMA\_TC.041 DMA Circular Buffer Wrap Interrupt**

If a DMA channel is configured for source circular buffer operation (ADICRz.SCBE = 1<sub>B</sub>), the DMA shall correctly calculate the DMA source addresses. When the DMA source address wraps, the DMA is unreliable in updating the wrap source buffer status (CHCSRz.WRPS). If the wrap source buffer interrupt is enabled (ADICRz.WRPSE = 1<sub>B</sub>), the DMA is unreliable in triggering a source wrap buffer interrupt.

If a DMA channel is configured for destination circular buffer operation ( $ADICRz.DCBE = 1_B$ ), the DMA shall correctly calculate the DMA destination addresses. When the DMA destination address wraps, the DMA is unreliable in updating the wrap destination buffer status ( $CHCSRz.WRPD$ ). If the wrap destination buffer interrupt is enabled ( $ADICRz.WRPDE = 1_B$ ), the DMA is unreliable in triggering a destination wrap buffer interrupt.

### Workaround

The source wrap buffer interrupt shall be disabled ( $ADICRz.WRPSE = 0_B$ ).

The destination wrap buffer interrupt shall be disabled ( $ADICRz.WRPDE = 0_B$ ).

If a DMA channel is configured for circular buffer operation ( $ADICRz.SCBE = 1_B$  or  $ADICRz.DCBE = 1_B$ ), the DMA channel shall be configured as follows:

- The size of the DMA transaction shall equal the size of the circular buffer.
- If a source circular buffer is configured ( $ADICRz.SCBE = 1_B$ ), the initial DMA source address shall be the start address of the source circular buffer.
- If a destination circular buffer is configured ( $ADICRz.DCBE = 1_B$ ), the initial DMA destination address shall be the start address of the destination circular buffer.
- The DMA channel interrupt control shall be configured to trigger an interrupt on completion of the DMA transaction ( $DMA\_ADICRz.INTCT = 10_B$  and  $DMA\_ADICRz.IRDV = 0000_B$ ).

If a DMA channel is configured for both source circular buffer operation ( $ADICRz.SCBE = 1_B$ ) AND destination circular buffer operation ( $ADICRz.DCBE = 1_B$ ), the size of the source circular buffer shall equal the size of the destination circular buffer.

### **DMA\_TC.042 DMA Interrupt from Channel reported before Completion of DMA Transaction**

The Interrupt from Channel (ICH) status bit should be set on completion of a DMA transaction. If the DMA channel is configured to append a DMA Timestamp then validation have discovered that the ICH bit is set before the DMA timestamp has been written.

### Workaround 1

On receipt of a DMA channel interrupt service request software shall poll the Move Engine (ME) Status Register(s) to confirm the DMA channel is no longer active.

1. Check active DMA channel in ME SR.
2. Check Write Status in ME SR.

If these fields in both ME are no longer the DMA channel that triggered the DMA channel interrupt service request then the DMA transaction has completed.

### Workaround 2

To avoid polling the Move Engine status, the user may use a DMA linked list to execute the following DMA transactions:

- DMA transaction 1:
  - move operation (DMA timestamp shall not be selected)
- DMA transaction 2:
  - single 32-bit DMA move to copy DMA timestamp from DMA TIME register to next 32-bit aligned destination after DMA transaction 1.

### **DMA\_TC.043 DMA Write Move Data Corruption for non 32-byte Aligned Cacheable Source Address**

If the DMA channel TCS selects a 256-bit channel data width and a non 32-byte aligned source address then the beat order of the DMA write move will be different for DMA read moves to cacheable (segments 8 and 9) and non-cacheable (segments A and B) source addresses. The effect is data corruption for accesses to cacheable addresses.

### Workarounds

1. Use 32-byte aligned source addresses for DMA read move to cacheable addresses (segments 8 and 9).
2. Use non-cacheable source addresses (segments A and B).

**DMA\_TC.044 Clock Switch after SPB Error Reported results in Spurious SRI Error**

If an SPB error is reported, and then immediately the SRI:SPB clock ratio is changed, then if the next DMA read move is to an SRI source address a spurious error may be reported.

**Workaround**

1. The system shall not change the SRI:SPB clock ratio while the DMA is active.
2. The DMA error handler should monitor the reporting of SPB and SRI errors after a clock switch.

**DMA\_TC.045 DMA Reconfigures DMA Channels Lockup**

If two or more DMA channels are used to re-configure other DMA channels (i.e. perform a DMA write move to DMA address space) the DMA may lock up if the re-configuration DMA channels are assigned to different DMA hardware resource partitions.

The effect of the DMA lock up is to lock up other SPB master interfaces which attempt a write access to DMA address space.

**Workaround**

All DMA channels used to re-configure other DMA channels shall be assigned to the same hardware resource partition in their corresponding DMA Channel Hardware Resource Registers HRRz.

**DMA\_TC.046 Shadow Operation Read Only Mode**

If a DMA channel is configured for Source Address Buffering Read Only (ADICR.SHCT = 0001<sub>B</sub>) or Destination Address Buffering Read Only (ADICR.SHCT = 0010<sub>B</sub>), the DMA is unreliable when performing a shadow address update. In these modes, the SADR/DADR registers may get directly

updated (instead of SHADR) in the middle of a transaction, potentially resulting in a DMA data transfer corruption.

### Workaround

The DMA channel configuration for Read Only Modes (SHCT = 0001<sub>B</sub> or SHCT = 0010<sub>B</sub>) must not be used.

Instead, to update the SADR/DADR in the middle of a transaction, use the corresponding Direct Write Mode for Source Address Buffering (ADICR.SHCT = 0101<sub>B</sub>) or Destination Address Buffering (ADICR.SHCT = 0110<sub>B</sub>), and write the new address to the SHADR register.

### **DMA\_TC.049 Bus Error Reported During LL TCS Load**

If a DMA channel is configured for Linked List (LL) operation AND a bus error is reported during the load of a new Transaction Control Set (TCS), the DMA shall set the DMA\_ERRSRx.DLLER status bit (Move Engine x DMA Linked List Error).

Erroneously, the DMA additionally sets the DMA\_ERRSRx.SER status bit (Move Engine x Source Error).

### Workaround

None.

### **DMA\_TC.050 Clearing CHCSR.FROZEN during Double Buffering**

If a DMA channel is configured for one of the following Double Buffering operations:

- 1001<sub>B</sub> Double Source Buffering Automatic Hardware and Software Switch
- 1011<sub>B</sub> Double Destination Buffering Automatic Hardware and Software Switch

AND the active buffer fills/empties before software has cleared the DMA channel CHCSRz.FROZEN bit, the DMA shall overflow/underflow the active buffer.

Erroneously, the DMA will not trigger a Transaction Request Lost (TRL) error.

### Workaround

Software shall clear DMA channel CHCSRz.FROZEN before the active buffer overflows/underflows.

### **DMA\_TC.052 SER and DER During Linked List Operations**

Software may configure a DMA channel for one of the DMA linked list operations:

- DMA linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1100<sub>B</sub>),
- Accumulated linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1101<sub>B</sub>),
- Safe linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1110<sub>B</sub>),
- Conditional linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1111<sub>B</sub>).

If the DMA is servicing a DMA request for a DMA channel configured for one of the linked list operations and the DMA indicates a Source Error (SER) (i.e. DMA\_ERRSRx.SER = 1<sub>B</sub>) or a Destination Error (DER) (i.e. DMA\_ERRSRx.DER = 1<sub>B</sub>), the DMA completes the current DMA transaction. If the DMA channel is configured for conditional linked list, the DMA disables pattern matching for each DMA read move reporting a SER. When the DMA completes the current DMA transaction, the DMA stops servicing the linked list operation and the DMA will not load the next transaction control set to allow debug of the current DMA transaction.

Erroneously, upon a SER or DER, the DMA does not reliably stop the linked list operation (when it should) on completion of the current DMA transaction.

If the Move Engine is configured to enable DMA error interrupt service request for SER (DMA\_EERx.ESER = 1<sub>B</sub>) and for DER (DMA\_EERx.EDER = 1<sub>B</sub>), the DMA triggers a DMA error interrupt service request.

The application software should include a DMA error handler to resolve all DMA errors including SER and DER.

**Workaround**

None.

**DMA\_TC.053 TS16\_ERR Type of Error Reporting Unreliable**

During debugging, the error trigger set (TS16\_ERR) may be used to identify the type of DMA error and the number of the DMA channel. After TS16\_ERR reports an error the error type bits (ME0SE, ME0DE, ME1SE and ME1DE) are not cleared. If TS16\_ERR reports a subsequent error, the type of error reporting is unreliable.

**Workaround**

After TS16\_ERR reports an error, the error type bits must be cleared.

**DMA\_TC.054 DMA Channel Halt Acknowledge Unreliable**

Software may halt a DMA channel by writing to the halt request bit (TSRz.HLTREQ = 1<sub>B</sub>). When a DMA channel enters the halt state, the DMA reports DMA channel halt acknowledge (TSRz.HLTACK = 1<sub>B</sub>).

The reporting of DMA channel halt acknowledge is unreliable when software sets the TSRz.HLTREQ bit just as channel z is about to be scheduled to a move engine. In this case, the DMA may report a DMA channel is halted when the DMA channel is active in a move engine.

**Workaround**

If the DMA reports a DMA channel is halted, the software should check the DMA channel is not active in a move engine by monitoring the active channel in the move engine status register(s).

### **DMA\_TC.055 ICU to DMA Interface in Sleep Mode**

The Interrupt Router triggers DMA hardware requests via the ICU interface. If the DMA is in sleep mode, the DMA will not acknowledge DMA hardware requests. The effect is to lock up the ICU to DMA interface.

#### **Workaround**

The application must disable the triggering of DMA hardware requests before placing the DMA in sleep mode.

### **DMA\_TC.056 TSR and SUSENR Access Protection Unreliable**

The DMA access protection is part of a system wide access protection scheme to restrict write accesses to DMA registers to individual on-chip bus masters.

If the application software configures DMA freedom from interference measures (i.e. when any on-chip bus master write to the DMA is prohibited by a DMA access enable setting), then on-chip bus master writes to the DMA channel TSR and SUSENR registers are unreliable and may result in the following effects:

#### **1. Safety Related Effects**

- 1.1. An illegal write access to a DMA channel TSR register will succeed with no indication.

The safety related effects (in point 1.1) relate to the DMA channel reset, halt and hardware request control functions in the TSR register. The most severe safety effect is that a DMA operation may be lost.

#### **Workaround (for 1.1):**

If the application software implements temporal monitoring of DMA transactions (e.g. using DMA timestamp) to detect lost DMA operations, the application software will detect the effect of the illegal access to DMA channel TSR register.

## 2. Non Safety Related Effects

- 2.1. An illegal write access to a DMA channel SUSENR register may succeed with no indication.
  - Impact of 2.1: The SUSENR register is a debug only register. No impact is foreseen during a normal application.
- 2.2. A legal write access to a DMA channel TSR register may fail with an indication - this means unexpected bus errors may be triggered when accessing TSR registers.
- 2.3. A legal write access to a DMA channel SUSENR register may fail with an indication - this means unexpected bus errors may be triggered when accessing SUSENR registers.
  - Impact of 2.2 & 2.3: Unexpected SPB bus errors and hence CPU traps and SPB error alarms may occur during application run.

### Workaround (for 2.2 & 2.3):

If the system implements DMA freedom from interference measures, then the Impact of 2.2 & 2.3 will occur, and cause unexpected SPB bus errors and hence CPU traps and SPB error alarms when writing to TSR and SUSENR registers.

In order to work around this problem, the application software shall implement all of the following steps:

- W1: Before an intended write access to a DMA channel TSR or SUSENR register, perform an additional preceding write access to a DMA channel Transaction Control Set (TCS) register of the same DMA channel.
  - TCS registers include the DMA channel RDCRC, SDCRC, SADR, DADR, SHADR, ADICR, CHCSR and CHCFGR registers.
- W2: Ensure that this additional preceding write access to a DMA channel TCS register has no real effect. Recommendation: Simply read and write back the RDCRCR register.
- W3: Perform the write access to the DMA channel TSR register.

Ensure that no other on-chip bus master can access any DMA register of a different resource partition between steps W2 and W3 in the workaround above.

### Example Code Snippet:

To update TSR register of DMA channel 25 with value:

1. Uint32 temp = DMA\_RDCRCR25.U;
2. DMA\_RDCRCR25.U = temp;
3. DMA\_TSR25.U = value;

### **DMA\_TC.058 Linked List Load Transaction Control Set (TCS) Integrity Error**

If DMA channel z is configured for one of the following linked list operations:

- DMA Linked List
  - (DMA channel ADICRz.SHCT = 1100<sub>B</sub>)
- Accumulated Linked List
  - (DMA channel ADICRz.SHCT = 1101<sub>B</sub>)
- Safe Linked List
  - (DMA channel ADICRz.SHCT = 1110<sub>B</sub>)
- Conditional Linked List
  - (DMA channel ADICRz.SHCT = 1111<sub>B</sub>)

Then on completion of a DMA transaction a new TCS is loaded into DMA channel z from the on-chip bus.

The DMA ignores data integrity errors in the new TCS:

- The DMA does not trigger an alarm to the SMU.
- The DMA does not store any DMA error status.
- The DMA may execute a corrupted DMA transaction.

Detection of most corrupted DMA transactions is provided by the DMA safety mechanisms as follows:

- Use of the DMA address checksum to detect address generation faults.
- Use of the DMA timestamp<sup>1)</sup> to detect temporal faults.

### **Workaround**

None.

---

1) Conditional Linked List does not support the appendage of timestamps (ADICRz.STAMP = 0<sub>B</sub>).

## **DMA\_TC.061 DMA Double Buffering Operations**

*Note: This erratum DMA\_TC.061 (DMA Double Buffering Operations) substitutes the following errata text modules*

- *DMA\_TC.029 (DMA Double Buffering Overflow),*
  - *DMA\_TC.047 (DMA Double Buffering Buffer Switch), and*
  - *DMA\_TC.057 (Double Buffering Overflow Causes Other Channel Corruption)*
- included in previous TC2xx errata sheet releases.*

Software may configure a DMA channel for one of the DMA double buffering operations:

- DMA Double Source Buffering Software Switch Only
  - (DMA channel DMA\_ADICRz.SHCT = 1000<sub>B</sub>),
- DMA Double Source Buffering Automatic Hardware and Software Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1001<sub>B</sub>),
- DMA Double Destination Buffering Software Switch Only
  - (DMA channel DMA\_ADICRz.SHCT = 1010<sub>B</sub>),
- DMA Double Destination Buffering Automatic Hardware and Software Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1011<sub>B</sub>).

If the DMA is servicing a DMA request for a DMA channel configured for one of the double buffering operations AND the software executes a Software Buffer Switch operation (DMA\_CHCSRz.SWB = 1<sub>B</sub>), the DMA will not perform the buffer switch reliably.

The following sections provide recommendations for the implementation of DMA double buffering operations.

### **Supported DMA double buffering operations:**

As a consequence, the software should configure for a limited number of DMA double buffering operations:

- DMA Double Source Buffering Automatic Hardware Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1001<sub>B</sub>),
- DMA Double Destination Buffering Automatic Hardware Switch

- (DMA channel DMA\_ADICRz.SHCT = 1011<sub>B</sub>).

The software must

- NOT perform a Software Buffer Switch (DMA\_CHCSRz.SWB = 0<sub>B</sub>),
- NOT set the frozen bit (DMA\_CHCSRz.FROZEN = 1<sub>B</sub>).

#### **DMA channel ETRL configuration:**

The software must set the Enable Transaction Request Lost (ETRL) bit (DMA\_ADICRz.ETRL = 1<sub>B</sub>) to prevent the DMA locking up during a DMA double buffering operation.

#### **DMA channel monitoring:**

The software should configure the DMA to trigger a DMA channel interrupt service request when the DMA empties (source buffering) or fills (destination buffering) a buffer on the completion of a DMA transaction. The software must service the DMA channel interrupt service requests. As soon as the software has analysed a buffer, the software must clear the frozen bit (DMA\_CHCSRz.FROZEN = 0<sub>B</sub>) and re-initialise the buffer address pointer.

#### **DMA channel underflow or overflow:**

If the software fails to analyse a frozen buffer before the next DMA channel interrupt service request, the DMA channel will underflow (source buffering) or overflow (destination buffering) on receiving the next DMA request. Erroneously, the DMA will not trigger a DMA error interrupt service request.

As soon as the CPU receives a DMA channel interrupt service request, the software must check for an underflow or overflow by monitoring the DMA transaction count. If the software reads a zero transaction count (DMA\_CHCSRz.TCOUNT = 0<sub>D</sub>), the DMA channel is in an underflow or overflow state.

#### **DMA channel interference:**

Erroneously a DMA channel underflow or overflow may cause the setting of the TRL flag and the clearing of a DMA request in one or more other DMA channels (note: dependent on the scheduling of DMA channels around this DMA

request). The DMA channel interference is independent of resource partition assignment.

#### **DMA channel reset:**

If the software detects a DMA channel underflow or overflow, the software must apply a DMA channel reset to all used DMA channels. On completion of the DMA channel reset, the software must re-configure all used DMA channels.

Alternatively, the software may apply an application reset.

#### **Workaround**

None.

#### **DMA TC.062 Termination of DMA Transaction for Pattern Match**

If a DMA channel is configured for pattern detection and the DMA detects a pattern match, the DMA should terminate the DMA transaction. The DMA should provide the software with the capability to use the DMA channel status to identify the transfer number of the DMA move data.

Erroneously, the DMA may decrement 1 from the TCOUNT value making identification of the DMA move data unreliable.

#### **Workaround**

None.

#### **DMA TC.063 DMA Timestamp Destination Address**

If software configures a DMA channel

- for increment of DMA destination address ( $\text{DMA\_ADICRz.INCD} = 1_{\text{B}}$ ) AND
- to append a DMA timestamp ( $\text{DMA\_ADICRz.STAMP} = 1_{\text{B}}$ );

and the intended write address of the DMA timestamp is in a different 32 Kbyte page to the last DMA destination address to write DMA move data, the DMA erroneously calculates the DMA timestamp write address. The DMA writes the

DMA timestamp to an incorrect address inside the same 32 Kbyte page as the last DMA destination address.

### Workaround

The last DMA destination address and the write address of the DMA timestamp shall exist in the same 32 Kbyte page (i.e. and shall not cross the 32 Kbyte page boundary).

### **DMA TC.064 DMA Daisy Chain Request**

If software configures a DMA channel for one of the following DMA operations:

- DMA Pattern Detection
  - (DMA channel DMA\_CHCFGRz.PATSEL[1:0] != 00<sub>B</sub>),
- DMA Double Source Buffering Software Switch Only
  - (DMA channel DMA\_ADICRz.SHCT = 1000<sub>B</sub>),
- DMA Double Source Buffering Automatic Hardware and Software Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1001<sub>B</sub>),
- DMA Double Destination Buffering Software Switch Only
  - (DMA channel DMA\_ADICRz.SHCT = 1010<sub>B</sub>),
- DMA Double Destination Buffering Automatic Hardware and Software Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1011<sub>B</sub>),
- DMA linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1100<sub>B</sub>),
- Accumulated linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1101<sub>B</sub>),
- Safe linked list
  - (DMA channel DMA\_ADICRz.SHCT = 1110<sub>B</sub>),
- Conditional linked list
  - (DMA channel ADICRz.SHCT = 1111<sub>B</sub>),

the software must not select daisy chain (DMA channel CHCFGRz.PRSEL = 0<sub>B</sub>).

**DMA\_TC.065 DMA Move Concurrent Bus Accesses**

The highest number DMA channel always wins arbitration to shared DMA resources (Move Engine and DMA on-chip bus master interfaces). The configuration of the DMA priority (DMA\_CHCFGRx.DMAPRIO) has no effect on internal DMA arbitration.

The DMA priority is used by the System Peripheral Bus (SPB) controller to arbitrate between requests from all the SPB master interfaces.

**Workaround**

None.

**DMA\_TC.066 DMA Double Buffering Operations - Update Address Pointer**

Software may configure a DMA channel for one of the DMA double buffering operations:

- DMA Double Source Buffering Software Switch Only
  - (DMA channel DMA\_ADICRz.SHCT = 1000<sub>B</sub>),
- DMA Double Source Buffering Automatic Hardware and Software Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1001<sub>B</sub>),
- DMA Double Destination Buffering Software Switch Only
  - (DMA channel DMA\_ADICRz.SHCT = 1010<sub>B</sub>),
- DMA Double Destination Buffering Automatic Hardware and Software Switch
  - (DMA channel DMA\_ADICRz.SHCT = 1011<sub>B</sub>).

If the software updates a buffer address pointer by BYTE or HALF-WORD writes, the resulting value of the address pointer is corrupted.

**Workaround**

If the software updates a buffer address pointer, the software should only use a 32-bit WORD access.

### **DTS\_TC.001 Temperature Sensor Formula**

The formula documented in older Data Sheet versions may result in an increased temperature error when calculating the junction temperature  $T_j$  of the device from a DTS temperature measurement.

To properly calculate the temperature measured by the DTS in [ $^{\circ}\text{C}$ ] from the RESULT bit field of register SCU\_DTSSTAT, it is recommended to use the following formulas depending on the contents of bit field SCU\_DTSCON[30:29]:

- While bit field SCU\_DTSCON[30:29] = 00<sub>B</sub>:  $T_j = (\text{RESULT} - 607_{\text{D}}) / 2.13$
- While bit field SCU\_DTSCON[30:29] = 01<sub>B</sub>:  $T_j = (\text{RESULT} - 646_{\text{D}}) / 2.11$

Bit field SCU\_DTSCON[30:29] can only deliver one of the two values (00<sub>B</sub>, 01<sub>B</sub>) listed above (constant for a given device).

Make sure the application software does not modify the values installed during device start-up in register SCU\_DTSCON.

*Note: The description in the Data Sheet will be updated appropriately.*

### **FLASH\_TC.052 Use of Write Page Once command**

When applying a Write Page Once (WPO) command to a pre-programmed or incompletely erased PFlash location, the WPO command will fail as expected, with both EVER (Erase Verify Error) and PVER (Program Verify Error) error flags being raised.

For an EVER failure in the WPO command, the read bias conditions on the NVM cells for the subsequent read operations will be incorrect. The incorrect bias conditions at the NVM cell terminals may lead to single-bit or multi-bit errors in the PFlash. Only zeroes (erased cells) will be affected by this phenomenon.

The physical content of the flash cells is not damaged by the incorrect read bias conditions, or by the WPO command failure.

*Note: As per the safety manual's Architecture for Management of Faults [SM\_AURIX\_PMU\_3], it is assumed that the WPO command is not used during application run time.*

## Workaround

The incorrect NVM read bias conditions can be fully recovered by performing one of the following actions immediately after the WPO failure:

- Request Flash module sleep mode and wake-up immediately after the WPO failure:
  - Request Sleep mode by setting bit FCON.SLEEP = 1<sub>B</sub>,
  - Poll the Flash Sleep Mode status bit FSR.SLM to make sure that the Flash is in sleep mode,
  - Initiate wake-up by clearing FCON.SLEEP = 0<sub>B</sub>,
  - Poll status bit FSR.SLM to make sure that the flash is in normal state again.

*Note: For more details about AURIX™ power-down modes, please refer to Application Note “AURIX™ standby power mode” (AP32332).*

- Perform System Reset immediately after the WPO failure.

## **FlexRay AI.087 After reception of a valid sync frame followed by a valid non-sync frame in the same static slot the received sync frame may be ignored**

### Description:

If in a static slot of an even cycle a valid sync frame followed by a valid non-sync frame is received, and the frame valid detection (prt\_frame\_decoded\_on\_X) of the DEC process occurs one sclk after valid frame detection of FSP process (fsp\_val\_syncfr\_chx), the sync frame is not taken into account by the CSP process (devte\_xxs\_reg).

### Scope:

The erratum is limited to the case where more than one valid frame is received in a static slot of an even cycle.

### Effects:

In the described case the sync frame is not considered by the CSP process. This may lead to a SyncCalcResult of MISSIMG\_TERM (error flag SFS.MRCS

set). As a result the POC state may switch to NORMAL\_PASSIVE or HALT or the Startup procedure is aborted.

### Workaround

Avoid static slot configurations long enough to receive two valid frames.

### **FlexRay\_AI.088 A sequence of received WUS may generate redundant SIR.WUPA/B events**

Description:

If a sequence of wakeup symbols (WUS) is received, all separated by appropriate idle phases, a valid wakeup pattern (WUP) should be detected after every second WUS. The E-Ray detects a valid wakeup pattern after the second WUS and then after each following WUS.

Scope:

The erratum is limited to the case where the application program frequently resets the appropriate SIR.WUPA/B bits.

Effects:

In the described case there are more SIR.WUPA/B events seen than expected.

### Workaround

Ignore redundant SIR.WUPA/B events.

### **FlexRay\_AI.089 Rate correction set to zero in case of SyncCalcResult=MISSING\_TERM**

Description:

In case a node receives too few sync frames for rate correction calculation and signals a SyncCalcResult of MISSING\_TERM, the rate correction value is set to zero instead of the last calculated value.

**Scope:**

The erratum is limited to the case of receiving too few sync frames for rate correction calculation (SyncCalcResult=MISSING\_TERM in an odd cycle).

**Effects:**

In the described case a rate correction value of zero is applied in NORMAL\_ACTIVE / NORMAL\_PASSIVE state instead of the last rate correction value calculated in NORMAL\_ACTIVE state. This may lead to a desynchronisation of the node although it may stay in NORMAL\_ACTIVE state (depending on gMaxWithoutClockCorrectionPassive) and decreases the probability to re-enter NORMAL\_ACTIVE state if it has switched to NORMAL\_PASSIVE (pAllowHaltDueToClock=false).

**Workaround**

It is recommended to set gMaxWithoutClockCorrectionPassive to 1. If missing sync frames cause the node to enter NORMAL\_PASSIVE state, use higher level application software to leave this state and to initiate a re-integration into the cluster. HALT state can also be used instead of NORMAL\_PASSIVE state by setting pAllowHaltDueToClock to true.

**FlexRay AI.090 Flag SFS.MRCS is set erroneously although at least one valid sync frame pair is received****Description:**

If in an odd cycle  $2c+1$  after reception of a sync frame in slot  $n$  the total number of different sync frames per double cycle has exceeded gSyncNodeMax and the node receives in slot  $n+1$  a sync frame that matches with a sync frame received in the even cycle  $2c$ , the sync frame pair is not taken into account by CSP process. This may cause the flags SFS.MRCS and EIR.CCF to be set erroneously.

**Scope:**

The erratum is limited to the case of a faulty cluster configuration where different sets of sync frames are transmitted in even and odd cycles and the total number of different sync frames is greater than gSyncNodeMax.

Effects:

In the described case the error interrupt flag `EIR.CCF` is set and the node may enter either the POC state `NORMAL_PASSIVE` or `HALT`.

### Workaround

Correct configuration of gSyncNodeMax.

### **FlexRay AI.091 Incorrect rate and/or offset correction value if second Secondary Time Reference Point (STRP) coincides with the action point after detection of a valid frame**

Description:

If a valid sync frame is received before the action point and additionally noise or a second frame leads to a STRP coinciding with the action point, an incorrect deviation value of zero is used for further calculations of rate and/or offset correction values.

Scope:

The erratum is limited to configurations with an action point offset greater than static frame length.

Effects:

In the described case a deviation value of zero is used for further calculations of rate and/or offset correction values. This may lead to an incorrect rate and/or offset correction of the node.

### Workaround

Configure action point offset smaller than static frame length.

**FlexRay\_AI.092 Initial rate correction value of an integrating node is zero if pMicroInitialOffsetA,B = 0x00**

## Description:

The initial rate correction value as calculated in figure 8-8 of protocol spec v2.1 is zero if parameter pMicroInitialOffsetA,B was configured to be zero.

## Scope:

The erratum is limited to the case where pMicroInitialOffsetA,B is configured to zero.

## Effects:

Starting with an initial rate correction value of zero leads to an adjustment of the rate correction earliest 3 cycles later (see figure 7-10 of protocol spec v2.1). In a worst case scenario, if the whole cluster is drifting away too fast, the integrating node would not be able to follow and therefore abort integration.

**Workaround**

Avoid configurations with pMicroInitialOffsetA,B equal to zero. If the related configuration constraint of the protocol specification results in pMicroInitialOffsetA,B equal to zero, configure it to one instead. This will lead to a correct initial rate correction value, it will delay the startup of the node by only one microtick.

**FlexRay\_AI.093 Acceptance of startup frames received after reception of more than gSyncNodeMax sync frames**

## Description:

If a node receives in an even cycle a startup frame after it has received more than gSyncNodeMax sync frames, this startup frame is added erroneously by process CSP to the number of valid startup frames (zStartupNodes). The faulty number of startup frames is delivered to the process POC. As a consequence this node may integrate erroneously to the running cluster because it assumes that it has received the required number of startup frames.

**Scope:**

The erratum is limited to the case of more than gSyncNodeMax sync frames.

**Effects:**

In the described case a node may erroneously integrate successfully into a running cluster.

**Workaround**

Use frame schedules where all startup frames are placed in the first static slots. gSyncNodeMax should be configured to be greater than or equal to the number of sync frames in the cluster.

**FlexRay AI.094 Sync frame overflow flag `EIR.SFO` may be set if slot counter is greater than 1024****Description:**

If in the static segment the number of transmitted and received sync frames reaches gSyncNodeMax and the slot counter in the dynamic segment reaches the value  $cStaticSlotIDMax + gSyncNodeMax = 1023 + gSyncNodeMax$ , the sync frame overflow flag `EIR.SFO` is set erroneously.

**Scope:**

The erratum is limited to configurations where the number of transmitted and received sync frames equals to gSyncNodeMax and the number of static slots plus the number of dynamic slots is greater or equal than  $1023 + gSyncNodeMax$ .

**Effects:**

In the described case the sync frame overflow flag `EIR.SFO` is set erroneously. This has no effect to the POC state.

## Workaround

Configure gSyncNodeMax to number of transmitted and received sync frames plus one or avoid configurations where the total of static and dynamic slots is greater than cStaticSlotIDMax.

### **FlexRay AI.095 Register RCV displays wrong value**

Description:

If the calculated rate correction value is in the range of [-pClusterDriftDamping .. +pClusterDriftDamping], vRateCorrection of the CSP process is set to zero. In this case register RCV should be updated with this value. Erroneously RCV.RCV[11:0] holds the calculated value in the range [-pClusterDriftDamping .. +pClusterDriftDamping] instead of zero.

Scope:

The erratum is limited to the case where the calculated rate correction value is in the range of [-pClusterDriftDamping .. +pClusterDriftDamping].

Effects:

The displayed rate correction value RCV.RCV[11:0] is in the range of [-pClusterDriftDamping .. +pClusterDriftDamping] instead of zero. The error of the displayed value is limited to the range of [-pClusterDriftDamping .. +pClusterDriftDamping]. For rate correction in the next double cycle always the correct value of zero is used.

## Workaround

A value of RCV.RCV[11:0] in the range of [-pClusterDriftDamping .. +pClusterDriftDamping] has to be interpreted as zero.

### **FlexRay AI.096 Noise following a dynamic frame that delays idle detection may fail to stop slot**

Description:

If (in case of noise) the time between 'potential idle start on X' and 'CHIRP on X' (see Protocol Spec. v2.1, Figure 5-21) is greater than `gdDynamicSlotIdlePhase`, the E-Ray will not remain for the remainder of the current dynamic segment in the state 'wait for the end of dynamic slot rx'. Instead, the E-Ray continues slot counting. This may enable the node to further transmissions in the current dynamic segment.

**Scope:**

The erratum is limited to noise that is seen only locally and that is detected in the time window between the end of a dynamic frame's DTS and idle detection ('CHIRP on X').

**Effects:**

In the described case the faulty node may not stop slot counting and may continue to transmit dynamic frames. This may lead to a frame collision in the current dynamic segment.

**Workaround**

None.

**FlexRay AI.097 Loop back mode operates only at 10 MBit/s****Description:**

The looped back data is falsified at the two lower baud rates of 5 and 2.5 MBit/s.

**Scope:**

The erratum is limited to test cases where loop back is used with the baud rate prescaler (`PRTC1.BRP[1:0]`) configured to 5 or 2.5 MBit/s.

**Effects:**

The loop back self test is only possible at the highest baud rate.

## Workaround

Run loop back tests with 10 MBit/s (`PRTC1.BRP[1:0] = 00B`).

### **FlexRay\_AI.099 Erroneous cycle offset during startup after abort of startup or normal operation**

#### Description:

An abort of startup or normal operation by a READY command near the macotick border may lead to the effect that the state INITIALIZE\_SCHEDULE is one macrotick too short during the first following integration attempt. This leads to an early cycle start in state INTEGRATION\_COLDSTART\_CHECK or INTEGRATION\_CONSISTENCY\_CHECK.

As a result the integrating node calculates a cycle offset of one macrotick at the end of the first even/odd cycle pair in the states INTEGRATION\_COLDSTART\_CHECK or INTEGRATION\_CONSISTENCY\_CHECK and tries to correct this offset.

If the node is able to correct the offset of one macrotick (`pOffsetCorrectionOut >> gdMacrotick`), the node enters NORMAL\_ACTIVE with the first startup attempt.

If the node is not able to correct the offset error because `pOffsetCorrectionOut ≤ gdMacrotick`, the node enters ABORT\_STARTUP and is ready to try startup again. The next (second) startup attempt is not effected by this erratum.

#### Scope:

The erratum is limited to applications where READY command is used to leave STARTUP, NORMAL\_ACTIVE, or NORMAL\_PASSIVE state.

#### Effects:

In the described case the integrating node tries to correct an erroneous cycle offset of one macrotick during startup.

### Workaround

With a configuration of `pOffsetCorrectionOut >> gdMacrotick • (1+cClockDeviationMax)` the node will be able to correct the offset and therefore also be able to successfully integrate.

### **FlexRay AI.100 First WUS following received valid WUP may be ignored**

#### Description:

When the protocol engine is in state WAKEUP\_LISTEN and receives a valid wakeup pattern (WUP), it transfers into state READY and updates the wakeup status vector `CCSV.WSV[2:0]` as well as the status interrupt flags `SIR.WST` and `SIR.WUPA/B`. If the received wakeup pattern continues, the protocol engine may ignore the first wakeup symbol (WUS) following the state transition and signals the next `SIR.WUPA/B` at the third instead of the second WUS.

#### Scope:

The erratum is limited to the reception of redundant wakeup patterns.

#### Effects:

Delayed setting of status interrupt flags `SIR.WUPA/B` for redundant wakeup patterns.

### Workaround

None.

### **FlexRay AI.101 READY command accepted in READY state**

#### Description:

The E-Ray module does not ignore a READY command while in READY state.

#### Scope:

The erratum is limited to the READY state.

Effects:

Flag `CCSV.CSI` is set. Cold starting needs to be enabled by POC command `ALLOW_COLDSTART` (`SUCC1.CMD = 1001B`).

### Workaround

None.

### **FlexRay AI.102 Slot Status vPOC!SlotMode is reset immediately when entering HALT state**

Description:

When the protocol engine is in the states `NORMAL_ACTIVE` or `NORMAL_PASSIVE`, a `HALT` or `FREEZE` command issued by the Host resets `vPOC!SlotMode` immediately to `SINGLE` slot mode (`CCSV.SLM[1:0] = 00B`). According to the FlexRay protocol specification, the slot mode should not be reset to `SINGLE` slot mode before the following state transition from `HALT` to `DEFAULT_CONFIG` state.

Scope:

The erratum is limited to the `HALT` state.

Effects:

The slot status `vPOC!SlotMode` is reset to `SINGLE` when entering `HALT` state.

### Workaround

None.

### **FlexRay AI.103 Received messages not stored in Message RAM when in Loop Back Mode**

After a FREEZE or HALT command has been asserted in NORMAL\_ACTIVE state, and if state LOOP\_BACK is then entered by transition from HALT state via DEF\_CONFIG and CONFIG, it may happen that acceptance filtering for received messages is not started, and therefore these messages are not stored in the respective receive buffer in the Message RAM.

Scope:

The erratum is limited to the case where Loop Back Mode is entered after NORMAL\_ACTIVE state was left by FREEZE or HALT command.

Effects:

Received messages are not stored in Message RAM because acceptance filtering is not started.

### **Workaround**

Leave HALT state by hardware reset.

### **FlexRay AI.104 Missing startup frame in cycle 0 at coldstart after FREEZE or READY command**

When the E-Ray is restarted as leading coldstarter after it has been stopped by FREEZE or READY command, it may happen, depending on the internal state of the module, that the E-Ray does not transmit its startup frame in cycle 0. Only E-Ray configurations with startup frames configured for slots 1 to 7 are affected by this behaviour.

Scope:

The erratum is limited to the case when a coldstart is initialized after the E-Ray has been stopped by FREEZE or READY command. Coldstart after hardware reset is not affected.

Effects:

During coldstart it may happen that no startup frame is sent in cycle 0 after entering COLDSTART\_COLLISION\_RESOLUTION state from COLDSTART\_LISTEN state.

Severity:

Low, as the next coldstart attempt is no longer affected. Coldstart sequence is lengthened but coldstart of FlexRay system is not prohibited by this behaviour.

### Workaround

Use a static slot greater or equal 8 for the startup / sync message.

### FlexRay AL105 RAM select signals of IBF1/IBF2 and OBF1/OBF2 in RAM test mode

When accessing Input Buffer RAM 1,2 (IBF1,2) or Output Buffer RAM 1,2 (OBF1,2) in RAM test mode, the following behaviour can be observed when entering RAM test mode after hardware reset.

- Read or write access to IBF2:
  - In this case also IBF1 RAM select **eray\_ibf1\_cen** is activated initiating a read access of the addressed IBF1 RAM word. The data read from IBF1 is evaluated by the respective parity checker.
- Read or write access to OBF1:
  - In this case also OBF2 RAM select **eray\_obf2\_cen** is activated initiating a read access of the addressed OBF2 RAM word. The data read from OBF2 is evaluated by the respective parity checker.

If the parity logic of the erroneously selected IBF1 resp. OBF2 detects a parity error, bit **MHDS.PIBF** resp. **MHDS.POBF** in the E-Ray Message Handler Status register is set although the addressed IBF2 resp. OBF1 had not error. The logic for setting **MHDS.PIBF** / **MHDS.POBF** does not distinguish between set conditions from IBF1 or IBF2 resp. OBF1 or OBF2.

Due to the IBF / OBF swap mechanism as described in section 5.11.2 in the E-Ray Specification, the inverted behaviour with respect to IBF1,2 and OBF1,2 can be observed depending on the IBF / OBF access history.

### Scope:

The erratum is limited to the case when IBF1,2 or OBF1,2 are accessed in RAM test mode. The problem does not occur when the E-Ray is in normal operation mode.

**Effects:**

When reading or writing IBF1,2 / OBF1,2 in RAM test mode, it may happen, that the parity logic of IBF1,2 / OBF1,2 signals a parity error.

**Severity:**

Low, workaround available.

**Workaround**

For RAM testing after hardware reset, the Input / Output Buffer RAMs have to be first written and then read in the following order: IBF1 before IBF2 and OBF2 before OBF1

**FlexRay AI.106 Data transfer overrun for message transfers Message RAM to Output Buffer (OBF) or from Input Buffer (IBF) to Message RAM**

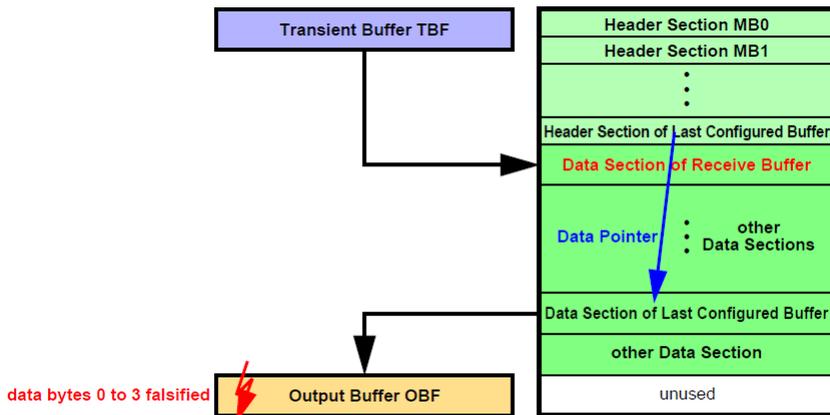
The problem occurs under the following conditions:

- 1) A received message is transferred from the Transient Buffer RAM (TBF) to the message buffer that has its data pointer pointing to the first word of the Message RAM's Data Partition located directly after the last header word of the Header Partition of the Last Configured Buffer as defined by **MRC.LCB**.
- 2) The Host triggers a transfer from / to the Last Configured Buffer in the Message RAM with a specific time relation to the start of the TBF transfer described under 1).

Under these conditions the following transfers triggered by the Host may be affected:

- a) Message buffer transfer from Message RAM to OBF

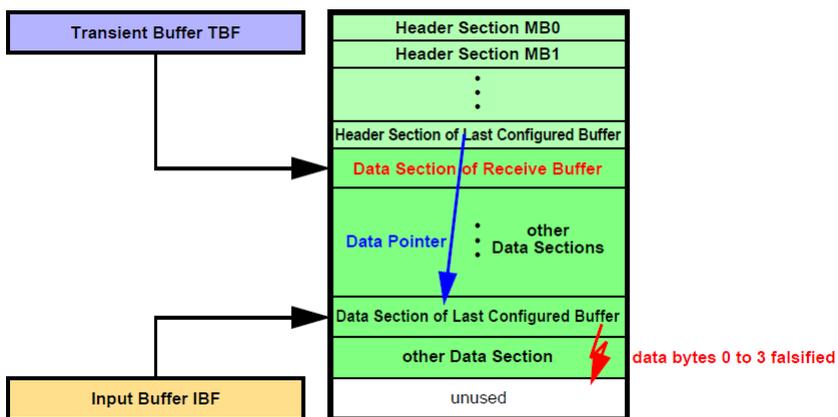
When the message buffer has its payload configured to maximum length (**PLC** = 127), the OBF word on address 00h (payload data bytes 0 to 3) is overwritten with unexpected data at the end of the transfer.



**Figure 1 Message buffer transfer from Message RAM to OBF**

b) Message buffer transfer from IBF to Message RAM

After the Data Section of the selected message buffer in the Message RAM has been written, one additional write access overwrites the following word in the Message RAM which might be the first word of the next Data Section.



**Figure 2 Message buffer transfer from IBF to Message RAM**

**Scope:**

The erratum is limited to the case when (see [Figure 3](#) “Bad Case”):

1) The first Data Section in the Data Partition is assigned to a receive buffer (incl. FIFO buffers)

**AND**

2) The Data Partition in the Message RAM starts directly after the Header Partition (no unused Message RAM word in between)

**Effects:**

a) When a message is transferred from the Last Configured Buffer in the Message RAM to the OBF and **PLC** = 127 it may happen, that at the end of the transfer the OBF word on address 00h (payload data bytes 0 to 3) is overwritten with unexpected data (see [Figure 1](#)).

b) When a message is transferred from IBF to the Last Configured Buffer in the Message RAM, it may happen, that at the end of the transfer of the Data Section one additional write access overwrites the following word, which may be the first word of another message's Data Section in the Message RAM (see [Figure 2](#)).

**Severity:**

Medium, workaround available, check of configuration necessary.

**Workaround**

1) Leave at least one unused word in the Message RAM between Header Section and Data Section.

**OR**

2) Ensure that the Data Section directly following the Header Partition is assigned to a transmit buffer.

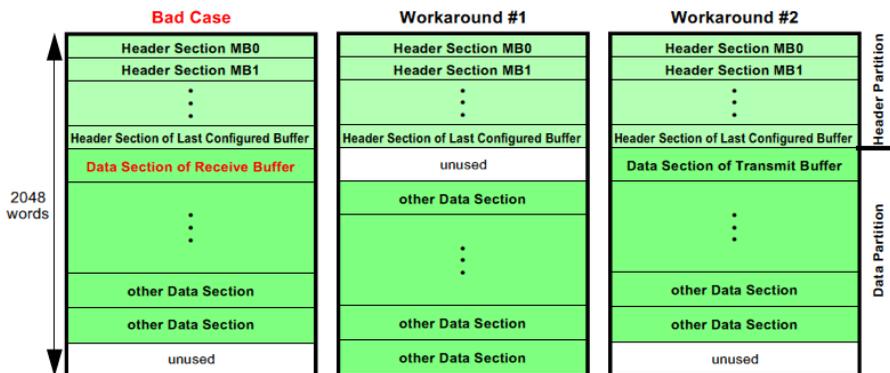


Figure 3 Message RAM Configurations

**GTM\_AI.132** GTM\_TOP level: AEI write to BRIDGE\_MODE register can result in blocking of AEI configuration interface

If the GTM bus bridge operates in MSK\_WR\_RESP=1 mode, a requested change of the GTM\_IP bridge mode (Bit BRG\_MODE) can result in blocking of the bus interface.

**Scope**

All AEI protocols.

**Effects**

GTM Bus interface does not issue aei\_ready/aei\_response\_ready which could lead to bus timeout of the serving bus master.

**Workaround**

Ensure that the write command to the BRIDGE\_MODE register bit BRG\_MODE which switches the mode of the bridge (ASYNC/SYNC) is assigned only when in addition the bit BRG\_RST is set to '1'.

**GTM\_AI.141 TIM: Incorrect data captured to GPR registers and routed via ARU when EGPRi\_SEL,GPRi\_SEL= 100 in TIM channel mode TIEM, TPWM, TIPM, TPIM, TGPS**

In case of a TIM channel capture event issued by a rising edge at TIM[i]\_CH[x]\_FOUT the capturing of the TIM[i]\_CH[x]\_ECNT register to the TIM[i]\_CH[x]\_GPRi register is incorrect. The captured value will be ECNT\_REG+2; bit 0 (signal level) will be 0. The correct operation would be to capture ECNT\_REG+1; bit 1 (signal level) would be 1.

### Scope

TIM.

*Note: The described effects related to the ARU do not apply to devices where no ARU is implemented.*

### Effects

- a) Inconsistency of ARU signal level bit and bit[0] of ARU word which shows the captured ECNT.
- b) Reading of TIM[i]\_CH[x]\_GPRi shows inconsistency when comparing bits [31:24] to [7:0]. At the point in time of capture event the bits [31:24] contain the correct value and are subject to be changed with new incoming edge.

### Workaround

- a) When using captured data via ARU routing the correct data can be reconstructed by:

```
IF ARU_SIGNAL_LEVEL ==1 AND ARU_DATA[0] == 0 THEN ARU_DATA = ARU_DATA -1;
```

- b) When reading TIM[i]\_CH[x]\_GPRi by configuration interface the data can be corrected as long as there is no GPR overflow and no new edge by:

```
IF TIM[i]_CH[x]_GPRi[24] == 1 AND TIM[i]_CH[x]_GPRi[0] == 0 THEN TIM[i]_CH[x]_GPRi[23:0] = TIM[i]_CH[x]_GPRi[23:0] -1
```

**GTM\_AI.142 TIM: Incorrect data captured to GPR registers and routed via ARU when EGPRi\_SEL, GPRi\_SEL= 100 in TIM channel mode TBCM**

In case of a TIM channel capture event issued by an input pattern match to condition TIM[i]\_CH[x]\_CNTS the capturing of the TIM[i]\_CH[x]\_ECNT register to the TIM[i]\_CH[x]\_GPRi register can be incorrect. Starting at t=0 with counter value ECNT\_REG(t=0), the captured values of two consecutive edges can be ECNT\_REG(t=0)+2 followed by ECNT\_REG(t=0)+2 instead of ECNT\_REG(t=0)+1 followed by ECNT\_REG(t=0)+2.

**Scope**

TIM.

*Note: The described effects related to the ARU do not apply to devices where no ARU is implemented.*

**Effects**

- a) In 2 following ARU transfers the ARU word which shows the captured ECNT do not increment by 1.
- b) Reading of TIM[i]\_CH[x]\_GPRi shows inconsistency between [31:24] and [7:0]

**Workaround**

- a) Ignore captured data via ARU and build with MCS independent counter which increments on each ARU transfer.
- b) When reading TIM[i]\_CH[x]\_GPRi by configuration interface use only TIM[i]\_CH[x]\_GPRi[31:24] as EDGE counter; don't use TIM[i]\_CH[x]\_GPRi[23:0].

**GTM\_AI.143 GTM\_TOP level: AEI pipelined write to GTM\_BRIDGE\_MODE register directly after setting aei\_reset='0' can result in blocking of AEI configuration interface**

If the GTM bus bridge is reset with aei\_reset= '0' (this means reset by application or module/kernel reset) and the next AEI transfer is a write command to GTM\_BRIDGE\_MODE register the AEI configuration interface can be blocked.

**Scope**

AEI pipelined protocol.

**Effects**

GTM Bus interface does not issue aei\_ready which could lead to bus timeout of the serving bus master.

**Workaround**

Ensure that after setting aei\_reset to inactive state (this means after resetting the GTM by application or module/kernel reset) the next command must be a read to any other register except GTM\_BRIDGE\_MODE. Issue desired write to GTM\_BRIDGE\_MODE register afterwards.

**GTM\_AI.144 TIM: TIM interrupts as trigger source from TIM to TOM/ATOM not functional**

According to specification one could select with the configuration bits EXT\_CAP\_SRCx(2:0) of register TIM[i]\_CH[x]\_CTRL one of six TIM channel x+1 interrupts as a source for signal TIM\_EXT\_CAPTURE(x).

The signal is used internally in TIM channel x and forwarded to a corresponding ATOM/TOM channel.

For the signal path TIM\_EXT\_CAPTURE(x) which is forwarded to ATOM/TOM the selection is incorrect for the values of EXT\_CAP\_SRCx(2:0) = 000, 010, 100, 101, 110, 111.

Only the selection of TIM\_IN(x-1), TIM\_IN(x) or AUX\_IN(x) is possible with the values EXT\_CAP\_SRCx(2:0) = 001 or 011.

For the signal path TIM\_EXT\_CAPTURE(x) which is used inside TIM channel x, the selection works as specified.

### Scope

TIM.

### Effects

The selection of an interrupt of TIM channel x+1 by EXT\_CAP\_SRCx(2:0) = 000, 010, 100, 101, 110, 111 to trigger corresponding TOM/ATOM channel leads to erroneous trigger behavior.

As a result the TOM/ATOM does not react on the intended interrupt.

### Workaround

None.

Do not use the configuration EXT\_CAP\_SRCx(2:0) = 000, 010, 100, 101, 110, 111.

### **GTM\_AI.153 TIM: Incorrect data captured to CNTS register when TIM channel operates in mode TPWM or TPIM and CNTS\_SEL = 1 and selected CMU\_CLK ≠ sys\_clk**

In case of CNTS\_SEL = 1 and TIM\_MODE = TPWM or TPIM in the CNTS\_REG register the value of TBU\_TS0 shall be captured. This does not happen when the selected CMU\_CLK ≠ sys\_clk.

### Scope

TIM.

### Effects

Unexpected values in CNTS\_REG.

## Workaround

Setup the TIM channel to operate on a CMU\_CLK (Divider =1) which is identical to sys\_clk. Please notice that the measurement with TIM\_CNT has resolution of sys\_clk.

### **GTM\_AI.154 TOM: Incorrect duty cycle in PCM mode (bit reversed mode)**

The generated duty cycle on the TOM output in PCM mode is always one smaller than the configured value in the CM1 register. So if the value 1 is configured, a duty cycle of 0% will be generated. Configuring the max value (0xFFFF) in the CM1 register results in a duty cycle of max-1. Expected is 100% duty cycle in this case. A zero in CM1 register results in 100% duty cycle.

## Scope

TOM.

## Effects

Unexpected duty cycle in PCM mode.

## Workaround

Configure always the value for the expected duty cycle in the CM1 register with expected duty cycle + 1.

To get 0% duty cycle, value 1 has to be configured. To get 100% duty cycle, 0 has to be configured to CM1 register while CM0 is always configured with max. value of 0xFFFF. Configuring CM0=0x1000 and CM1=0xFFFF will also get a duty cycle of 100%.

### **GTM\_AI.157 CMU: Incorrect AEI status by writing 1 to bit 24 of register CMU\_CLK\_6/7\_CTRL**

If according to GTM device configuration no DPLL is available, bit 24 of register CMU\_CLK\_6\_CTRL and CMU\_CLK\_7\_CTRL is reserved.

Erroneously, writing a '1' to bit 24 is possible and leads to AEI status 0.

## Scope

CMU: GTM device configurations without DPLL.

## Effects

No functional influence to specified GTM.

After writing a '1' to bit 24 of register CMU\_CLK\_6\_CTRL or CMU\_CLK\_7\_CTRL, a '1' is read back from this register bit.

Writing a '1' to bit 24 of register CMU\_CLK\_6\_CTRL or CMU\_CLK\_7\_CTRL leads to AEI status 0.

## Workaround

Do not write '1' to bit 24 of register CMU\_CLK\_6/7\_CTRL.

## **GTM\_AI.163 TIM: timeout signaled when TDU unit is reenabled**

In the following situation an undesired timeout event is signaled:

After stopping the TDU the TO\_CNT bitfield will have an arbitrary value  $TO\_CNT0 \leq TOV0$  bitfield. Assume TOV will be reconfigured to value TOV1 with  $TOV1 \leq TO\_CNT0$ . If the TDU will be enabled again by writing to TOCTRL a value  $\neq 0$  and at the same time the TCS selected CMU\_CLK has an active edge an unintended timeout is signaled. This results due to the fact that for one clock cycle  $TO\_CNT0 \geq TOV1$ .

## Scope

TIM.

## Effects

Unexpected timeout event when TIM TDU is enabled.

## Workaround

If TDU unit has to be reenabled with a TOV value TOV1 which is less than the previous one in use TOV0 (2 alternatives are available):

- a) Wait with disabling TDU until condition  $TOV1 > TO\_CNT$  is fulfilled. Configure TOV with TOV1 reenable TDU Unit.
- b) Disable TDU; if  $TOV1 \leq TO\_CNT$  write TOV with  $FF_H$ ; enable TDU unit; reconfigure TOV to desired value TOV1.

**GTM AI.164 TIM: capturing of data into TIM[i]\_CH[x]\_CNTS with setting CNTS\_SEL=1 not functional in TPWM and TPIM mode**

If CNTS\_SEL=1 is selected and a new input edge is signaled by the TIM Filter unit while the selected CMU\_CLK has no rising edge the register TIM[i]\_CH[x]\_CNTS will capture data TIM[i]\_CH[x]\_CNT instead of TBU\_TS0.

**Scope**

TIM.

**Effects**

Captured data in TIM[i]\_CH[x]\_CNTS is not as expected.

**Workaround**

- a) Select with CLK\_SEL a CMU\_CLK which is identical to sys\_clk (clock divider=1 applied in CMU channel and for global fractional divider).
- b) Use TIEM mode to capture TBU\_TS0 for rising and falling input edges.
- c) PWM mode: Use CNTS\_SEL=0 with CMU\_CLK source selected as in use for TBU\_TS0 counting. Capture with EGPR0\_SEL=0, GPR0\_SEL=0 in GPR0\_REG TBU\_TS0 and with EGPR1\_SEL=0, GPR1\_SEL= 3 in GPR1\_REG CNT. Calculate the desired timestamp with  $GPR0\_REG - GPR1\_REG + CNTS\_REG$ .

**GTM AI.181 TIM: Incorrect signal level bit ECNT[0] in mode TIEM, TPWM, TIPM, TPIM, TGPS**

In case of re-enabling a previously disabled TIM channel the bit ECNT[0] might not reflect the actual signal level of the corresponding input TIM[i]\_CH[x]\_FOUT

until the next input edge occurs. This situation can only occur if between disabling and re-enabling the ECNT register is not read.

### Scope

TIM.

### Effects

Inconsistency of input signal level with ECNT bit[0].

### Workaround

- After disabling the TIM channel, ensure that the ECNT register is read at least once and afterwards the TIM channel can be re-enabled.
- Before re-enabling a TIM channel, issue a TIM channel reset and reconfigure the TIM channel control registers.

### **GTM\_AI.202 (A)TOM: no CCU1 interrupt in case of CM1=0 or 1 and RST\_CCU0=1**

In case of channel x has configuration of RST\_CCU0=1 (i.e. CN0 is reset by trigger input) and CN0 counts from 0 to MAX:

- if CM1=0, CM0>0 -> no CCU1 interrupt is generated
- if CM1=1, CM0=MAX+1 -> only one time a CCU1 interrupt is generated

### Scope

TOM / ATOM SOMP mode.

### Effects

For the described configuration no CCU1 interrupt is generated.

### Workaround

Use for triggering channel y (i.e. the channel that triggers on channel x the reset of counter CN0) the configuration of CM0=MAX, CM1=1.

In case of duty cycle configuration of  $CM1=0$  and  $CM0>0$  on channel x use instead of CCU1 interrupt on channel x the CCU0 interrupt of triggering channel y.

In case of duty cycle configuration of  $CM1=1$  and  $CM0=MAX+1$  on channel x use instead of CCU1 interrupt on channel x the CCU1 interrupt of triggering channel y.

### **GTM\_AI.205 TIM: unexpected CNTS register update in TPWM OSM mode**

If  $OSM=1$  and  $TIM\_MODE="000"$  (TPWM) an active edge defined by DSL will stop the measurement. In case of an inactive edge following after 1 GTM system clock cycle the active edge the CNTS register will be reset unexpectedly.

#### **Scope**

TIM.

#### **Effects**

Unexpected CNTS register content.

#### **Workaround**

- a) Use CMU clock in TIM channel with frequency lesser than system clock.
- b) Enable filter and configure filter parameter in a way that two consecutive edges will never occur with distance of GTM system clock.

### **GTM\_AI.209 TOM/ATOM: no update of CM0/CM1/CLK\_SRC via trigger signal from preceding instance if selected CMU\_CLKx is not SYS\_CLK**

The trigger signal between (A)TOM instances (e.g. signal  $TOM\_TRIG\_i$ ) is registered between each TOM and between each 2nd ATOM and with this delayed by one  $SYS\_CLK$  period to break long combinational path.

For each register in the trigger path between (A)TOM instance i and the succeeding (A)TOM instance i+1, this trigger from instance i does not trigger the update of register CM0, CM1 and CLK\_SRC with content of SR0, SR1 and

CLK\_SRC\_SR if the triggered channel of instance i+1 is not running with a selected CMU\_CLKx = SYS\_CLK.

### Scope

TOM/ATOM.

### Effects

In the described configuration no update of CM0, CM1 and CLK\_SRC is done although the update is enabled by register TOM[i]\_TGC[y]\_GLB\_CTRL / ATOM[i]\_AGC\_GLB\_CTRL.

### Workaround

For each register in trigger path between (A)TOM instance i and (A)TOM instance i+1, the channel of instance i+1 that should be triggered has to use a clock of period identical to SYS\_CLK period.

A second workaround could be to set up on instance i+1 a redundant channel to trigger other channel of instance i+1 like it was set up on instance i to trigger other channel. Then, start both instances synchronously by using the TBU time base comparator of AGC/TGCx unit (i.e. the ATOM[i]\_AGC\_ATC\_TB / TOM[i]\_TGC[y]\_ACT\_TB register).

### **GTM AI.260 TOM/ATOM: Async. update in SOMP mode with CM1=0 and selected CMU clock unequal sys\_clk not functional**

*Note: In TC23x/TC22x/TC21x devices, this problem relates to the following scenario in TOM: Async. update with CM1=0 and selected CMU\_FXCLK unequal to sys\_clk not functional.*

An asynchronous update of the duty cycle by writing value 0 to CM1 register while a CMU clock unequal sys\_clk is selected is not working. It is expected that the output signal level is set immediately to inactive level but it will remain at actual level.

### Scope

TOM/ATOM.

## Effects

The output signal level is not set to inactive level. It will remain at actual level.

## Workaround

Writing value 1 instead of 0 to CM1 register will set the output to inactive level in the actual generated PWM period.

If the duty cycle duration should be zero also for the following period, the user has to take care, that the CM1 register is loaded with a 0 at the beginning of the next PWM period.

Otherwise, if the content of register CM1 remains at 1, a peak of one clock cycle with the selected CMU clock will be observed, with the next PWM period.

## **GTM\_AI.270 (A)TOM: output signal is postponed one period for the values CM0=1 and CM1>CM0 if CN0 is reset by the trigger of a preceding channel (RST\_CCU0=1)**

If counter CN0 is reset by the trigger of a preceding channel (bit RST\_CCU0 of register TOM[i]\_CH[x]\_CTRL/ATOM[i]\_CH[x]\_CTRL is set), then the value of CM0 defines the signal edge to SL (signal level), whereas CM1 defines the edge to !SL (inverted signal level).

If - in this case - the value 1 is configured for the output edge to SL (CM0=1) and CM1 is configured to greater than CM0 (CM1>CM0) the expected output edge will be postponed by one period.

## Scope

TOM, ATOM SOMP mode

## Effects

The expected output edge will be postponed by one period.

## Workaround

Instead of configuring CM0=1 it is also possible to configure CM1=1 and to invert SL to get the expected edge at counter value 1 (CN0=1).

**GTM\_AI.298 TOM/ATOM: wrong output behaviour in SOMP oneshot mode when oneshot pulse is triggered by TIM\_EXT\_CAPTURE(x)**

If TOM/ATOM is configured in SOMP oneshot mode (OSM = 1) and the oneshot trigger is configured to TIM\_EXT\_CAPTURE(x) (OSM\_TRIG = 1, EXT\_TRIG = 1) the output behaviour is not as expected depending on the selected CMU clock.

1. If the selected CMU clock is configured to sys\_clk (ATOM: CMU\_CLK\_[z]\_CTRL = 0, TOM: CMU\_FXCLK0 used) no initial oneshot period (CN0 is set to zero and then counts until CN0 >= CM0) is executed and the output is set to SL immediately and not as expected after the first initial period.
2. If the selected CMU clock is configured to CMU\_CLK\_[z]\_CTRL > 0 (ATOM)/CMU\_FXCLK[1..n] (TOM) then an initial period is executed but the output is set immediately to SL and not as expected when the second oneshot period starts.

**Scope**

TOM/ATOM SOMP oneshot mode

**Effects**

The TOM/ATOM output is set immediately to SL and not as expected with a delay of the first initial oneshot period.

**Workaround**

For GTM generation v2 no workaround is available.

If it is possible configure the selected CMU clock to sys\_clk period. Then the generated oneshot pulse length is correct but without executing of the initial period.

**GTM\_AI.299 TOM/ATOM: wrong output behaviour in SOMP oneshot mode when oneshot pulse is triggered by trig\_[x-1]**

If TOM/ATOM is configured in SOMP oneshot mode (OSM = 1) and the oneshot trigger is configured to trigger signal from trigger chain trig\_[x-1] (OSM\_TRIG = 1, EXT\_TRIG = 0) the output signal is set immediately to SL and not as expected after a delay of the first initial oneshot period (CN0 counts from 0 until it reaches the value of CM0). The first initial oneshot period isn't executed.

**Scope**

TOM/ATOM SOMP oneshot mode

**Effects**

The TOM/ATOM output is set immediately to SL and not as expected with a delay of the first initial oneshot period.

**Workaround**

For GTM generation v2 no workaround is available.

If it is possible work without the initial period for GTM generation v2 because the generated pulse length is correct.

**GTM\_AI.336 GTM Bus Bridge: Incorrect AEI access execution in case the previous AEI access was aborted with the access timeout abort function**

In case the GTM internal AEI access timeout abort function is in use (GTM\_CTRL.TO\_VAL != 0 and GTM\_CTRL.TO\_MODE=1), a following AEI access can be corrupted:

- a) A write access might not be executed (register/ memory not written to the specified value)
- b) A read access can return random data (read value does not reflect the content of the addressed register / memory).

Hint: As a timeout based abort of a GTM register access is assumed to be an error scenario, the internal state of the GTM might be exposed. To ensure the

proper behavior after such a severe incident, the GTM IP should be re-initialized as part of a recovery action on system level.

### Scope

CPU interface accesses

### Effects

Read access returns random data.

Write access does not change the content of the target address.

### Workaround

Do not use the AEI access abort mode, use the observe mode instead (Set GTM\_CTRL.TO\_MODE=0).

Enable additionally the timeout observe IRQ by setting GTM\_IRQ\_EN.AEI\_TO\_XPT\_IRQ=1 to invoke higher level recovery mechanisms for GTM re-initialization.

(e.g. abort the pending access to the GTM and re-initialize the GTM\_IP from hardware reset).

### **GTM\_AI.340 TOM/ATOM: Generation of TRIG\_CCU0/TRIG\_CCU1 trigger signals skipped in initial phase of A/TOM SOMP one-shot mode**

#### Configuration in use:

- A/TOM[i]\_CH[x]\_CTRL.OSM=1
- A/TOM[i]\_CH[x]\_CTRL.OSM\_TRIG=0
- A/TOM[i]\_CH[x]\_CTRL.UDMODE=00
- ATOM[i]\_CH[x]\_CTRL.MODE=10

#### Expected behavior:

The generation of one-shot pulses in A/TOM can be initiated by a write to CN0. In this case the pulse generation comprises of an initial phase where the signal level at A/TOM output is inactive followed by a pulse. The duration of the initial phase can be controlled by the written value of CN0, where the duration is

defined by CM0-CN0. After the counter CN0 reaches the value of CM0-1, the pulse starts with its active edge, CN0 is reset, and starts counting again. When CN0 reaches CM1-1, the inactive edge of the pulse occurs. Due to the fact, that the capture compare units CCU0 and CCU1 compare also in the initial phase of the pulse generation, the trigger conditions for these comparators apply also in this initial phase. Thus, the TRIG\_CCU0 and TRIG\_CCU1 signals also occur in the initial phase of the one-shot pulse. When these trigger signals are enabled in the A/TOM[i]\_CH[x]\_IRQ\_EN, an interrupt signal is generated by A/TOM on the CCU0TC and CCU1TC trigger conditions and the corresponding A/TOM[i]\_CH[x]\_IRQ\_NOTIFY bits are set.

**Observed behavior:**

For certain start values of CN0 and dependent on the history of pulse generation, the trigger signals TRIG\_CCU0 and TRIG\_CCU1 are skipped. As a consequence, this can led to missing interrupts CCU0TC and CCU1TC on behalf of their missing trigger signals TRIG\_CCU0 and TRIG\_CCU1.

For the first pulse generation after enabling the channel, all trigger signals TRIG\_CCU0 and TRIG\_CCU1 appear as expected and described in the section expected behavior. If the channel stays enabled and a new value CN0 is written to trigger a subsequent one-shot pulse, the TRIG\_CCU0/TRIG\_CCU1 triggers in the initial phases of subsequent one-shot pulses are skipped under the following conditions:

- For TRIG\_CCU0 trigger: if the one-shot pulse is started by writing a value to CN0 greater or equal to CM0-1.
- For TRIG\_CCU1 trigger: if the one-shot pulse is started by writing a value to CN0 greater or equal to CM1-1.

**Scope**

TOM/ATOM

**Effects**

Missing TRIG\_CCU0 and TRIG\_CCU1 trigger signals in initial phase of subsequent pulses in A/TOM one-shot mode, when one shot-mode is started with writing to CN0 values greater equal CM0-1 or CM1-1.

### Workaround 1

Disabling, resetting (channel reset), re-enabling and initializing of the channel between each one-shot pulse will ensure the correct behavior of CCU0TC and CCU1TC interrupt source.

### Workaround 2

Starting a new one-shot pulse by writing twice the counter CN0 whereas the first value, which is written to CN0 should be zero followed by the value which defines the length of the initial phase.

Be aware that in this case, the total length of the initial phase until the pulse is started, is influenced by the time between the two write accesses to CN0.

**GTM\_AI.341 TOM/ATOM: False generation of TRIG\_CCU1 trigger signal in SOMP one-shot mode with OSM\_TRIG=1 when CM1 is set to value 1**

#### Configuration in use:

- A/TOM[i]\_CH[x]\_CTRL.OSM=1
- A/TOM[i]\_CH[x]\_CTRL.OSM\_TRIG=1
- A/TOM[i]\_CH[x]\_CTRL.UDMODE=00
- ATOM[i]\_CH[x]\_CTRL.MODE=10

#### Expected behavior:

The generation of one-shot pulses in A/TOM can be initiated by the trigger event TRIG\_[x-1] from trigger chain or by TIM\_EXT\_CAPTURE(x) trigger event from TIM, whereas the counter CN0 is reset to zero and starts counting. In this case the pulse generation comprises of an initial phase where the signal level at A/TOM output is inactive followed by a pulse. The duration of the initial phase is always as long until the counter CN0 reaches CM0-1.

After the counter CN0 reaches the value of CM0-1, the pulse starts with its active edge, CN0 is reset, and starts counting again. When CN0 reaches CM1-1, the inactive edge of the pulse occurs. Due to the fact, that the capture compare units CCU0 and CCU1 compare also in the initial phase of the pulse generation, the trigger conditions for these comparators apply also in this initial phase. Thus, the TRIG\_CCU0 and TRIG\_CCU1 signals also occur in the initial

phase of the one-shot pulse. When these trigger signals are enabled in the A/TOM[i]\_CH[x]\_IRQ\_EN, an interrupt signal is generated by A/TOM on the CCU0TC and CCU1TC trigger conditions and the corresponding A/TOM[i]\_CH[x]\_IRQ\_NOTIFY bits are set.

### Observed behavior:

If the compare register CM1 is set to 1 and a new one-shot pulse is triggered, two effects can be observed:

- The first observed behavior is that the capture compare unit doesn't generate the TRIG\_CCU1 trigger signal in the initial phase of the one-shot cycle.
- The second observed behavior is that at the end of the operation phase of the one-shot cycle, where CN0 reaches CM0-1 a second time, the capture compare unit generates a TRIG\_CCU1 trigger signal which is not expected at this point in time.

### Scope

TOM/ATOM

### Effects

Missing TRIG\_CCU1 trigger signal in initial phase of the one-shot cycle and unexpected TRIG\_CCU1 trigger signal at the end of the operation phase of the one-shot cycle.

### Workaround

Instead of using value 1 for CM1 it could be possible to generate the same pulse length by using a higher CMU\_FXCLK/CMU\_CLK frequency. Then, to get the same pulse length, the value of CM1 has to be multiplied by the difference of the two CMU\_FXCLK/CMU\_CLK frequencies.

Be aware that this workaround is only possible, if you are not already using the CMU\_FXCLK(0) because there is no higher CMU\_FXCLK frequency to select.

**Example for TOM:** Instead of using CMU\_FXCLK(1), which has the divider value  $2^{**4}$ , use CMU\_FXCLK(0), which has the divider value  $2^{**0}$ . In this case, CM1 has to be configured with value  $2^{**4}$  minus  $2^{**0}$  which is equal to  $2^{**4}-1=16$ .

**Hint:** To get the same length of period, which defines the length of the initial phase, the value for the period in CM0 has to be multiplied by the same value.

A second limitation is that the maximum length of the period, which is configured in CM0, is limited. Using a higher CMU\_FXCLK/CMU\_CLK frequency reduces the maximum possible period.

### **GTM\_AI.347 TOM/ATOM: Reset of (A)TOM[i]\_CH[x]\_CN0 with TIM\_EXT\_CAPTURE are not correctly synchronized to selected CMU\_CLK/CMU\_FXCLK**

To reset the counter (A)TOM[i]\_CH[x]\_CN0 (SOMP mode in ATOM), the input signal TIM\_EXT\_CAPTURE can be used by configuration of (A)TOM[i]\_CH[x]\_CTRL.EXT\_TRIG=1 and (A)TOM[i]\_CH[x]\_CTRL.RST\_CCU0=1.

The reset of the counter (A)TOM[i]\_CH[x]\_CN0 should happen synchronously to the internal selected CMU clock CMU\_CLK/CMU\_FXCLK. Therefore a synchronisation stage is implemented to synchronize the input signal TIM\_EXT\_CAPTURE to the internal selected CMU clock CMU\_CLK/CMU\_FXCLK.

It can be observed, that the reset of the counter is done immediately with the occurrence of the input signal TIM\_EXT\_CAPTURE and not as expected synchronously to the selected CMU clock enable CMU\_CLK/CMU\_FXCLK.

As a consequence of this, the output signal for the compare values 0 and 1 of (A)TOM[i]\_CH[x]\_CM1.CM1 and (A)TOM[i]\_CH[x]\_CM0.CM0 will not be set correctly.

#### **Scope**

ATOM, TOM

#### **Effects**

The output signal (A)TOM[i]\_CH[x]\_OUT is not set correctly for the compare values 0 and 1 of the operation register bitfields (A)TOM[i]\_CH[x]\_CM1.CM1 and (A)TOM[i]\_CH[x]\_CM0.CM0.

### Workaround 1

Select a CMU clock enable signal CMU\_CLK/CMU\_FXCLK by appropriate setting of (A)TOM[i]\_CH[x]\_CTRL.CLK\_SRC which is setup inside the CMU module in that way, that each system clock is enabled. In other words this means that the selected clock enable signal CMU\_CLK/CMU\_FXCLK should be always active high.

*Note: No frequency divider should be used for CMU\_CLKz (only CMU\_CLK\_z\_CTRL.B.CNT = 0) and CMU\_FXCLKx (only CMU\_FXCLK0).*

### Workaround 2

Avoid the compare values 0 and 1 for the operation register bitfields (A)TOM[i]\_CH[x]\_CM1.CM1 and (A)TOM[i]\_CH[x]\_CM0.CM0.

### **GTM\_AI.361 IRQ: Missing pulse in single-pulse interrupt mode on simultaneous interrupt and clear event**

In single-pulse interrupt mode ([MODULE]\_IRQ\_MODE = 0b11) only the first interrupt event of the interrupt bits of the interrupt notify register inside this module generates a pulse on the output signal IRQ\_line, if the associated interrupt is enabled ([MODULE]\_IRQ\_EN=1). All further interrupt events have no effect on the output signal IRQ\_line until all enabled interrupts are cleared, except when an interrupt and a clear event (HW\_clear or a SW\_clear) occur at the same time.

#### **Expected behaviour:**

On simultaneous occurrence of an interrupt and clear event, a pulse on the output signal IRQ\_line is generated.

#### **Observed behaviour:**

If the associated notify register bit of the interrupt event is not set and another bit of the same notify register is set and this interrupt is enabled, no pulse on the output signal IRQ\_line is generated.

All modules ([MODULE]) are affected by this ERRATUM, which are able to generate interrupts and which have multiple interrupt sources which are ORed to the output. Not affected are the modules DPLL and ARU.

## Scope

IRQ

## Effects

Missing pulse on interrupt signal IRQ\_line.

All modules, which deliver an interrupt signal and have more than one internal interrupt source which are ORed are affected. The only exceptions are the modules ARU and DPLL.

## Workaround

On a SW clear prevent HW clear events and read the interrupt notify register to check on new interrupts without a received interrupt pulse on IRQ\_line. In this case repeat the SW clear step to enable interrupt generation again.

When disabling the HW clear is not an option refrain from using the single-pulse interrupt mode.

## **GTM\_AI.380 (A)TOM: potentially wrong output signal in case of RST\_CCU0=1 and CM0=1 on triggered channel in SOMP mode**

When the reset of (A)TOM\_CHx\_CN0 of a TOM or ATOM channel is triggered by a preceding channel or assigned TIM module (RST\_CCU0=1) and the ATOM channel is configured in SOMP mode, the CM0 value defines the edge to SL and CM1 defines the edge to !SL.

## **Expected behavior:**

When SR0 is configured to '1', and CM0 is updated with SR0=1 on trigger signal coming from previous channel, an edge to SL is expected, when CN0=CM0=1.

**Observed behavior:**

When CM0 is updated synchronously from SR0 for the next period, and  $CM0 > 1$  at the actual period, no edge to SL is generated when  $CM0 = CN0 = 1$  for the first period after  $CM0 = 1$  becomes active (was updated to  $CM0 = 1$  from SR0).

**Scope**

TOM, ATOM

**Effects**

For the configuration  $RST\_CCU0 = 1$  and  $CM0 = 1$ ,  $CM0 < CM1$  no edge is generated for the first period, after  $CM0$  is updated from SR0 with '1' and  $CM0 > 1$  in the period before.

**Workaround**

In addition to configuring  $SR0 = 1$  and letting the (A)TOM channel update  $CM0$  with '1' at the start of the next period, a hot reconfiguration of  $CM0 = 1$  can be done. However, the hot reconfiguration needs to be done after the edge to SL was performed in the actual period. Otherwise the  $CM0$  value would be overwritten by '1' and the edge to SL would be generated immediately after hot reconfiguration and not at the intended old  $CM0$  value.

The workaround is applicable where the system can update the  $CM0$  value in time; otherwise the setting of  $CM0 = 1$  should not be used.

**GTM\_AI.408 (A)TOM-RTL: Missing edge on output signal (A)TOM\_OUT when CN0 is reset with force update event****Description**

The channel is configured in continuous up-counter mode. Then a new period is started with a force update event and reset of  $CN0$  is activated.

**Configuration for TOM:**

`TOM[i]_CH[x]_CTRL.UDMODE=0`

`TOM[i]_TGC[g]_FUPD_CTRL.FUPD_CTRL[k]=1`

TOM[i]\_TGC[g]\_FUPD\_CTRL.RSTCN0\_CH[k]=1

### Configuration for ATOM:

ATOM[i]\_CH[x]\_CTRL.MODE=0b10 (SOMP mode)

ATOM[i]\_CH[x]\_CTRL.UDMODE=0

ATOM[i]\_AGC\_FUPD\_CTRL.FUPD\_CTRL[k]=1

ATOM[i]\_AGC\_FUPD\_CTRL.RSTCN0\_CH[k]=1

### Expected behavior:

After the counter (A)TOM[i]\_CH[x]\_CN0.CN0 has been reset and therefore a new period has to be started and the output signal (A)TOM\_OUT has to be set immediately to the SL value (ATOM[i]\_CH[x]\_CTRL\_SOMP.SL, TOM[i]\_CH[x]\_CTRL.SL), and after the counter reaches (A)TOM[i]\_CH[x]\_CM1.CM1, an edge on (A)TOM\_OUT to the inverted SL value (ATOM[i]\_CH[x]\_CTRL\_SOMP.SL, TOM[i]\_CH[x]\_CTRL.SL) is expected.

### Observed behavior:

An edge on the output signal (A)TOM\_OUT to the SL value (ATOM[i]\_CH[x]\_CTRL\_SOMP.SL, TOM[i]\_CH[x]\_CTRL.SL) at the beginning of the new period does not happen. Instead, the output signal (A)TOM\_OUT holds its last value.

A second observation is in case the SL value (ATOM[i]\_CH[x]\_CTRL\_SOMP.SL, TOM[i]\_CH[x]\_CTRL.SL) changes synchronously together with the force update event, an edge on (A)TOM\_OUT to the inverted SL value (ATOM[i]\_CH[x]\_CTRL\_SOMP.SL, TOM[i]\_CH[x]\_CTRL.SL) when (A)TOM[i]\_CH[x]\_CN0.CN0 reaches (A)TOM[i]\_CH[x]\_CM1.CM1 does not happen.

### Scope

TOM, ATOM

### Effects

Missing edge and false output signal level on (A)TOM\_OUT

## Workaround

No workaround available.

### **GTM\_AI.411** A change of the BRIDGE\_MODE register might be delayed indefinitely

#### Description

After a write access to the BRIDGE\_MODE register, the bit fields BRG\_MODE and BYPASS\_SYNC will not be updated until the transaction buffer is empty. In split mode the bridge allows new transactions to be added to the buffer, even when an update of these bits is pending.

Polling the register in split mode might prevent the buffer from getting empty and as a result prevent the actual update of the described bit fields.

*Note: bit field BYPASS\_SYNC is not specified for TC2xx.*

#### Scope

GTM\_AEI

#### Effects

Frequently polling the BRIDGE\_MODE register ends in a deadlock.

#### Workaround 1

After every failed attempt to read back the new values, increase the wait time before issuing the next read transaction.

#### Workaround 2

Use standard mode (which is entered by setting AEI\_PIPE and AEI\_SPLIT at zero while asserting AEI\_SEL) to write and read back the affected bits.

*Note: This workaround is only possible in devices without AXIS.*

**GTM\_AI.419 TIM: Potentially wrong capture values****Description****Configuration**

The TIM channel is configured in TIEM, TIPM, TGPS or TSSM mode by setting of  $TIM[i]_{CH[x]}_{CTRL}.TIM\_MODE = \{010_B, 011_B, 101_B, 110_B\}$ . The TIM channel is disabled ( $TIM[i]_{CH[x]}_{CTRL}.TIM\_EN = 0$ ) and later enabled again ( $TIM[i]_{CH[x]}_{CTRL}.TIM\_EN = 1$ ).

**Expected behavior for TIEM/TIPM/TGPS mode**

The registers  $TIM[i]_{CH[x]}_{CNT}$ ,  $TIM[i]_{CH[x]}_{ECNT}.ECNT[15:1]$ ,  $TIM[i]_{CH[x]}_{GPR0}$  and  $TIM[i]_{CH[x]}_{GPR1}$  are set to their reset values. In case of an input signal edge or an input capture event or an active selected CMU clock (TGPS mode) at the same time as the channel is enabled, this event has to be taken into account and the  $TIM[i]_{CH[x]}_{CNT}$  register must be updated/incremented based on its reset value. Due to this a capture event can happen depending on the configured TIM mode and the register values.

**Expected behavior for TSSM mode**

The registers  $TIM[i]_{CH[x]}_{CNT}$ ,  $TIM[i]_{CH[x]}_{ECNT}.ECNT[15:1]$ ,  $TIM[i]_{CH[x]}_{GPR0}$  and  $TIM[i]_{CH[x]}_{GPR1}$  are set to their initial values. The initial value for  $TIM[i]_{CH[x]}_{CNT}$  register depends on  $TIM[i]_{CH[x]}_{CTRL}.ISL$  and  $TIM[i]_{CH[x]}_{CNTS}.CNTS(22)$ . If  $TIM[i]_{CH[x]}_{CNTS}.CNTS(22)$  is set to 0 and  $TIM[i]_{CH[x]}_{CTRL}.ISL$  is set to 0 the initial value of  $TIM[i]_{CH[x]}_{CNT}$  is 0x000000. An input signal event simultaneously to the channel enable is not taken into account.

**Observed behavior for TIEM/TIPM/TGPS mode**

If no input signal event or input capture event or active selected CMU clock (TGPS mode) occurs, the registers  $TIM[i]_{CH[x]}_{CNT}$ ,  $TIM[i]_{CH[x]}_{ECNT}.ECNT[15:1]$ ,  $TIM[i]_{CH[x]}_{GPR0}$  and  $TIM[i]_{CH[x]}_{GPR1}$  are set to their reset values as expected.

If an input signal event or an input capture event or an active selected CMU clock (TGPS mode) occurs at same time as the channel gets enabled, the

TIM[i]\_CH[x]\_CNT register continues to count (or update) based on the previous (old) value. As a result, a capture could be performed too early and/or with the wrong values.

The TIM[i]\_CH[x]\_ECNT.ECNT[15:1] register is set to its reset value as expected.

### Observed behavior for TSSM mode

The register TIM[i]\_CH[x]\_CNT is not set to its initial value of 0x000000 on channel enabling when TIM[i]\_CH[x]\_CNTS.CNTS(22) is set to 0 and TIM[i]\_CH[x]\_CTRL.ISL is set to 0.

*Note: The TIM channel modes TPWM, TPIM and TBCM*

*(TIM[i]\_CH[x]\_CTRL.TIM\_MODE = {000<sub>B</sub>, 001<sub>B</sub>, 100<sub>B</sub>}) are not affected.*

### Scope

TIM

### Effects

TIM[i]\_CH[x]\_CNT register is not reset and wrong values could be captured into TIM[i]\_CH[x]\_GPR0 and TIM[i]\_CH[x]\_GPR1 registers.

### Workaround 1

Reset the TIM channel by setting of TIM[i]\_RST.RST\_CH[x] = 1 before enabling the TIM channel.

### Workaround 2

The following sequence has to be executed on the disabled channel but before the actual enabling of the channel, to ensure that the TIM[i]\_CH[x]\_CNT register is set to its reset value when the channel is enabled:

1. Configure TIM[i]\_CH[x]\_CNTS = 0
2. Enable the TIM channel with the following configuration inside the TIM[i]\_CH[x]\_CTRL register:
  - TIM\_EN = 1
  - TIM\_MODE = 101<sub>B</sub> (TGPS)
  - ISL = 1

- OSM = 1
  - ARU\_EN = 0
  - select a fast CMU\_CLK\_RES, for example CLK\_SEL = 000<sub>B</sub>
3. Wait until an edge on the selected CMU\_CLK\_RES occurs. This can be observed on the NEWVAL IRQ notify register. This event sets the TIM[i]\_CH[x]\_CNT register to its reset value.
  4. Disable TIM channel (TIM[i]\_CH[x]\_CTRL.TIM\_EN = 0)
  5. Configure the former TIM channel configuration in TIM[i]\_CH[x]\_CTRL register and enable the TIM channel again.

### **GTM\_AI.429 TIM: Missing glitch detection interrupt event**

#### **Description**

#### **Configuration**

TIM filter is configured in immediate edge propagation mode by setting

TIM[i]\_CH[x]\_CTRL.FLT\_MODE\_RE = 0 or

TIM[i]\_CH[x]\_CTRL.FLT\_MODE\_FE = 0.

The filter is enabled by setting TIM[i]\_CH[x]\_CTRL.FLT\_EN = 1.

#### **Expected behavior**

As long as the filter threshold is not reached and the input signal level unexpectedly changes (it is an input glitch occurs), the internal glitch detection interrupt event signal (TIM\_GLITCHDET\_IRQ) should have a HIGH pulse of one cluster clock cycle.

#### **Observed behavior**

When the input signal glitch occurs at the same time the filter counter reaches its threshold, the internal glitch detection interrupt event signal (TIM\_GLITCHDET\_IRQ) does not occur.

#### **Scope**

TIM

## Effects

The TIM[i]\_CH[x]\_IRQ\_NOTIFY.GLITCHDET bit is not set. Thus, no interrupt is triggered. Furthermore, the external capture source EXT\_CAPTURE(x) is not triggered if its source is set to TIM\_GLITCHDET\_IRQ.

## Workaround

The filter counter threshold can be set to the next higher value. Thus, a former not detected glitch would be detected. In that case, the output signal would be changed (one clock cycle longer) when the input signal is a single cycle pulse.

## **GTM AI.430 TIM: Unexpected increment of filter counter**

### Description

#### Configuration

TIM filter is configured in immediate edge propagation mode by setting

TIM[i]\_CH[x]\_CTRL.FLT\_MODE\_RE = 0 or

TIM[i]\_CH[x]\_CTRL.FLT\_MODE\_FE = 0.

The filter is enabled by setting TIM[i]\_CH[x]\_CTRL.FLT\_EN = 1.

The filter counter threshold is set to 0 by setting either

TIM[i]\_CH[x]\_FLT\_RE.FTL\_RE = 0 or TIM[i]\_CH[x]\_FLT\_FE.FTL\_FE = 0.

#### Expected behavior

When the input signal level changes, the filter counter should stay at 0.

#### Observed behavior

When the input signal level changes, the filter counter counts to 1 and is not reset.

#### Scope

TIM

## Effects

If an input edge occurs during the acceptance time, the following output signal change will happen one selected CMU clock cycle earlier than expected.

## Workaround

If acceptable, use a threshold greater than 0. Otherwise there is no workaround available.

## **GTM\_AI.431 TIM: Glitch detection interrupt event of filter is not a single cycle pulse**

### Description

### Configuration

The TIM filter must be enabled by setting `TIM[i]_CH[x]_CTRL.FLT_EN = 1`.

### Expected behavior

As long as the filter threshold is not reached and the input signal level changes unexpectedly, the glitch detection interrupt event signal (TIM\_GLITCHDET\_IRQ) should have a single cycle HIGH pulse.

### Observed behavior

When the input signal level changes unexpectedly for longer than one clock cycle, the glitch detection interrupt event signal (TIM\_GLITCHDET\_IRQ) is HIGH as long as the unexpected signal change is present.

### Scope

TIM

### Effects

- Effect 1: The longer lasting HIGH signal of the glitch detection interrupt event signal (TIM\_GLITCHDET\_IRQ) may lead to an unexpected behavior

within the GTM only if TIM\_GLITCHDET\_IRQ is used for the external capture signal EXT\_CAPTURE(x).

- Effect 2: If the related interrupt notify register (TIM[i]\_CH[x]\_IRQ\_NOTIFY) is cleared by software while the TIM\_GLITCHDET\_IRQ signal is still HIGH, the interrupt will unexpectedly retrigger.

### Workaround

No workaround in hardware.

For the unexpected retrigger of the interrupt directly after an interrupt clear step, the interrupt routine has to consider that the interrupt might be invalid.

### **GTM AI.462** (A)TOM: Missing CCU0TC\_IRQ interrupt signal

#### Description

#### Configuration

The channel is configured in SOMP (ATOM) up-counter mode with up/down counter mode disabled ((A)TOM[i]\_CH[x]\_CTRL.UDMODE=0) or not existing and triggering by a preceding channel with configuration of (A)TOM[i]\_CH[x]\_CTRL.RST\_CCU0=1.

#### Expected behavior

When the counter (A)TOM[i]\_CH[x]\_CN0.CN0 reaches the value of (A)TOM[i]\_CH[x]\_CM0.CM0, the interrupt signal CCU0TC\_IRQ must be triggered.

#### Observed behavior

In the first period after (A)TOM[i]\_CH[x]\_CM0.CM0 is changed to the value 0 or 1, no CCU0TC\_IRQ interrupt signal is triggered.

*Note: When the second period starts after (A)TOM[i]\_CH[x]\_CM0.CM0 is changed to the value 0 or 1 and stays at that value, then the CCU0TC\_IRQ interrupt signal generation works correctly.*

**Scope**

TOM, ATOM

**Effects**

Interrupt signal CCU0TC\_IRQ is not triggered.

**Workaround**

No workaround available.

It needs to be checked if the application can accept the interrupt occurring with the second period.

**GTM\_TC.009 TBU signals not wired to debug logic**

The TBU signals are not wired from the GTM kernel to the OTGB interface. Therefore, TBU signals are not available for debug purposes.

The other GTM signals are connected to the debug system as specified.

*Note: This limitation will not be present on TC23x Emulation Devices.*

**GTM\_TC.012 Read Access Control by Register ODA**

Specific GTM registers have by default “destructive read” behavior as their normal read behavior (see section “GTM Software Debugger Support” in the GTM chapter of the User’s Manual for further details.)

Depending on the reading master and the configuration of bits DREN and DDREN in register GTM\_ODA (OCDS Debug Access Register), the read can be performed “non-destructive” for debug related read operation.

According to the User’s Manual the read is performed “non-destructive” (i.e. debug related read operation)

- for all masters when ODA.DREN = 1<sub>B</sub>,
- for the Cerberus (OCDS) FPI master when ODA.DREN = 0<sub>B</sub> and ODA.DDREN = 0<sub>B</sub>.

## Problem Description

In the current implementation the read is performed “non-destructive” (i.e. debug related read operation)

- for all masters when ODA.DREN = 1<sub>B</sub>,
- for the DMA Partition 2 FPI master when ODA.DREN = 0<sub>B</sub> and ODA.DDREN = 0<sub>B</sub>.

## Workaround

The problem described above has 2 aspects:

### 1. For DMA Partition 2 Access to GTM

When the DMA Partition 2 FPI master is used to perform a normal (“destructive”) read of the GTM registers that by default have “destructive read” behavior as their normal read behavior, setting ODA.DREN = 0<sub>B</sub> and ODA.DDREN = 1<sub>B</sub> is required to avoid an unintended debug related (“non-destructive”) read access that would be caused by this issue.

### 2. For Cerberus (OCDS) Access to GTM

When ODA.DREN = 0<sub>B</sub> and ODA.DDREN = 0<sub>B</sub>, any read access of the Cerberus (OCDS) FPI master to the registers that by default have “destructive read” behavior as their normal read behavior will cause the normal (“destructive”) read behavior. To get the intended debug related (“non-destructive”) read behavior, ODA.DREN needs to be set to 1<sub>B</sub> before each access of the Cerberus and set back to 0<sub>B</sub> afterwards to not affect the access of other FPI masters on the registers described above.

## **IOM TC.002 Missed or spurious IOM events when pulse length exceeds Event Window counter range**

When using the Logic Analyzer Module (LAM) of the IOM, if the 24-bit counter for the Event Window exceeds its maximum value (0xFFFFF) it wraps around and starts counting again from 0x0.

If the Event Window is not inverted (LAMCFG.IVW = 0<sub>B</sub>), for example for measuring long pulses, and the edge that generates an event comes after the

counter exceeded its maximum value, the event will not be generated if the counter, due to the rollover, is again below the threshold value (LAMEWS.THR), outside of the Event Window.

As an additional side effect of the wraparound, spurious events may be generated when expecting an alarm only in case of pulses that are too short, if a pulse is longer than the counter can handle.

### Workaround

Avoid measuring pulses longer than the Event Window counter range.

### **IOM TC.003 Unexpected Event upon Kernel Reset**

If a kernel reset (via bits RST in registers KRST0/1) is performed on the IOM, an unexpected event may be signalled to the SMU.

### Workaround

Before triggering a kernel reset via software, set the alarm reaction in SMU to “No Action” to avoid reaction on the unexpected event.

### **IOM TC.004 Write to IOM register space when IOM\_CLC.RMC > 1**

If a clock divider value RMC > 1 is selected in register IOM\_CLC, more than one write access may be performed to the IOM register address space within one IOM clock cycle.

This will cause unpredictable effects on the internal state for the following scenarios where two (or even multiples of 2) write accesses are performed within one IOM clock cycle to the following register groups:

- ECM registers ECMCCFG and/or ECMSELR, or
- ECM Event Trigger History registers ECMETH0 and/or ECMETH1, or
- FPC registers FPCEsr, FPCCTRk and/or FPCTIMk, or
- LAM registers LAMCFGm and/or LAMEWSm.

*Note: No problem will occur for read accesses.*

## Workaround

Set IOM\_CLC.RMC = 1 when configuring (writing to the registers of) the IOM.  
During runtime (not configuring IOM) IOM\_CLC.RMC > 1 is not an issue.

## **MTU\_TC.005 Access to MCx\_ECCD and MCx\_ETRRi while MBIST disabled**

It is possible to access the memory controller registers MCx\_ECCD and MCx\_ETRRi without the need of the MBIST mode being enabled (i.e. without MTU\_MEMTEST.MEMxEN = 1<sub>B</sub>). This may be used to avoid a complete SRAM initialization on certain security relevant SRAMs.

However, when a MBIST controller is disabled (MTU\_MEMTEST.MEMxEN = 0<sub>B</sub>), there is an inevitable corner case that causes the value read/written from/to registers MCx\_ECCD and MCx\_ETRRi of a disabled MBIST controller to be wrong. There is also a possibility that an SPB error is triggered when accessing the MCx\_ECCD and MCx\_ETRRi registers if other masters concurrently use the SPB bus in this situation.

*Note: No workaround is required to access the registers of an enabled MBIST controller.*

## Workaround

When MBIST mode is disabled (MTU\_MEMTEST.MEMxEN = 0<sub>B</sub>) for a MBIST controller,

- ensure that the module kernel clock is enabled for the access to MCx\_ECCD and MCx\_ETRRi,
- and perform a dummy write to MCx\_ECCD with value 780F<sub>H</sub> before any read/write access to MCx\_ECCD or MCx\_ETRRi.

*Note: The module kernel clock (of the module in which the SRAM is present) does not need to be enabled if it can be ensured that no concurrent SPB bus accesses by other masters (CPU, DMA, HSM, debugger, ..) to other modules are performed during the MCx\_ECCD/ETRRi access while the module kernel clock is disabled.*

The module kernel clock is enabled under the following conditions:

1. For CPU memories, the clock is enabled after reset (for CPU<sub>x</sub> with  $x > 0$  even when CPU<sub>x</sub> is still in BOOT-HALT mode), when the CPU is not explicitly put into IDLE mode by software.
2. For SRAMs in peripherals, the module kernel clock is enabled when the module clock is enabled via the CLC register.

The value 780F<sub>H</sub> has been chosen as an example based on the following use cases and assumptions:

- If error reporting is turned on (i.e. notification enable bits \*ENE are set), it does not disturb the system to write back 780F<sub>H</sub> to register ECCD (write back of reset values, write to read-only bits and write of 1<sub>B</sub> to error indication bits has no effect).
- If error reporting is turned off (i.e. notification enable bits \*ENE are cleared), write back of 780F<sub>H</sub> to register ECCD may trigger SMU alarms (if SMU is configured). It is assumed that the corresponding errors are already known by the system since error reporting had previously been deactivated.

### **MTU\_TC.007 Error Overflow Indication ECCD.EOV**

The Error Overflow Indication bit EOv in register ECCD does not work correctly in specific cases for the following modules:

E-Ray, ETH, GTM, CIF (in ADAS devices), MCDS (in emulation devices), DAM.

The problem occurs in the following cases:

- If an error (correctable, uncorrectable, address error) was detected at address 0, this error is correctly stored in the ETRR register (ETRR.ADDR = 0x0, flags SERR, CERR, UERR are set accordingly), and bit ECCD.5 (least significant bit of bit field VAL) is set to 1<sub>B</sub>.
  - However, a subsequent error on a different address doesn't generate an error overflow, i.e. bit ECCD.EOV isn't set to 1<sub>B</sub>.
- If an error (correctable, uncorrectable, address error) was detected at an address  $\text{addr}_x > 0$ , this error is correctly stored in the ETRR register (ETRR.ADDR =  $\text{addr}_x$ , flags SERR, CERR, UERR are set accordingly), and bit ECCD.5 (least significant bit of bit field VAL) is set to 1<sub>B</sub>.

- However, a subsequent read from same `addr_x` which is still erroneous will erroneously generate an error overflow, i.e. bit `ECCD.EOV` is set to  $1_B$ .

### Workaround

Test address `0x0` by software to identify whether the device is sensitive to the first effect described above.

Periodically check bit field `VAL` in register `ECCD`. If `VAL`  $\neq 0$ , save the contents stored in `ETRR` (`ADDR` and `MBI`), and clear `ECCD` and `ETRR` afterwards (via `ECCD.TRC = 1_B`). If an error overflow is signaled, compare the current contents of `ETRR` with the saved value to identify an unmotivated error overflow (second effect described above).

### **MTU\_TC.011 MBIST Bitmap not working for w0 - r1**

The simple test case of writing all 0 and checking for 1 should return a full bitmap.

However, in this device step, only one (the last) address of the SRAM is returned.

### Workaround

Use the reverse test `w1 - r0`, which is working as expected and returns the full bitmap.

### **MTU\_TC.012 Security of CPU Cache Memories During Runtime is Limited**

MTU chapter “Security Applications” in the User’s Manual describes that selected memories with potentially security relevant content are initialized under certain conditions to prevent reading of their data or supplying manipulated data.

The description is correct, but the initialization of CPU cache and cache tag memories triggered by MBIST enable/disable and when mapping/un-mapping

these memories to/from system address space using MEMMAP register is of limited value:

- These memories stay functional as cache in the address mapped state. Therefore software can enable address mapping and afterwards watch cache usage of the application (this is a debug feature). Even manipulation of the cache content is feasible.
- It is possible to abort an ongoing memory initialization.

The security of memory initialization during startup is not affected. Also protection of FSI0 and HSM memories is not limited.

### Workaround

Handle security relevant data exclusively inside HSM. Protect the application code by locking external access (e.g. lock debug interface, prevent boot via serial interface). Consider validation of application code by HSM secure boot.

### **MultiCAN\_TC.043 CAN FD: Idle Condition**

*Note: This problem does not affect the nodes in TC23xED and ADAS devices. In TC23x step AC, this problem only affects node 2 of modules MultiCAN and MultiCAN1.*

The CAN FD ISO draft hardens the idle condition for the integration phase. Now it is also required that the 11 consecutive recessive bits occur without any synchronization in between. This is to ensure that integration of a node during a fast CAN FD data phase is not irritated by the fast baudrate.

### Problem Description

If a fast CAN FD data phase is overlapping with the integration phase, the CAN module might end the integration phase too early.

As a result, the node is not properly synchronized and might produce an error on the bus.

### Workaround

None

## **MultiCAN\_TC.044 CAN FD: Missing Hardsync**

*Note: This problem does not affect the nodes in TC23xED and ADAS devices. In TC23x step AC, this problem only affects node 2 of modules MultiCAN and MultiCAN1.*

The CAN FD specification requires hard synchronization from the transition recessive FDF bit to dominant res bit being enabled. Hard-synchronization insures perfect synchronization, even for large timing offsets, due to arbitration loss.

### **Current Implementation**

Soft synchronization is implemented at FDF bit, no hard synchronization as required.

### **Problem Description**

Soft-synchronization is enabled for the transition recessive FDF bit to dominant res bit. In case of a time offset higher than the SJW (Synchronization Jump Width), only partial synchronization is achieved. This might not be sufficient for high speed CAN FD data phase.

### **Incorrect Behavior**

If a transmitter is losing arbitration late (e.g. in an extended frame - see [Figure 4](#) the node which wins arbitration and [Figure 5](#) the transmitting node losing arbitration), then the FDF bit to res bit is the last chance to synchronize to the winning node (here as a receiver [Figure 6](#) the same node as in [Figure 5](#)). A timing offset will occur between the now receiving node and the actual transmitter.

### **Consequence**

The device is producing more error frames within a network than a device with a hardsync at the FDF bit.

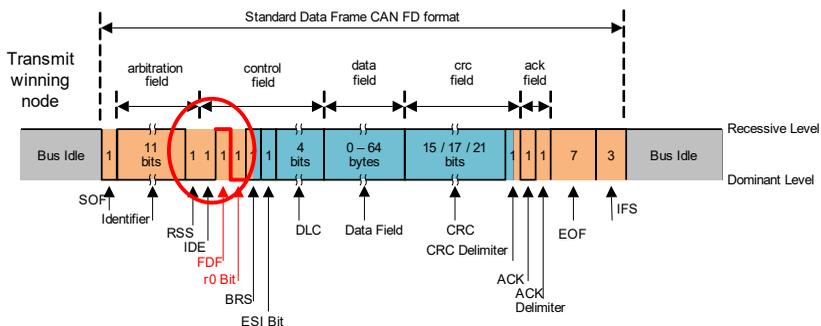


Figure 4 Transmitter - winning CAN FD node

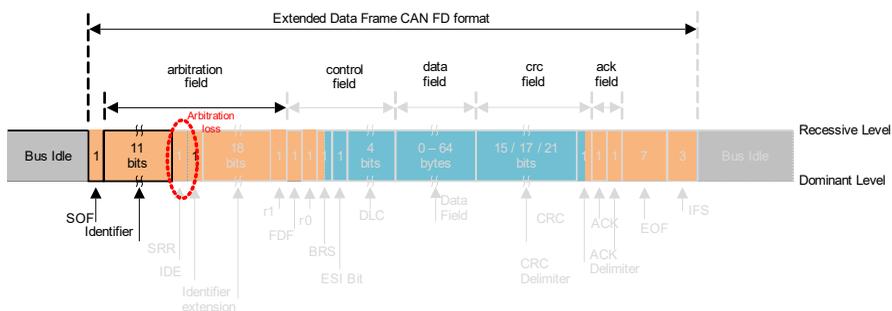


Figure 5 Transmitter - losing CAN FD node

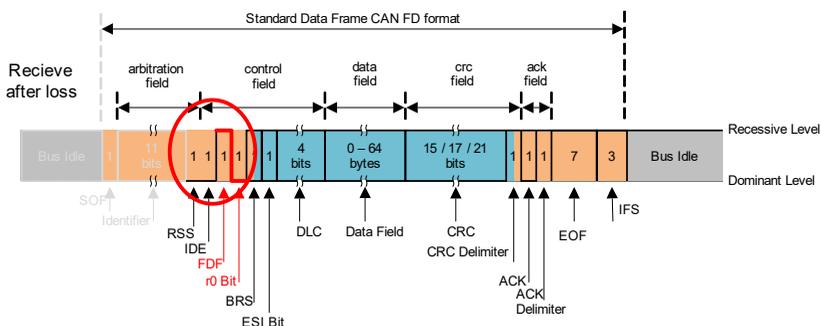


Figure 6 Now receiver (same node as previous figure) - losing CAN FD node

## Workaround

This workaround is not 100% solving the “Missing Hardsync“ issue, but setting the SJW to the highest possible value will hide the problem in most cases.

### **MultiCAN\_TC.045 Simultaneous communication of ISO 11898-1:2015 CAN FD and Non-ISO CAN FD nodes**

When ISO CAN FD communication is enabled for CAN node 0, then ongoing CAN protocol activities of a Non-ISO CAN FD supporting node disturb the communication of CAN node 0 (see table at the end of MultiCAN\_TC.H008 for a summary of nodes supporting ISO/Non-ISO CAN FD).

- For **TC23x**, this means:
  - When ISO CAN FD communication is enabled for CAN node 0, then ongoing CAN protocol activities of CAN **node 2** disturb communication of CAN node 0.
- For **TC26x**, this means:
  - When ISO CAN FD communication is enabled for CAN node 0, then ongoing CAN protocol activities of CAN **node 4** disturb communication of CAN node 0.

Thus, no overlapping communication of ISO CAN FD in node 0 and Classical CAN in a Non-ISO CAN FD supporting node is possible.

*Note: Other ISO CAN FD supporting nodes (node 1 in TC23x, node 1..3 in TC26x) are not impacted by this issue.*

## Workarounds

### Option 1:

Do not use the Non-ISO CAN FD supporting node of the device.

- For **TC23x**: disable CAN **node 2**.
- For **TC26x**: disable CAN **node 4**.

### Option 2:

Do not use CAN node 0 in the CAN FD mode. Use CAN node 0 only in Classical CAN mode.

**OCDS\_TC.038 Disconnecting a debugger without device reset (“hot detach”) may require reading of OCS registers**

If a debugger disconnects, it should activate at least the Debug Reset. This will reset all the main OCDS resources like CPUs, Cerberus, etc. However for peripherals having a BPI interface, there is the following issue: The Debug Reset is implemented as a synchronous clear on this level. If the OCDS registers are not clocked (e.g. for power saving reasons), the effect of this synchronous clear will be delayed to the next activation of the clock.

In general this will be more a theoretical problem. It's very unlikely that there is a use case, where a hot detach is required and critical OCDS resources of peripherals were used before. In nearly all cases this effect is invisible for a user, since any register access of the peripheral will generate the clock cycles which are required for the synchronous clear.

**Workaround**

In case of a hot detach, a tool should - after the Debug Reset activation - read the OCS registers of all peripherals where it used critical OCDS resources. These reads will initiate the required peripheral kernel clocks for the synchronous clear of the OCDS resources.

**OCDS\_TC.042 OTGS capture registers can miss single clock cycle triggers**

The Cerberus OTGS capture registers (TCTL, TCCB, TCCH, TCIP, TCTGB, TCM) can fail to capture a trigger if the trigger is of single clock cycle duration and arrives in the same cycle as the same trigger register is being read by the bus.

**Workaround**

Avoid polling of OTGS capture registers while the system is running.

If polling while running can't be avoided use TLCCx counters for capturing critical Trigger Lines.

### **OCDS TC.043 Read-Modify-Write Bus Transactions to Cerberus Registers**

During read-modify-write (RMW) bus transactions to writable registers in the Cerberus (CBS), the target register is incorrectly updated with an undefined value during the Read-part. The correct value is always returned to the bus master for the Read-part, and the correct value is written to the register when the Write-part completes. But the register may contain an undefined value for a number of clock cycles between the Read-part and the Write-part.

The bus master (CPU) will see the RMW complete normally, but any logic driven by the hardware register's writable bits may be unexpectedly toggled.

This effects all registers that can be written by the SPB (using the FPI protocol) in the CBS block. It does not effect external access from the tool via JTAG/DAP.

#### **Workaround**

Do not use RMW bus operations targeting the CBS registers.

### **PINNING TC.001 Port functions and pinning tables for TC234\* devices without ETH module in TQFP-144 package - Documentation update**

The port function and pinning tables included in chapter “PG-TQFP-144-27 Package Variant Pin Configuration of TC23x-ADAS” of the current version of the Data Sheet correctly only apply to the TC234\* variants that include an ETH module. They are identified by the sales names

- TC234LA-\* and TC234LX-\*

and are only available for TC23x design step AB. In addition, these devices include 4 VADC kernels (G0..G3) and EMEM.

Other variants of TC23x design step AB do not have an ETH and EMEM module, and only have 2 VADC kernels (G0 and G1).

TC23x design step AC does not have an ETH and EMEM module, and only has 2 VADC kernels (G0 and G1).

#### **Port functions and pinning tables for TC234\* devices without ETH module**

**in TQFP-144 package - Documentation update**

For these devices, the following modifications must be considered in the respective rows and columns when using the port function and pinning tables in chapter “PG-TQFP-144-27 Package Variant Pin Configuration of TC23x-ADAS”:

**Table 7 Modifications to port function and pinning tables for TC23x without ETH module in TQFP-144 package**

Pin	Symbol	Buffer Type	Function
	ETH0_*		All ETH0_* inputs and outputs are not existing (device does not contain an ETH module)
	P10.1	A1 pad (instead of A1+)	
	P10.2	A1 pad (instead of A1+)	
	P10.3	A1 pad (instead of A1+)	
	P13.2	A1 pad (instead of A1+)	
	P13.3	A1 pad (instead of A1+)	
	P15.2	A1 pad (instead of A1+)	
	P15.3	A1 pad (instead of A1+)	
	P15.4	A1 pad (instead of A1+)	
	P40.1		Input VADCG2_0 is not existing (device does not contain VADC group G2)

**Table 7 Modifications to port function and pinning tables for TC23x without ETH module in TQFP-144 package (cont'd)**

Pin	Symbol	Buffer Type	Function
	P41.1		Input VADCG3_0 is not existing (device does not contain VADC group G3)
10	VDD		Pin 10 is only used as VDD but not as VDDSB because EMEM is not available

**PLL\_ERAY\_TC.001 PLL\_ERAY Initialization after Cold Power-up or Wake-up from Standby mode**

When the PLL\_ERAY is configured by the application software after cold power-on reset or wake-up from Standby mode, it may not always reach the intended target frequency (either lock at a lower frequency, or go into unlock state), in particular at high temperature.

**Workaround**

The following code sequence, executed after power-on reset or wake-up from Standby mode and before initializing the PLL\_ERAY, avoids the problem:

```

SCU_PLLELAYCON0.B.PLLPWD = 0; // set PLL_ERAY to power
                               // saving mode
wait(10);                      // wait 10µs
SCU_PLLELAYCON0.B.PLLPWD = 1; // set PLL_ERAY to normal
                               // behavior
...                             // initialize PLL_ERAY

```

**PLL\_TC.005 PLL Initialization after Cold Power-up or Wake-up from Standby mode**

When the system PLL is configured by the application software after cold power-on reset or wake-up from Standby mode, it may not always reach the

intended target frequency (either lock at a lower frequency, or go into unlock state), in particular at high temperature.

### Workaround

The following code sequence, executed after power-on reset or wake-up from Standby mode and before initializing the system PLL, avoids the problem:

```
SCU_CCUCON0.B.CLKSEL = 0; // switch system clock to
    another source different from PLL, e.g. back-up clock
SCU_CCUCON0.B.UP = 1; // request update
while(SCU_CCUCON0.B.LCK == 1); // wait for update handshake
    (see separate Note below)
SCU_PLLCON0.B.PLLPWD = 0; // set PLL to power saving mode
wait(10); // wait 10µs
SCU_PLLCON0.B.PLLPWD = 1; // set PLL to normal behavior
... // initialize PLL
```

*Note: For devices with PLL\_ERAY, see also problem PLL\_ERAY\_TC.001*

### Note on update handshake

LCK = 0 indicates the end of the update handshake. Instead of polling LCK, other instructions may be executed that bridge this time.

The minimum number of instruction cycles  $n_{UH}$  (cycles of  $f_{SRI}$ ) required to bridge the maximum time for the update handshake depends on the least common multiple of the active clock divider factors  $> 0$  that are effective in CCUCON0/1/2/5 before the update by CCUCON0.B.UP = 1 in the sequence above is requested. For LPDIV = 0, this set includes FSIDIV, FSI2DIV, SPBDIV, SRIDIV, BAUD2DIV, BAUD1DIV, ASCLINSDIV, ASCLINFDIV, GTMDIV, STMDIV, CANDIV, MAXDIV, BBDIV.

This results in the following range when SRIDIV = 1 ( $f_{SRI} = f_{SOURCE}$ ):

- $n_{UH} \geq 17$  if the active clock divider factors are any of the elements of the set {1, 2, 3, 4, 6, 12}.
- $n_{UH} \geq 29$  if factor **8** is included in the set of {1, 2, 3, 4, 6, **8**, 12}.
- $n_{UH} \geq 65$  if factor **5** (or multiples of 5) are included in the set of {1, 2, 3, 4, **5**, 6, **10**, 12, **15**}.

- $n_{UH} \geq 125$  if factors **5** (or multiples of 5) and **8** are included in the set of {1, 2, 3, 4, **5**, 6, **8**, **10**, 12, **15**}.

When  $SRDIV = n > 1$ , only  $n_{UH}/n$  instruction cycles are required to bridge the maximum time for the update handshake, as the instructions take  $n$  times longer.

For  $LPDIV > 0$ , the divider factors for  $f_{SRI}$ ,  $f_{SPB}$ ,  $f_{BBB}$  and  $f_{MAX}$  are determined by  $LPDIV$ . As instruction execution is slowed down by the ratio defined by  $LPDIV$ , the number of instructions to bridge the time for the update handshake is scaled accordingly.

*Note: For the allowed clock ratios see table “CCU allowed Clock Ratios” in the User’s Manual.*

### **PLL\_TC.007 PLL Loss of lock when oscillator shaper is used**

Under certain conditions the PLL loses lock when the oscillator shaper is used ( $OSCCON.SHBY = 0_B$ , recommended system configuration, default after reset).

The fail behavior is not observed for oscillator frequencies  $f_{OSC} \leq 25$  MHz when using an external crystal / ceramic resonator or supplying the clock signal directly.

### **Workaround**

It is recommended to use input clock frequencies  $f_{OSC} \leq 25$  MHz.

*Note: For devices with PLL\_ERAY, the problem also applies to PLL\_ERAY.*

### **PMC\_TC.002 Switch Capacitor Regulator Mode, Frequency Spreading - Documentation Update to Register EVRSDCTRL1**

The documentation of bit fields  $SDFREQSPRD$ ,  $TON$ ,  $TOFF$  in register  $EVRSDCTRL1$  will be updated with the next revision of the User’s Manual.

## 1. Documentation Update to SDFREQSPRD

The correct encoding of bit field SDFREQSPRD (Frequency Spread Mode) in register EVRSDCTRL1 is as documented in [Table 8](#):

**Table 8 EVR13 SD Control Register 1 (EVRSDCTRL1) - Frequency Spread Mode Encoding**

Field	Bits	Type	Description
<b>SDFREQSPRD</b>	[3:0]	rw	<p><b>Frequency Spread Mode</b></p> <p>This bit field defines the maximum number of back-up clock cycles (<math>f_{BACK}</math>) which are added to both charge (TON) and discharge (TOFF) switching phases thus increasing the average switching period. The number of clock cycles added are randomized equally over the listed cycle count range. The resulting TON and TOFF phase lengths are the same and 50% duty cycle is maintained.</p> <p>0<sub>H</sub> No frequency spreading activated.            1<sub>H</sub> 0 to 3 clock cycles are added (default).            2<sub>H</sub> 0 to 7 clock cycles are added.            All other values are reserved.</p>

## 2. Documentation Update to TON/TOFF

The effect of bit field EVRSDCTRL1.SDFREQSPRD on the SC DCDC switching period is as documented in [Table 9](#):

**Table 9 EVR13 SD Control Register 1 (EVRSDCTRL1) - Switching Period**

Field	Bits	Type	Description
<b>TON</b>	[15:8]	rw	<b>Charge Phase Length</b> The charge phase length is defined in back-up clock cycles ( $f_{BACK}$ , nominal 100 MHz): <ul style="list-style-type: none"> <li>• In case SDFREQSPRD = 0:               <ul style="list-style-type: none"> <li>– Switching period (in cycles) = TON + TOFF + 18.</li> </ul> </li> <li>• In case SDFREQSPRD <math>\neq</math> 0:               <ul style="list-style-type: none"> <li>– Switching period (in cycles) = TON + TOFF + 18 + 2 * <math>[2^{(SDFREQSPRD + 1)} - 1]</math>.</li> </ul> </li> </ul>
<b>TOFF</b>	[23:16]	rw	<b>Discharge Phase Length</b> The discharge phase length in clock cycles should be the same as the charge phase length.

**PMC\_TC\_003 Usecase limitation of LDO mode with on chip pass device for SAL devices**

*Note: This problem only affects devices with temperature range classification "SAL" (ambient temperature range -40/150°C). For devices classified as "SAK" (-40/125°C),  $T_J$  is limited to 150°C anyway (see Data Sheet).*

The LDO mode of the internal EVR13 regulator with on-chip pass device shall not be used at condition  $T_J > 150^\circ\text{C}$ .

Junction temperature  $T_J > 150^\circ\text{C}$  leads to degradation of the pass device characteristics resulting in a deviation of the static accuracy ( $V_{OUTT}$ ) of the EVR13 core regulator.

If the junction temperature is increased above  $150^\circ\text{C}$  in an application with the on-chip pass device, deviation of the static accuracy down to -3% of the nominal voltage is possible for  $I_{DD}$  load currents above 150 mA at the respective device junction temperature. The application shall be tolerable to such deviation of the output voltage accuracy considering limits for the dynamic regulation of the EVRC.

**QSPI TC.006 Baud rate error detection in slave mode (error indication in current frame)**

According to the specification, a baud rate error is detected if the incoming shift clock supplied by the master has less than half or more than double the expected baud rate (determined by bit field GLOBALCON.TQ).

However, in this design step, a baud rate error is detected not only if the incoming shift clock has less than half the expected baud rate (as specified), but also already when the incoming shift clock is somewhat (i.e. less than double) higher than the expected baud rate.

In this case, the baud rate error is indicated in the current frame.

**Workaround**

It is recommended not to rely on the baud rate error detection feature, and not to use the corresponding automatic reset enable feature (i.e. keep GLOBALCON.AREN=0<sub>B</sub>).

The baud rate error detection feature in slave mode is of conceptually limited use and is not related to data integrity. Data integrity can be ensured e.g. by parity, CRC, etc., while clocking problems of an AURIX™ master are detected by mechanisms implemented in the master.

Protection against the effects of high frequency glitches is provided by the spike detection feature in slave mode.

**QSPI TC.017 Slave: Reset when receiving an unexpected number of bits**

A deactivation of the slave select input (SLSI) by a master is expected to automatically reset the bit counter of the QSPI module when configured as a slave.

This reset should help slaves to recover from messages where faults in the master or glitches on SCLK lead to an incorrect number of clocks on SCLK (= incorrect number of bits per SPI frame).

However, in this design step, the reset of the bit counter is unreliable.

### Workaround

The slave should enable the Phase Transition interrupt (PT2EN = 1<sub>B</sub> in register GLOBALCON1) to be triggered after the PT2 event “SLSI deselection” (PT2 = 101<sub>B</sub>).

In the interrupt service routine, after ensuring that the receive data has been copied, the software should issue a reset of the bit counter and the state machine via GLOBALCON.RESETS = 0111<sub>B</sub>.

### **RESET\_TC.005 Indication of Power Fail Events in SCU\_RSTSTAT**

In case of consecutive cold resets triggered by EVR13, EVR33 or SWD power fail events, then only the last power fail event is registered in register SCU\_RSTSTAT. It is not possible to distinguish individually between EVR13, EVR33 or SWD power fail events from RSTSTAT information.

### Workaround

In case any power fail reset indication bit is set among EVR13, EVR33 or SWD power fail events in register SCU\_RSTSTAT, it has to be assumed that all power fail events may have happened before.

### **SCU\_TC.034 TESTMODE pin shall be held at static high level during LBIST**

For a stable MISR signature, the level on pin TESTMODE (P20.2) must not change during LBIST execution.

Therefore, always keep pin TESTMODE (P20.2) at a static **high** level during LBIST execution.

### **SMU\_TC.006 OCDS Trigger Bus OTGB during Application Reset**

The SMU provides an alarm trigger and trace interface (Trigger Set TS16\_SMU) using the OCDS Trigger Bus OTGB.

While the Application Reset is active, the SMU outputs the reset state of the OTGB interface instead of TS16\_SMU.

This OTGB interface reset state is identical to TS16\_SMU when no alarm is active.

After the Application Reset TS16\_SMU is output again.

### Workaround

Just ignore the phase in the OTGB trace where an alarm seems to become inactive while the Application Reset is active.

### **SMU\_TC.007 Size and Position of Field ACNT in Register SMU\_AFCNT**

*Note: This erratum might affect the SFR C Header Definitions. In such cases, SFR usage in the software shall be analyzed within the applications for their correct handling.*

In the SMU chapter of the User's Manual, in the description of register SMU\_AFCNT (Alarm and Fault Counter),

- Size and position of field ACNT (Alarm Counter) are incorrectly described as SMU\_AFCNT.[15:8], and
- Bits SMU\_AFCNT.[7:4] are incorrectly shown as "Reserved; read as 0".

The **correct** size and position of field ACNT (Alarm Counter) in register SMU\_AFCNT is SMU\_AFCNT.[15:4], as shown in the following **Table 10**.

The position of the "Reserved" bits is aligned accordingly.

**Table 10 Field ACNT in Register SMU\_AFCNT - Correction**

Field	Bits	Type	Description
ACNT	[15:4]	rh	<b>Alarm Counter</b> This field is incremented by hardware when the SMU processes an <b>internal</b> action related to an alarm event (see Figure " <b>Alarm operation</b> "). The counter value holds if the maximum value is reached.

**Table 10 Field ACNT in Register SMU\_AFCNT - Correction (cont'd)**

Field	Bits	Type	Description
0	[29:16]	r	<b>Reserved</b> Read as 0; should be written with 0.

*Note: The other fields (ACO, FCO, FCNT) of register SMU\_AFCNT are correctly described in the User's Manual.*

### **SMU\_TC.008 Behavior of Action Counter ACNT**

Register SMU\_AFCNT (Alarm and Fault Counter) implements a Fault Counter (FCNT) that counts the number of transitions from the RUN state to the FAULT state. Register AFCNT is only reset by a power-on-reset.

Whenever a pending alarm event is processed, the corresponding status bit is set to 1<sub>B</sub> by hardware in the Alarm Status register AG<x>.

If an internal SMU action is configured for this alarm, the Action Counter (ACNT) in register AFCNT is incremented anytime the SMU processes this internal action.

#### **Corner Case**

In this device step, some of the alarm signals may increment the Action Counter ACNT multiple times for a single alarm event.

#### **Workaround**

Do not rely on the value in the action counter ACNT.

### **SMU\_TC.010 Transfer to SMU\_AD register not triggered correctly**

#### **Background**

The SMU contains Alarm Debug registers which can be used for diagnostic purposes. If an alarm which is configured to generate a reset (application or system reset) is sent to the SMU, a copy of the Alarm Status registers – AGi – into the Alarm Debug registers – ADi – is automatically triggered.

The AGi are reset by Application reset while the ADi are reset only by power-on reset.

### Corner Case

In the case that a first SMU alarm AGi[j] generates a reset request, and a second alarm AGx[y] (where x=i and y=j is possible) configured for a reset occurs a few cycles before the reset is actually executed, then the reset values of the AGi registers will be transferred to the ADi register.

In this case, the ADi registers will not reflect the root cause that lead to a SMU alarm/reset.

*Note: This corner case will always be met for level alarms.*

### **SMU\_TC.012 Unexpected alarms when registers FSP or RTC are written**

Due to a synchronization issue, ALM3[27] is sporadically triggered if the PRE2 field of register FSP is written while the SMU is configured in Time Switching protocol (FSP.MODE = 10<sub>B</sub>) and FSP[0] is toggling with a defined T<sub>SMU\_FFS</sub> period.

Also, ALM3[27] is sporadically triggered if the PRE1 or TFSP\_HIGH fields of register FSP are written while the SMU is in the Fault State and T<sub>FSP\_FS</sub> has not yet been reached (STS.FSTS=0<sub>B</sub>) (regardless of the FSP.MODE configuration).

In addition, an unexpected ALM2[29] or ALM2[30] is sporadically triggered if field FSP.PRE1 or RTC.RTD is written, and at least one recovery timer is running based on a defined T<sub>SMU\_FS</sub> period (regardless of the FSP.MODE configuration).

The alarms can only be cleared with cold or warm Power-On reset.

### Workaround

To avoid unexpected alarms, perform the configuration of the PRE1, PRE2 or TFSP\_HIGH fields only when the SMU is not in the Fault State and FSP is in Bi-stable protocol mode (FSP.MODE = 00<sub>B</sub>). Mode switching and configuration shall not be done with the same write access to register FSP.

This means that in the Fault Free State:

- before writing to PRE1, PRE2 or TFSP\_HIGH while Time Switching protocol is enabled:
  - disable Time Switching protocol by setting FSP in Bi-stable protocol mode (FSP.MODE = 00<sub>B</sub>);
  - wait until Bi-stable protocol mode is active (read back register FSP twice);
  - write desired value to PRE1, PRE2 or TFSP\_HIGH;
  - then switch FSP.MODE to the desired protocol (optional step).
- If the mode shall be changed after writing to PRE1, PRE2 or TFSP\_HIGH while in Bi-Stable protocol mode (FSP.MODE = 00<sub>B</sub>):
  - write desired value to PRE1, PRE2 or TFSP\_HIGH;
  - then switch FSP.MODE to Time Switching protocol.

If field FSP.PRE1 or RTC.RTD shall be written, make sure no recovery timer is running. It is not allowed to write to the PRE1 or RTD field when at least one recovery timer is running (indicated by bits RTS0 and RTS1 in the STS register).

### **SRI TC.003 XBAR\_PRIOL/H Register Layout and Reset Values**

*Note: This erratum might affect the SFR C Header Definitions. In such cases, SFR usage in the software shall be analyzed within the applications for their correct handling.*

The CPU0 SRI masters (CPU0.DMI, CPU0.PMI) are mapped to the XBar\_SRI Master Connection Interfaces MCI12 and MCI13 as described in table “Mapping of TC21x/TC22x/TC23x SRI master devices to MCI” of the User’s Manual.

*Note: This implementation in the TC23x .. TC21x devices is compatible with the other devices (TC29x .. TC26x) of the AURIX™ family.*

However, the description of the register layout and reset values for the XBAR\_PRIOL/H registers in chapter “TC21x/TC22x/TC23x Control Registers” of the TC21x/TC22x/TC23x Family User’s Manual V1.1 is partially incorrect.

The corrected parts of the description are shown in the following tables.

**Table 11 XBAR\_PRIOL Registers - Reset Values TC23x**

Short Name	Description	Reset Value
XBAR_PRIOLD	Arbiter Priority Register D	0020 0002 <sub>H</sub>
XBAR_PRIOL0	Arbiter Priority Register 0	0020 0002 <sub>H</sub>
XBAR_PRIOL4	Arbiter Priority Register 4	0020 0002 <sub>H</sub>
XBAR_PRIOLx (x = 6-7)	Arbiter Priority Register x	0020 0002 <sub>H</sub>

**Table 12 XBAR\_PRIOL Registers - Fields TC23x**

Field	Bits	Type	Description
<b>MASTER0</b>	[2:0]	rw	<b>Master 0 Priority</b> (Priority of DMA Access)
<b>MASTER5</b>	[22:20]	rw	<b>Master 5 Priority</b> (Priority of SFI Access)
<b>0</b>	[31:23], [19:3]	r	<b>Reserved</b> Read as 0; should be written with 0.

**Table 13 XBAR\_PRIOH Registers - Reset Values TC23x .. TC21x**

Short Name	Description	Reset Value
XBAR_PRIOHD	Arbiter Priority Register D	0055 0000 <sub>H</sub>
XBAR_PRIOH0	Arbiter Priority Register 0	0055 0000 <sub>H</sub>
XBAR_PRIOH4	Arbiter Priority Register 4	0055 0000 <sub>H</sub>
XBAR_PRIOHx (x = 6-7)	Arbiter Priority Register x	0055 0000 <sub>H</sub>

**Table 14 XBAR\_PRI0H Registers - Fields TC23x .. TC21x**

<b>Field</b>	<b>Bits</b>	<b>Type</b>	<b>Description</b>
<b>MASTER12</b>	[18:16]	rw	<b>Master 12 Priority</b> (Priority of CPU0.DMI Access)
<b>MASTER13</b>	[22:20]	rw	<b>Master 13 Priority</b> (Priority of CPU0.PMI Access)
<b>0</b>	[31:23], 19, [15:0]	r	<b>Reserved</b> Read as 0; should be written with 0.

### 3 **Deviations from Electrical- and Timing Specification**

#### **ADC\_TC.P010 Increased Gain Error ( $EA_{GAIN}$ ) for $T_J < 0^\circ\text{C}$**

For devices with Analog-Digital-Converters (VADC) providing 16:1 analog multiplexers (TC26x, TC23x..TC21x), the maximum Gain Error ( $EA_{GAIN}$ ) increases as follows for  $T_J < 0^\circ\text{C}$ :

- from  $\pm 3.5 \text{ LSB}_{12}$  to  $\pm 4.5 \text{ LSB}_{12}$  when  $V_{DDM} = 4.5 \text{ V}$  to  $5.5 \text{ V}$  (upper voltage range) and sample time  $t_s < 200 \text{ ns}$ ,
- from  $\pm 5.5 \text{ LSB}_{12}$  to  $\pm 6.5 \text{ LSB}_{12}$  when  $V_{DDM} = 2.97 \text{ V}$  to  $4.5 \text{ V}$  (lower voltage range) and sample time  $t_s < 400 \text{ ns}$ .

**Note:**

1. The resulting Total Unadjusted Error (TUE) is not affected and remains as specified in the corresponding Data Sheet.
2. For temperatures  $T_J \geq 0^\circ\text{C}$ , the Gain Error ( $EA_{GAIN}$ ) remains as specified in the corresponding Data Sheet.
3. For  $t_s \geq 200 \text{ ns}$  (upper voltage range) or  $t_s \geq 400 \text{ ns}$  (lower voltage range), the Gain Error ( $EA_{GAIN}$ ) remains as specified in the corresponding Data Sheet.

#### **IDD\_TC.H001 IPC Limits used in Production Test for IDD Max Power Pattern**

Instructions per cycle for a CPU is measured by dividing ICNT instruction counter value with the CCNT clock counter value.

*Note: For a complete description of registers ICNT and CCNT refer to the TriCore Architecture Manual, chapter "Performance Counter Registers".*

Parameters using the max power pattern for device individual testing of power consumption limits (IDD) are tested for a maximum IPC rate of 1.3 for all CPUs available in the device.

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**Deviations from Electrical- and Timing Specification****IEVRSB\_TC.P001 Test Condition for  $I_{EVRSB}$  (sum of all currents in standby mode) - Data Sheet correction**

In table “Power Supply” in the current version of the Data Sheet, in column “Note/Test Condition” for parameter “Sum of all currents (STANDBY mode)” (symbol  $I_{EVRSB}$ ), the following term is **incorrect**:

- $V_{EVRSB} = 5\text{ V}$

**Correction**

The correct value for  $V_{EVRSB}$  in the test condition for  $I_{EVRSB}$  shall be

- $V_{EVRSB} = 3.3\text{ V}$

**PADS\_TC.H004 PN-Junction Characteristics for Pad Type S**

As described in chapter “Package and Pinning Definitions” in the Data Sheet, symbol “S” in column “Type” is defined as class D ADC input with digital input. Consequently, for pad type S, the PN-junction characteristics for pad type D apply.

The corresponding values for  $U_{IN}$  are listed in tables “PN-Junction Characteristics for positive Overload” and “PN-Junction Characteristics for negative Overload” in chapter “Pin Reliability in Overload” in the Data Sheet.

**RTH\_TC.H001 Thermal characteristics of the package - Footnote update for LF-BGA-292-6 package**

The references to the JEDEC standards (JESD51-3/5/7) for RQJA in the footnote for the LF-BGA-292-6 package in table “Thermal characteristics of the package” are not correct. They only apply to the TQFP package.

**Correction**

The correct footnote for the LF-BGA-292-6 package is:

- <sup>3)</sup> Value is defined in accordance with JESD51-9.

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**Deviations from Electrical- and Timing Specification****VDDPPA TC.H001 Voltage to ensure defined pad states - Footnote update**

In the footnote for parameter “Voltage to ensure defined pad states” (symbol  $V_{DDPPA}$ ) in table “Operating Conditions” of the Data Sheet,  $V_{DDP3}$  is mentioned as representative for “non-core supply voltages” in the text.

**Update**

The footnote for  $V_{DDPPA}$  should be extended to include all “non-core supply voltages” as follows:

\*) This parameter is valid under the assumption the PORST signal is constantly at low level during the power-up/power-down of the “non-core supply voltages” ( $V_{DDP3}$ ,  $V_{EXT}$ ,  $V_{FLEX}$ ,  $V_{DDFL3}$ ,  $V_{DDM}$ , ..., depending on the respective TC2x device version).

## 4 Application Hints

### **ADC AI.H003 Injected conversion may be performed with sample time of aborted conversion**

For specific timing conditions and configuration parameters, a higher prioritized conversion  $c_i$  (including a synchronized request from another ADC kernel) in cancel-inject-repeat mode may erroneously be performed with the sample time parameters of the lower prioritized cancelled conversion  $c_c$ . This can lead to wrong sample results (depending on the source impedance), and may also shift the starting point of following conversions.

The conditions for this behavior are as follows (all 3 conditions must be met):

1. **Sample Time setting:** injected conversion  $c_i$  and cancelled conversion  $c_c$  use different sample time settings, i.e. bit fields  $STC^*$  in the corresponding Input Class Registers for  $c_c$  and for  $c_i$  ( $GxICLASS0/1$ ,  $GLOBICLASS0/1$ ) are programmed to different values.
2. **Timing condition:** conversion  $c_i$  starts during the first  $f_{ADCI}$  clock cycle of the sample phase of  $c_c$ .
3. **Configuration parameters:** the ratio between the analog clock  $f_{ADCI}$  and the arbiter speed is as follows:

$$N_A > N_D \cdot (N_{AR} + 3),$$

with

- a)  $N_A$  = ratio  $f_{ADC}/f_{ADCI}$  ( $N_A = 1 \dots 32$ , as defined in bit field  $DIVA$ ),
- b)  $N_D$  = ratio  $f_{ADC}/f_{ADCD}$  = number of  $f_{ADC}$  clock cycles per arbitration slot ( $N_D = 1 \dots 4$ , as defined in bit field  $DIVD$ ),
- c)  $N_{AR}$  = number of arbitration slots per arbitration round ( $N_{AR} = 4, 8, 16, \text{ or } 20$ , as defined in bit field  $GxARBCFG.ARBRRND$ ).

Bit fields  $DIVA$  and  $DIVD$  mentioned above are located in register  $GLOBCFG$ .

As can be seen from the formula above, a problem typically only occurs when the arbiter is running at maximum speed, and a divider  $N_A > 7$  is selected to obtain  $f_{ADCI}$ .

### Recommendation 1

Select the same sample time for injected conversions  $c_i$  and potentially cancelled conversions  $c_c$ , i.e. program all bit fields  $STC^*$  in the corresponding Input Class Registers for  $c_c$  and for  $c_i$  ( $GxICLASS0/1$ ,  $GLOBICLASS0/1$ ) to the same value.

### Recommendation 2

Select the parameters in register  $GLOBCFG$  and  $GxARBCFG$  according to the following relation:

$$N_A \leq N_D * (N_{AR} + 3).$$

### ADC\_TC.H011 Bit DCMSB in register GLOBCFG

The default setting for bit DCMSB (Double Clock for the MSB Conversion) in register  $GLOBCFG$  is  $0_B$ , i.e. one clock cycle for the MSB conversion step is selected.

$DCMSB = 1_B$  is reserved in future documentation and must not be used.

*Note: In devices supporting Workaround 4 of problem ADC\_TC.068,  $DCMSB = 1_B$  may be used to control synchronization of converter groups (for details, see ADC\_TC.068, Workaround 4).*

### ADC\_TC.H014 VADC Start-up Calibration

The formula for the duration of the start-up calibration in some versions of the TC2x User's Manuals is incorrect with respect to the used frequency, or missing.

In the following, the contents of chapter "Calibration" is reprinted, including the correct [Formula for Start-up Calibration](#) below.

#### Calibration

Calibration automatically compensates deviations caused by process, temperature, and voltage variations. This ensures precise results throughout the operation time.

An initial start-up calibration is required once after a reset for all converters. All converters must be enabled ( $ANONS = 11_B$ ). The start-up calibration is initiated globally by setting bit SUCAL in register GLOBCFG. Conversions may be started after the initial calibration sequence. This is indicated by bit  $CALS = 1_B$  AND bit  $CAL = 0_B$ .

### Formula for Start-up Calibration

The start-up calibration phase takes  $4352 f_{ADCI}$  cycles ( $4352 \times 50 \text{ ns} = 217.6 \mu\text{s}$  for  $f_{ADCI} = 20 \text{ MHz}$ ).

After that, postcalibration cycles will compensate the effects of drifting parameters. The postcalibration cycles can be disabled.

*Note: The ADC error depends on the temperature. Therefore, the calibration must be repeated periodically.*

### ADC\_TC.H015 Conversion Time with Broken Wire Detection

As described in a note in section “Broken Wire Detection” of the User’s Manual, the duration of the complete conversion is increased by the preparation phase (same as the sample phase) if the broken wire detection is enabled, i.e. the sample time doubles for standard conversions when broken wire detection is enabled ( $GxCHCTRY.BWDEN = 1_B$ ):

#### Formula for Standard Conversions without Broken Wire Detection

- $t_{CN} = t_s + (N + PC) \times t_{ADCI} + 2 \times t_{VADC}$  (see also User’s Manual/Data Sheet)

#### Formula for Standard Conversions with Broken Wire Detection

- $t_{CN} = 2 \times t_s + (N + PC) \times t_{ADCI} + 2 \times t_{VADC}$

where:

$$t_s = (2 + STC) \times t_{ADCI} \text{ for } STC \leq 15, \text{ and}$$

$$t_s = (2 + (STC-15) \times 16) \times t_{ADCI} \text{ for } STC \geq 16;$$

N = result width (8/10/12 bits);

PC = 2 if post-calibration selected, PC = 0 otherwise.

**Examples**

Conversion times for different configurations are shown in the following **Table 15** (without broken wire detection) and **Table 16** (with broken wire detection):

**Table 15 Conversion Time for Standard Conversions - Without Broken Wire Detection - Examples**

Result	Symbol	Time	Conditions
12-bit result	$t_{C12}$	$(16 + \text{STC}) \times t_{\text{ADCl}} + 2 \times t_{\text{VADC}}$	Post-calibration enabled, $\text{STC} \leq 15$
10-bit result	$t_{C10}$	$(12 + \text{STC}) \times t_{\text{ADCl}} + 2 \times t_{\text{VADC}}$	Post-calibration disabled, $\text{STC} \leq 15$
8-bit result	$t_{C8}$	$(10 + \text{STC}) \times t_{\text{ADCl}} + 2 \times t_{\text{VADC}}$	Post-calibration disabled, $\text{STC} \leq 15$

**Table 16 Conversion Time for Standard Conversions - With Broken Wire Detection - Examples**

Result	Symbol	Time	Conditions
12-bit result	$t_{C12B}$	$(18 + 2 \times \text{STC}) \times t_{\text{ADCl}} + 2 \times t_{\text{VADC}}$	Post-calibration enabled, $\text{STC} \leq 15$
10-bit result	$t_{C10B}$	$(14 + 2 \times \text{STC}) \times t_{\text{ADCl}} + 2 \times t_{\text{VADC}}$	Post-calibration disabled, $\text{STC} \leq 15$
8-bit result	$t_{C8B}$	$(12 + 2 \times \text{STC}) \times t_{\text{ADCl}} + 2 \times t_{\text{VADC}}$	Post-calibration disabled, $\text{STC} \leq 15$

**ADC\_TC.H020 Minimum/Maximum Detection Compares 12 Bits Only**

In minimum or maximum detection mode ( $\text{FEN} = 11_{\text{B}}$  or  $10_{\text{B}}$ ) new results are compared to the lower 12 bits of the respective result register bitfield RESULT.

Therefore, a value  $\text{RESULT} = \text{XFFF}_{\text{H}}$  ( $\text{X} > 0_{\text{H}}$ ) will not be updated for a new result value of  $0\text{FFF}_{\text{H}}$  in minimum detection mode.

In a real application, this should be no problem, as the minimum detection usually sees values below  $0FFF_H$ .

### Recommendation

For minimum detection, use the start value  $0FFF_H$  (instead of  $FFFF_H$  as mentioned in the User's Manual).

For maximum detection, use the start value  $0000_H$  as mentioned in the User's Manual.

### ADC\_TC.H022 Sample Time Control - Formula

Table "Sample Time Coding" in section "Input Class Registers" of the VADC chapter in the User's Manual describes the additional clock cycles (selected in bit fields STCS and STCE) to be added to the minimum sample time of two analog clock cycles.

As can be seen from the table in the User's Manual, the step width in the coding depends on the MSB of  $STC_i$  ( $i = S$  or  $E$ ). The following [Table 17](#) has been copied from the User's Manual, with the corresponding formula added in the last column:

**Table 17 Sample Time Coding**

STCS / STCE	Additional Clock Cycles <sup>1)</sup>	Resulting Sample Time	Clock Cycle Formula
0 0000 <sub>B</sub>	0	$2 / f_{ADCI}$	2 + $STC_i$
0 0001 <sub>B</sub>	1	$3 / f_{ADCI}$	
...	...	...	
0 1111 <sub>B</sub>	15	$17 / f_{ADCI}$	2 + $(STC_i - 15) \times 16$
1 0000 <sub>B</sub>	16	$18 / f_{ADCI}$	
1 0001 <sub>B</sub>	32	$34 / f_{ADCI}$	
...	...	...	
1 1110 <sub>B</sub>	240	$242 / f_{ADCI}$	
1 1111 <sub>B</sub>	256	$258 / f_{ADCI}$	

- 1) The number of resulting additional clock cycles listed in this column corresponds to the term “STC” used in the conversion timing formulas in the Data Sheet.

### **ADC\_TC.H024 Documentation: Filter control only in registers GxRRC7/GxRRC15**

In sections “Finite Impulse Response Filter Mode (FIR)” and “Infinite Impulse Response Filter Mode (IIR)” of the VADC chapter in the User’s Manual,

- replace this sentence:  
“Several predefined sets of coefficients can be selected via bitfield DRCTR (coding listed in Table xx-6) in registers G0RCRy (y = 0 - 15)ff and GLOBRCR.”
- with this sentence:  
“Several predefined sets of coefficients can be selected via bitfield DRCTR (coding listed in Table xx-6) in registers **GxRRC7** and **GxRRC15**.”

### **ADC\_TC.H031 High precision bandgap voltage - documentation update**

The VADC is capable of providing measurement of the internal High Precision Bandgap Reference (HPBG) output voltage  $V_{hpb}$  shared by the PMS subsystem for additional safety supervision. The valid range of the  $V_{hpb}$  signal values currently is not specified.

#### **Detailed description**

The output voltage  $V_{hpb}$  of the HPBG is mapped on VADC channel G0CH12 via the double buffer. The expected nominal value of the signal seen by VADC equals to 2.4 V which is  $2 \times V_{hpb}$ .

The complete range of expected values during normal operation is 1.075 V .. 1.325 V, which corresponds to the range of ADC result values as 1761 (6E1<sub>H</sub>) .. 2171 (87B<sub>H</sub>) assuming VAREF = 5.0V.

The supervision signals are enabled by setting bit GLOBTF.RCEN. For operation at  $f_{ADC1} = 20$  MHz, the recommended sample time setting for this measurement is STC = 0x13.

**ADC\_TC.H038 Multiplexer Diagnostics Connection - Documentation update**

The multiplexer diagnostics feature can pull up the channel input line to  $V_{DDM}$  or pull it down to  $V_{SS}$ .

Figure “Signal Path Test” in the VADC chapter of the User’s Manual erroneously shows a connection to  $V_{DDP}$  instead of  $V_{DDM}$ . Pull-up to  $V_{DDP}$  is not possible.

**Correction**

In figure “Signal Path Test” in the VADC chapter of the User’s Manual, symbol “ $V_{DDP}$ ” shall be replaced by “ $V_{DDM}$ ”.

**ADC\_TC.H041 Offset address of register GxTRCTR - Correction to table “Registers Overview” in User’s Manual**

In table “Registers Overview” in the VADC chapter of the TC22x/TC23x Family User’s Manual V1.0 and V1.1, the offset address for register GxTRCTR (Trigger Control Register, Group x) is incorrectly documented as  $X550_H$ .

*Note: The offset address of register GxTRCTR is correctly specified in the register description of this register at the end of chapter “27.5.1 Queued Request Source Handling”.*

**Documentation correction**

The offset address for register GxTRCTR in table “Registers Overview” in the VADC chapter of the TC22x/TC23x Family User’s Manual V1.0 and V1.1 shall be corrected as follows:

**Table 18 Registers Overview - Correction of offset address for register GxTRCTR**

Register Short Name	Register Long Name	Offset Address
GxTRCTR	Trigger Control Register, Group x	$X554_H$

**ADC\_TC.H042 Precharging of capacitor C<sub>A<sub>IN</sub>SW</sub> - Documentation update**

The following paragraph, which is the last paragraph in section “Input Signal Path” in the VADC chapter of the TC22x/TC23x Family User’s Manual V1.0 and V1.1:

“The capacitor C<sub>A<sub>IN</sub>SW</sub> is automatically precharged to a voltage of approximately half the standard reference voltage V<sub>A<sub>REF</sub></sub> to minimize the average difference between V<sub>A<sub>IN</sub>x</sub> and V<sub>C</sub> at the beginning of a sample phase. Due to varying parameters and parasitic effects, the precharge voltage of C<sub>A<sub>IN</sub>SW</sub> is typically smaller than V<sub>A<sub>REF</sub></sub> / 2.”

shall be replaced as described in the following.

**Documentation update:**

The capacitor C<sub>A<sub>IN</sub>SW</sub> is automatically precharged to a voltage of approximately half the standard reference voltage V<sub>A<sub>REF</sub></sub> while the converter is idle. Due to varying parameters and parasitic effects, the precharge voltage of C<sub>A<sub>IN</sub>SW</sub> is typically smaller than V<sub>A<sub>REF</sub></sub> / 2.

*Note: When conversions are executed in a sequence, or when a conversion cancels a running conversion, the sample phase starts immediately. The converter does not become idle in this case and C<sub>A<sub>IN</sub>SW</sub> is not precharged!*

**ASCLIN\_TC.H001 Bit field FRAMECON.IDLE in LIN slave tasks**

For LIN performing slave tasks, bit field FRAMECON.IDLE has to be set to 000<sub>B</sub> (default after reset), i.e. no pause will be inserted between transmission of bytes.

If FRAMECON.IDLE > 000<sub>B</sub>, the inter-byte spacing of the ASCLIN module is not working properly in all cases in LIN slave tasks (no bit errors are detected by the ASCLIN module within the inter-byte spacing).

### **ASCLIN\_TC.H003 Behavior of LIN Autobaud Detection Error Flag**

#### **Expected Behavior**

In ASCLIN, when auto baud detection (LINCON.ABD) is deactivated, the auto baud measurement should still be active and the Autobaud Detection Error Flag FLAGS.LA should be set when the value measured is outside the BRD.LOWERLIMIT and BRD.UPPERLIMIT range.

#### **Actual Behavior**

The Autobaud Detection Error Flag FLAGS.LA is not set, as the auto baud measurement is not active when auto baud detection is deactivated (LINCON.ABD = 0).

### **ASCLIN\_TC.H004 Changing the Transmit FIFO Inlet Width / Receive FIFO Outlet Width**

#### **Expected Behavior**

The Transmit FIFO should write the data to intended location of TxFIFO, even though the Transmit FIFO inlet width TXFIFOCON.INW is changed between the write operations.

The Receive FIFO should read the data from intended location, even though the Receive FIFO outlet width RXFIFOCON.OUTW is changed between the read operations.

#### **Actual Behavior (Transmit FIFO)**

The Transmit FIFO does not write the data in the intended location when TXFIFOCON.INW is changed in an increasing order (from 1 to 2 to 4) between write operations.

The Transmit FIFO writes the data only to aligned write index based on the number of bytes to be written (TXFIFOCON.INW).

**Example:** Assuming that the write index of TxFIFO is from 0 to 15 (16 bytes), when TXFIFOCON.INW = 2, the TxFIFO writes two bytes of data starting only from half-word aligned write index (0, 2, 4, ..., 14). Similarly when TxFIFO writes four bytes of data starting only from word aligned write index (0, 4, 8, 12).

*Note: This misbehavior is seen only when TXFIFOCON.INW is changed in-between write operations.*

### **Actual Behavior (Receive FIFO)**

The Receive FIFO does not read the data from intended location when RXFIFOCON.OUTW is changed in an increasing order (from 1 to 2 to 4) between read operations.

The Receive FIFO reads the data only from aligned read index based on the number of bytes to be read (RXFIFOCON.OUTW).

**Example:** Assuming that the read index of RxFIFO is from 0 to 15 (16 bytes), when RXFIFOCON.OUTW = 2, the RxFIFO reads two bytes of data starting only from half-word aligned write index (0, 2, 4, ..., 14). Similarly when RxFIFO reads four bytes of data starting only from word aligned read index (0, 4, 8, 12).

*Note: This misbehavior is seen only when RXFIFOCON.OUTW is changed in-between read operations.*

### **Effect**

Previously written data in TxFIFO will be over-written by the new data, when the TxFIFO write index is not aligned with number of data bytes to be written.

Previously read data will be read again, when the RxFIFO read index is not aligned with number of data bytes to be read.

### **Recommendation**

Flush the TxFIFO (TXFIFOCON.FLUSH) or RxFIFO (RXFIFOCON.FLUSH) before TXFIFOCON.INW or RXFIFOCON.OUTW is changed respectively.

### **ASCLIN\_TC.H005 Collision detection error reported twice in LIN slave mode**

An ASCLIN module configured as LIN slave node could report a wrong collision detection error during reception of LIN header after detecting a first correct collision detection error during the transmission of a response field of the previous LIN frame.

This misbehavior is observed under the following sequence:

- The LIN slave node detects a collision detection error when there is a bit error in its transmitted response frame, and then it goes to the idle state as expected.
- The master transmits a header onto the LIN bus, and the LIN slave node receives header and tries to capture the identifier inside the header.
- Then the LIN slave node reports another collision error which is wrongly detected during the reception of identifier although there is no corruption of LIN header on the bus.

### Recommendation

Ignore the collision detection error which happened during reception phase of a LIN slave node.

### **ASCLIN TC.H006 Sample point position when using three samples per bit - Documentation update**

As documented in the description of field BITCON.SAMPLEPOINT, "... if three sample points at position 7, 8, 9 are required, this bit field would contain 9".

In general, if three samples per bit are selected (BITCON.SM = 1<sub>B</sub>), field BITCON.SAMPLEPOINT defines the position of the last sample point.

### Documentation update

The text related to three sample points in figure "ASCLIN Bit Structure" in the ASCLIN chapter of the User's Manual should be updated as follows:

- 16x Oversampling, 3 sample points, relevant sample position 7, 8, 9 (BITCON.OVERSAMPLING = 16, BITCON.SM = 1, BITCON.SAMPLEPOINT = 9)
  - instead of "16x Oversampling, 3 sample points, relevant sample position 8"
- 8x Oversampling, 3 sample points, relevant sample position 3, 4, 5 (BITCON.OVERSAMPLING = 8, BITCON.SM = 1, BITCON.SAMPLEPOINT = 5)
  - instead of "8x Oversampling, 3 sample points, relevant sample position 4"

## **ASCLIN\_TC.H007 Handling TxFIFO and RxFIFO interrupts in single move mode – Documentation update**

### **Present description for TxFIFO single move mode**

As described in section “Single Move Mode” of chapter “TxFIFO interrupt generation” in the User’s Manual, the purpose of the Single Move Mode is to keep the TxFIFO as full as possible, refilling the TxFIFO by writing to it as soon as there is a free element. The single move mode supports primarily a DMA operation using single move per TxFIFO interrupt.

See also the note at the end of this section in the User’s Manual:

***Attention: In Single Move Mode multiple software writes or block DMA moves would lead to multiple interrupts and (false) transaction lost events. Therefore they should be avoided - only single moves should be used.***

To complement the above description, the following two sentences shall be added to the section before the reference to figure “Interrupt generation in the single move mode” in the User’s Manual:

### **Documentation update for TxFIFO single move mode**

If TxFIFO can handle new data, it generates an interrupt but expects just one data of the defined frame width. The DMA or the user should not write multiple data at once to avoid unexpected behavior.

### **Present description for RxFIFO single move mode**

As described in section “Single Move Mode” of chapter “RxFIFO interrupt generation” in the User’s Manual, the purpose of the Single Move Mode is to keep the RxFIFO as empty as possible, by fetching the received elements one by one as soon as possible. The single move mode supports primarily a DMA operation using single move per RxFIFO interrupt.

See also the note at the end of this section in the User’s Manual:

***Attention: In Single Move Mode multiple software reads or block DMA moves lead to multiple interrupts and (false) transaction lost events. Therefore they should be avoided - only single moves should be used.***

To complement the above description, the following two sentences shall be added to the section before the reference to figure “RXFIFO - Interrupt Triggering in the Single Move Mode” in the User’s Manual:

### **Documentation update for RxFIFO single move mode**

If RxFIFO can handle new data, it generates an interrupt but fetches just one data of the defined frame width. The DMA or the user should not read multiple data at once to avoid unexpected behavior.

### **ASCLIN\_TC.H008 SPI master timing – Additional information to Data Sheet characteristics**

The following note shall be added to chapter “ASCLIN SPI Master Timing” in the Data Sheet:

*Note: The specified timings describe the pad capabilities for the respective driver strength configuration. For the maximum achievable baud rate in a given application, the MRST input timings need to be considered in particular.*

### **Background information**

Chapter “ASCLIN SPI Master Timing” in the Data Sheet contains separate tables for different output driver configurations. As can be seen from these tables, the master output timings directly depend on the selected driver strength. The corresponding parameters are marked as controller characteristics with symbol “CC”.

The setup and hold timings for input data received from the slave are marked as system requirements with symbol “SR”. They must be provided by the system in which the device is designed in.

In a given application, the maximum rate at which data can be received from a slave on the master receive input MRST may be limited by the required setup time  $t_{s2}$  (MRST setup to ASCLKO latching edge). As data is shifted by the slave on one edge of ASCLKO and latched by the master on the opposite edge, one phase of ASCLKO must always be greater than the minimum required MRST

setup time (assuming the sampling point is in the middle). This means the ASCLKO period  $t_{50}$  must be  $> 2 \times t_{52}$ .

### **BCU\_TC.H001 HSM Transaction Information not captured**

No HSM transaction information is captured by the System Bus Control Unit (SBCU). Therefore the following HSM related control/status register bits in the SBCU do not have any function:

- Register **SBCU\_DBGRNT** (SBCU Debug Grant Mask Register):
  - **HSMCMI**: this control bit has no function. Behavior as described for SBCU\_DBGRNT.ONE0.
  - **HSMRMI**: this control bit has no function. Behavior as described for SBCU\_DBGRNT.ONE0.
- Register **SBCU\_DGNTT** (SBCU Debug Trapped Master Register):
  - **HSMCMI**: this control bit has no function. Behavior as described for SBCU\_DBGNTT.ONE0.
  - **HSMRMI**: this control bit has no function. Behavior as described for SBCU\_DBGNTT.ONE0.

### **BROM\_TC.H003 Information related to Register FLASH0\_PROCOND**

Chapters “TC2x BootROM Content” of the User’s Manuals contain a description of parts of the FLASH0\_PROCOND register as used by the firmware. This description in subchapter “Configuration by Boot Mode Index (BMI)” shows an incorrect address F800 1030<sub>H</sub>.

Correct is the description of this register in the PMU chapter with address F800 2030<sub>H</sub> (FLASH0 base address F800 1000<sub>H</sub> + offset 1030<sub>H</sub>).

### **BROM\_TC.H009 Re-Enabling Lockstep via BMHD**

For all CPUs with lockstep option, the lockstep functionality is controlled by Boot Mode Headers (BMHD) loaded during boot upon a reset trigger.

If lockstep is disabled for a CPUx with lockstep functionality, re-enabling (e.g. via a different BMHD) is not reliably possible if warm PORST, System or Application reset is executed.

### Recommendation

Use cold PORST if lockstep is disabled and shall be re-enabled upon the reset trigger.

### **BROM TC.H010 Interpretation of value UNIQUE\_CHIP\_ID\_32BIT**

As described in chapter “Debug System handling” in the AURIX™ TC2xx BootROM chapter, the value UNIQUE\_CHIP\_ID\_32BIT is written to the COMDATA register by firmware.

*Note: Unlike the name “UNIQUE\_CHIP\_ID\_32BIT” may suggest, this value only identifies a particular product variant, but not an individual device.*

### **BROM TC.H019 CRC32 ethernet polynomial - Footnote correction**

As documented in the FCE chapter of the User’s Manual, CRC calculation is based on IEEE 802.3, the CRC32 ethernet polynomial used is 0x04C11DB7.

In footnote <sup>2)</sup> below table “Boot Mode Header (BMHD) structure” in the BootROM chapter, the CRC32 ethernet polynomial is erroneously documented as 04C11DB7<sub>1H</sub>.

### Documentation correction

Footnote <sup>2)</sup> below table “Boot Mode Header (BMHD) structure” in the BootROM chapter shall be corrected (trailing “1” deleted) as follows:

- <sup>2)</sup> CRC calculation is based on IEEE 802.3, the CRC32 ethernet polynomial used is 04C11DB7<sub>H</sub>.

**BUS\_TC.H001 CPU access latency for TC21x/TC22x/TC23x - Documentation update**

TC21x/TC22x/TC23x devices have one TC1.6E CPU core with one PSPR and DSPR.

**Documentation update**

The rows in table “CPU access latency in CPU clock cycles for TC21x/TC22x/TC23x” in chapter “On-Chip System Buses and Bus Bridges” in the TC21x/TC22x/TC23x Family User’s Manual referring to

- “.. other PSPR” and
- “.. other DSPR”

do not apply to these devices and shall be ignored.

**BUS\_TC.H002 Reset value for register XBAR\_IDINTEN - Documentation update**

The reset value of the Transaction ID Interrupt Enable register XBAR\_IDINTEN is determined by the configured or enabled masters and slaves.

As no master device is connected to SRI Master Connection Interface MC14 in TC23x/TC22x/TC21x, the corresponding bit XBAR\_IDINTEN.20 is specified as “reserved” (read as 0, should be written with 0) in the register description in the TC23x/TC22x/TC21x User’s Manual. This contradicts the documented XBAR\_IDINTEN reset value of 3031 80D1<sub>H</sub>, where bit 20 is shown as 1<sub>B</sub>.

**Documentation update**

The reset value for register XBAR\_IDINTEN in the TC23x/TC22x/TC21x User’s Manual shall be changed as follows:

- XBAR\_IDINTEN reset value = 3021 80D1<sub>H</sub>

As the TC23x/TC22x/TC21x User’s Manual is a family user’s manual for all device variants, for specific device variants that do not include the full feature set less bits may be set to 1<sub>B</sub> after reset in register XBAR\_IDINTEN.

See also the notes below the description of register XBAR\_IDINTEN in the TC23x/TC22x/TC21x User’s Manual.

**CCU6\_AI.H001 Update of Register MCMOUT**

At every correct Hall event (CM\_CHE), the next Hall patterns are transferred from the shadow register MCMOUTS into MCMOUT (Hall pattern shadow transfer HP\_ST), and a new Hall pattern with its corresponding output pattern can be loaded (e.g. from a predefined table in memory) by software into MCMOUTS. For the Modulation patterns, signal MCM\_ST is used to trigger the transfer.

Loading this register can also be done by writing MCMOUTS.STRHP = 1<sub>B</sub> (for EXPH and CURH) or MCMOUTS.STRMCMP = 1<sub>B</sub> (for MCMP).

*Note: If in a corner case a hardware event occurs simultaneously with a software write where MCMOUTS.STRHP = 1<sub>B</sub> or MCMOUTS.STRMCMP = 1<sub>B</sub>, the current contents of MCMOUTS is copied to the corresponding bit fields of MCMOUT. The new value written to MCMOUTS will be loaded upon the next event.*

**CCU6\_AI.H002 Description of Bit RWHE in Register ISR**

Register ISR (Interrupt Status Reset Register) contains bits to individually clear the interrupt event flags by software. Writing a 1<sub>B</sub> clears the bit(s) in register IS at the corresponding bit position(s), writing a 0<sub>B</sub> has no effect.

In some versions of the User's Manual, the description of bit RWHE (Reset Wrong Hall Event Flag) in column "Description" of register ISR is wrong (description for status 0<sub>B</sub> and 1<sub>B</sub> inverted).

The correct description for bit RWHE is (like for all other implemented bits in register ISR) as shown in the following **Table 19**:

**Table 19 Bit RWHE in register ISR**

Field	Bits	Type	Description
<b>RWHE</b>	13	w	<b>Reset Wrong Hall Event Flag</b> 0 <sub>B</sub> No action 1 <sub>B</sub> Bit WHE will be cleared

**CCU6\_AI.H003 Bit TRPCTR.TRPM2 in Manual Mode - Documentation Update**

In CCU6 chapter “Trap Control Register” of the User’s Manual, the description for bit TRPCTR.TRPM2 = 1<sub>B</sub> (Manual Mode) incorrectly states:

“Manual Mode:

Bit TRPF stays **0** after the trap input condition is no longer valid. It has to be cleared by SW by writing ISR.RTRPF = 1.”

**Correction**

The correct description is as follows:

Manual Mode:

Bit TRPF stays **1** after the trap input condition is no longer valid. It has to be cleared by SW by writing ISR.RTRPF = 1.

**CCU\_TC.H001 Clock Monitor Check Limit Values**

The values for the check limits of the clock monitor have been updated as shown in **Table 20**. This table replaces the corresponding table in chapter “Clock Monitors” of the User’s Manual.

**Table 20 Target trimmed Check limits**

Target Frequency	LOWER value	UPPER value	SELXXX <sup>1)</sup>	Error can be detected for min. deviation	Error is detected for min. deviation
7.5 MHz	0x23	0x27	11 <sub>B</sub>	-4.07% +1.54%	-9.40% +6.35%
6.6 MHz	0x1F	0x23	10 <sub>B</sub>	-4.07% +2.75%	-9.40% +7.50%
6 MHz	0x1C	0x1F	01 <sub>B</sub>	-3.35% +1.54%	-8.43% +6.35%
5 MHz	0x17	0x1A	00 <sub>B</sub>	-2.76% +4.07%	-9.41% +7.50%

1) refers to corresponding bit field xxxSEL in respective CCUCON register

### **CCU\_TC.H002 Oscillator Gain Selection via OSCCON.GAINSEL**

The reset value of OSCCON.GAINSEL = 11<sub>B</sub> provides the default and recommended setting for the oscillator gain. It is not required to modify this value, as the adaptation to a crystal frequency is done via the external circuitry. Therefore, all other gain selections should be regarded as reserved for special application topics, as shown in the following [Table 21](#).

**Table 21 Oscillator Gain Selection via OSCCON.GAINSEL**

Field	Bits	Type	Description
<b>GAINSEL</b>	[4:3]	rw	<b>Oscillator Gain Selection</b> This value should not be changed from the reset value 11 <sub>B</sub> . 00 <sub>B</sub> Low gain 1: reserved for adaptations 01 <sub>B</sub> Low gain 2: reserved for adaptations 10 <sub>B</sub> Low gain 3: reserved for adaptations 11 <sub>B</sub> <b>Maximum gain: default setting</b>

### **Recommendation**

Always to keep the default configuration of OSCCON.GAINSEL = 11<sub>B</sub>.

### **CCU\_TC.H005 References to $f_{PLL2}$ , $f_{PLL2\_ERAY}$ and K3 Divider in User's Manual**

The VADC incorporated in this device uses clocks derived from  $f_{SPB}$ .

Previous design steps (e.g. TC27x Bx, TC26x Ax, TC29x Ax) incorporated a different VADC module also clocked by  $f_{ADC}$ , which could be derived via the K3 divider from  $f_{PLL2}$ ,  $f_{PLL2\_ERAY}$ . These clocks were selected in CCUCON0.[27:26], which is described as "Reserved/Should be written with 0" in the present version of the User's Manual.

Clocks  $f_{PLL2}$ ,  $f_{PLL2\_ERAY}$  and the K3 divider are still described in the present version of the User's Manual.

### Recommendation

- New software implementations should not consider  $f_{PLL2}$ ,  $f_{PLL2\_ERAY}$  and the K3 divider.
- Software ported from previous design steps with a VADC module clocked by  $f_{ADC}$  may be reused on this device step.

### **CCU\_TC.H006 Clock Monitor Support - Documentation Update**

The note at the end of section “Operating the Clock Monitors” in chapter “Clock Monitors”:

*Note: This feature is supported by the Infineon safety driver [safTlib] and there is no additional customer software required.*

should state more precisely:

*Note: The Infineon SafeTlib provides a test for the clock monitor. The clock monitor shall be configured by the application software.*

### **CCU\_TC.H007 Oscillator Watchdog Trigger Conditions for ALM3[0]**

As described in the User's Manual in section “Oscillator Watchdog”, the divider value OSCCON.OSCVAL has to be selected in a way that  $f_{OSCREF}$  is within the range of 2 MHz to 3 MHz, and should be as close as possible to 2.5 MHz.

The Oscillator Watchdog (OSC\_WDT) will trigger the “input clock out of range” alarm ALM3[0] under the following conditions:

- Boundary for **too high** frequencies:
  - for  $(OSCVAL+1) \times 6.25 \leq f_{OSC} \text{ [MHz]} \leq (OSCVAL+1) \times 7.5$ , an alarm can be generated, but there is no guarantee that it is generated,
  - for  $f_{OSC} \text{ [MHz]} > (OSCVAL+1) \times 7.5$ , an alarm is always generated.
- Boundary **for too** low frequencies:
  - for  $(OSCVAL+1) \times 1.25 \leq f_{OSC} \text{ [MHz]} \leq (OSCVAL+1) \times 1.67$ , an alarm can be generated, but there is no guarantee that it is generated,

- for  $f_{OSC}$  [MHz]  $< (OSCVAL+1) \times 1.25$ , an alarm is always generated.

The accuracy of these limits [in %] depends on the variation [in %] of the back up clock (see specification of  $f_{BACKUT}$  and  $f_{BACKT}$  in the Data Sheet).

### Example

- For  $f_{OSC} = 20$  MHz, selecting  $OSCVAL = 7$  results in  $f_{OSC} = 2.5$  MHz.
  - An alarm for too high frequencies can be generated for  $f_{OSC} \geq 50$  MHz,
  - An alarm for too high frequencies is always generated for  $f_{OSC} > 60$  MHz.
  - An alarm for too low frequencies can be generated for  $f_{OSC} \leq 13.36$  MHz,
  - An alarm for too low frequencies is always generated for  $f_{OSC} < 10$  MHz.

### **CCU\_TC.H010 Oscillator Mode control in register OSCCON - Documentation Update**

The description for setting  $OSCCON.MODE = 00_B$  in register OSCCON must be changed from

- “External Crystal / Ceramic Resonator Mode and External Input Clock Mode. The oscillator Power-Saving Mode is not entered.”

to:

- “External Crystal / Ceramic Resonator Mode. The oscillator Power-Saving Mode is not entered.”

### **Recommendation**

When using an external input clock signal connected to XTAL1 (XTAL2 open), do not use setting  $OSCCON.MODE = 00_B$ . Instead, use setting  $OSCCON.MODE = 10_B$ .

### **CPU\_TC.H006 Store Buffering in TC1.6/P/E Processors**

#### **Overview**

Store buffering is a method of increasing processor performance by decoupling memory write operations from the instruction execution flow within the CPU. All

write data is placed in a FIFO buffer (known as the store buffer) by the CPU prior to being read by the memory/bus interfaces and written to memory. This allows the processor to continue execution without waiting for the write data to be written to the target memory location. Data is written to the store buffer at processor speed and read from the store buffer at memory/bus speed. Typically the read bandwidth from the store buffer will exceed the write bandwidth from the processor, only if the store buffer fills will the processor stall.

To further increase performance memory read operations are prioritised ahead of memory write operations from the store buffer. This ensures that the processor does not stall on data loads while data writes are pending in the store buffer. A side effect of this prioritising is that memory may not be accessed in program order.

### Operational Details

The function of the store buffer is designed to be invisible to the end user under normal operation:

- All CPU load operations are checked against the store buffer contents. Data for matching load addresses is either immediately forwarded to the CPU from the store buffer (TC1.6, TC1.6P) or written to memory prior to the load operation proceeding (TC1.6E).
- All loads and store operations to peripheral regions (typically segments  $E_H$  and  $F_H$ ) are performed in strict program order (no load prioritisation).

The operation of the store buffer can become visible when in-order memory access is required to non-peripheral segments.

This can occur under the following circumstances:

- When programming flash memory.
- When performing memory testing with the processor.
- When data is required to be in memory for inter-core/inter-module communication.

In such cases the following solutions may be employed:

- The store buffer may be explicitly flushed by use of a DSYNC instruction.
- The store buffer may be disabled by setting `SMACON.IODT`. This should not be done during normal operation as it significantly impacts performance.

**Examples**

The following examples refer to memory accesses to non-peripheral regions (i.e. segments  $0_H .. D_H$ ):

**Example-1a Out of order memory access due to load prioritisation**

Program Flow	-	Memory Access
st-1		ld-4
st-2		ld-5
st-3		ld-6
ld-4		st-1
ld-5		st-2
ld-6		st-3

**Example-1b In order memory access enforced by DSYNC**

Program Flow	-	Memory Access
st-1		st-1
st-2		st-2
st-3		st-3
dsync		
ld-4		ld-4
ld-5		ld-5
ld-6		ld-6

**Example-2a Load forwarding from store buffer - no memory read (TC1.6/1.6P)**

Program Flow	-	Memory Access
st.w [a0], d0		
ld.w d1, [a0]		st.w [a0], d0

**Example-2b In order memory access enforced by DSYNC (TC1.6/1.6P)**

Program Flow	-	Memory Access
st.w [a0], d0		st.w [a0], d0
dsync		
ld.w d1, [a0]		ld.w d1, [a0]

## **CPU\_TC.H008 Instruction Memory Range Limitations**

To ensure the processor cores are provided with a constant stream of instructions the Instruction Fetch Units will speculatively fetch instructions from up to 64 bytes ahead of the current Program Counter (PC).

If the current PC is within 64 bytes of the top of an instruction memory the Instruction Fetch Unit may attempt to speculatively fetch instructions from beyond the physical range. This may then lead to error conditions and alarms being triggered by the bus and memory systems.

### **Recommendation**

It is therefore recommended that either the MPU is used to define the allowable executable range or that the upper 64 bytes of any memory be initialized but unused for instruction storage for the TC1.6.\* class processors. For TC1.3.\* class processors this may be reduced to 32 bytes.

## **CPU\_TC.H009 Details on CPU Clock Control**

As described in chapter “Clock Control Unit” of the User’s Manual, the effective CPU execution frequency may be reduced by programming the associated bit field CPUxDIV in register CCUCONn (where x is the core number, and n = x+6).

The effective execution frequency  $f_{\text{CPUx}}$  seen by CPUx is given by the following equation (where  $f_{\text{SRI}}$  is the base SRI frequency):

- $f_{\text{CPUx}} = f_{\text{SRI}} * (64 - \text{CPUxDIV}) / 64$

A CPUxDIV value of 0 results in the core CPUx being clocked at the SRI frequency (no frequency reduction).

To avoid synchronisation issues typically associated with clock division the clock control mechanism stalls the issue of instructions into the processor pipeline rather than by modifying the actual applied clock. An incoming instruction fetch packet is stalled for the number of cycles required to approximate the required execution frequency. The stall is seen by the processor as a stall in the instruction stream in the same way a stalling instruction memory would be seen.

In most scenarios this mechanism provides a good approximation to clock division based control. The actual reduction in effective frequency will be dependent on the code executed.

When determining IPC rates as described in AP32168 (Application Performance Optimization for TriCore V1.6 Architecture), note that for CPUxDIV > 0, field Count Value in register CCNT still represents SRI clock cycles.

### **CPU\_TC.H012 Behavior of bit-wise operations on certain peripheral register bits which need to be written back with the same value**

The LDMST, ST.T, CMPSWAP.W, SWAPMSK.W and SWAP.W instructions in the AURIX™ microcontrollers are instructions intended to provide atomicity as well as bit-wise operations to a targeted memory location or peripheral register. They are also referred to as Read-Modify-Write (RMW) instructions.

In some registers in certain modules, a bit has to be written with the same value (e.g. a bit set to  $1_B$  has to be written with a  $1_B$  to perform an operation).

When using a RMW instruction to write to such a bit, the write is masked away and will not happen at all.

*Note: Writing a different value (e.g. writing a  $1_B$  to a bit currently at  $0_B$ ) is not affected, and works as expected to modify only the selected bit.*

**Example:** Consider the GxVFR register in the VADC module:

GxVFR (x = 0 - 10)  
Valid Flag Register, Group x (x \* 0400<sub>H</sub> + 05F8<sub>H</sub>)      Reset Value: 0000 0000<sub>H</sub>

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
r	r	r	r	r	r	r	r	r	r	r	r	r	r	r	r
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
VF15	VF14	VF13	VF12	VF11	VF10	VF9	VF8	VF7	VF6	VF5	VF4	VF3	VF2	VF1	VF0
rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh	rwh

Field	Bits	Type	Description
VFy (y = 0 - 15)	y	rwh	Valid Flag of Result Register x Indicates a new result in bitfield RESULT or in bit FCR. 0 <sub>B</sub> Read access: No new valid data available Write access: No effect 1 <sub>B</sub> Read access: Result register x contains valid data and has not yet been read, or bit FCR has been updated Write access: Clear this valid flag and bitfield DRC in register GxRESy (overrides a hardware set action)

**Figure 7 Register GxVFR in the VADC Module of TC2xx Devices**

The bits in the GxVFR register have to be written with 1<sub>B</sub> to clear a valid flag VFy indicating a valid result. Assuming VFy = 1<sub>B</sub>, if one of the RMW instructions listed above is used, the write to VFy would never happen since VFy is already set to 1<sub>B</sub>. This means that the next read of VFy may lead to incorrect conclusions by software.

### Affected Modules and Registers in the AURIX™ Platform

- CCU6: IMON
- VADC: GxVFR, GxSEFLAG, GxCEFLAG, GxREFLAG, GLOBEFLAG.

*Note: VADC is located outside the addressable range of ST.T, so ST.T need not be considered in the context of VADC.*

### Recommendation

In the affected modules, use only direct writes (i.e, write the whole register as a 32-bit word), and do not use RMW operations to write to such bits.

For example, to clear bit VF0 in the GxVFR register, the software should write:

```
VADC_GxVFR.U = 0x00000001;
```

Here .U implies writing the whole 32-bit register as an unsigned integer.

### **CPU\_TC.H014 ACCEN\* Protection for Write Access to Safety Protection Registers - Documentation Update**

The access protection symbol 'P' to indicate protection by the ACCEN\* register mechanism is missing in column "Access Mode - Write" in table "Safety Protection Registers" in the CPU chapter of the User's Manual for RGN\*x registers with an index  $x \geq 4$ , and for register ACCENA.

Actually, these registers also have write access attribute 'P'.

### **CPU\_TC.H015 Register Access Modes for Safety Protection Registers - Documentation Update**

The access protection symbol 'U' is erroneously included and should be removed in column "Access Mode - Write" for all registers in table "Safety Protection Registers" in the CPU chapter of the User's Manual.

The note below this table is rephrased as follows:

*Note: A disallowed access to any CPU register (e.g. attempted write to non-existent register, attempted write to read only register, attempted access to E without Endinit, etc.) will NOT result in a Bus Error*

### **CPU\_TC.H017 MSUB.Q does not match MUL.Q+SUB - Documentation Update**

The AURIX™ implementation of MSUB.Q uses infinitely precise intermediate results. In contrast with AUDO™ devices this can lead to different observable results for MSUB.Q when compared with a MUL.Q+SUB sequence.

The following table describes these differences in the MSUB.Q behaviour in AURIX™ 1st and 2nd generation products.

*Note: The TriCore™ TC1.6.2 Core Architecture Manual (Vol.2 Instruction Set) V1.1 and following for 2nd Generation AURIX™ (TC3xx) contains these new definitions.*

*Note: For 1st generation AURIX™ devices (TC2xx), this is a documentation update to the TriCore™ TC1.6P & TC1.6E Core Architecture Manual V1.0D15 (Vol.2 Instruction Set).*

**Table 22 MSUB.Q Definitions in AURIX™ different from AUDO™**

<b>Secondary Opcode [23:18]</b>	<b>Instruction Mnemonic</b>	<b>Updated Description</b>
0x00	<b>MSUB.Q D[c], D[d], D[a], D[b] U, n</b> 32 - (32 * 16U)Up --> 32	result = ({D[d], 16'h0000} - ((D[a] * D[b][31:16] << n)) >> 16; D[c] = result[31:0]; // Fraction
0x01	<b>MSUB.Q D[c], D[d], D[a], D[b] L, n</b> 32 - (32 * 16L)Up --> 32	result = ({D[d], 16'h0000} - ((D[a] * D[b][15:0] << n)) >> 16; D[c] = result[31:0]; // Fraction
0x02	<b>MSUB.Q D[c], D[d], D[a], D[b], n</b> 32 - (32 * 32)Up --> 32	result = ({D[d], 32'h0000_0000} - ((D[a] * D[b]) << n)) >> 32; D[c] = result[31:0]; // Fraction
0x20	<b>MSUBS.Q D[c], D[d], D[a], D[b] U, n</b> 32 - (32 * 16U)Up --> 32	result = ({D[d], 16'h0000} - ((D[a] * D[b][31:16] << n)) >> 16; D[c] = ssov(result, 32); // Fraction
0x21	<b>MSUBS.Q D[c], D[d], D[a], D[b] L, n</b> 32 - (32 * 16L)Up --> 32	result = ({D[d], 16'h0000} - ((D[a] * D[b][15:0] << n)) >> 16; D[c] = ssov(result, 32); // Fraction
0x22	<b>MSUBS.Q D[c], D[d], D[a], D[b], n</b> 32 - (32 * 32)Up --> 32	result = ({D[d], 32'h0000_0000} - ((D[a] * D[b]) << n)) >> 32; D[c] = ssov(result, 32); // Fraction

**DAP\_TC.H002 DAP client\_blockread in Combination with TGIP and all Parcels with CRC6**

*Note: This problem is only relevant for tool development, not for application development.*

When issuing a DAP client\_blockread telegram together with the TGIP (Trigger in Protocol) option (DAPISC.TGIP = 1) the TGIP extra bit is appended for each parcel in case “all parcels with CRC6” is enabled. This causes a slight increase in the communication length compared to the correct behavior of having a TGIP bit only for the last parcel.

**Recommendation**

Do not use the TGIP and “CRC6 for all parcels” features together in case this extra bit can not be tolerated. If the Trigger in Protocol and increased communication safety is required TGIP can be used together with the CRC32 option (see also DAP\_TC.002 DAP client\_blockread has Performance issue in Specific Operation Modes).

**DAP\_TC.H003 Not acknowledged DAP telegrams in noisy environments**

*Note: This problem is only relevant for tool development, not for application development.*

DAP telegrams always follow a request-reply scheme. The request is driven by the tool, the reply by the AURIX™. The AURIX™ acknowledges a correctly received telegram always by a reply, which consists at least of a start-bit. DAP communication in noisy environments might result in invalid telegrams. This can leave the IOClient in an intermediate state which requires an IOClient reset.

If AURIX™ receives an invalid telegram with a wrong CRC6 or length field, it does not reply at all and in some cases the selected IOClient might be left in an intermediate state in case of a detected client\_write/blockwrite/readwrite tool request.

### Recommendation

If a tool does not receive a start bit as an acknowledge for an IOClient request, a client\_reset must be sent as the next telegram for the selected IOClient. Tool interaction with the DAP module itself is not affected and can be done in between.

### **DMA\_TC.H002 Bit CHCSRz.BUFFER can be toggled when not in Double Buffer Mode**

The purpose of bit CHCSRz.BUFFER is to indicate which buffer is read or filled during DMA double buffering (selected in bitfield ADICRz.SHCT).

However, bit CHCSRz.BUFFER can also be toggled by writing bit CHCSRz.SWB = 1<sub>B</sub> when not in Double Buffer Mode.

### Recommendation

Do not write bit CHCSRz.SWB = 1<sub>B</sub> when not in Double Buffer Mode.

### **DMA\_TC.H004 Transaction Request Lost upon software trigger with pattern match**

If a DMA channel is configured for pattern detection and software triggering of each DMA transfer (CHCSRz.RROAT = 0<sub>B</sub>), then if there is a new DMA software request received while a DMA transfer is executing then a Transaction Request Lost event may be lost.

### Recommendation

The loss of TRL status is a debug feature. A DMA channel should be used such that TRL is not set.

The user must ensure that the CPU triggers a new DMA software request when no DMA access is pending. The software could poll the TSRz.CH bit to confirm it is 0<sub>B</sub> before issuing a DMA software trigger.

### **DMA\_TC.H005 Linked List Transfer leading to loading of non-Linked List TCS causes corruption**

If on completion of a Linked List (LL) a non-LL Transaction Control Set (TCS) is loaded with shadow address buffering enabled (read only and direct write) then the new non-LL TCS can be corrupted.

#### **Recommendation**

Shadow address buffering must be disabled in the non-LL TCS (SHCT[3:0] = 0000<sub>B</sub>)

### **DMA\_TC.H006 Clearing of HTRE when DMA channel is configured for Single Mode**

The DMA may be used to support a peripheral with a high interrupt rate where the interrupts are generated in quick succession (e.g. a QSPI filling a TXFIFO).

The DMA channel z is configured with the following settings:

- Single Mode (HTRE is reset by hardware on completion of a DMA transaction)
  - TSRz.CHMODE = 0<sub>B</sub>
- Request required for each DMA Transfer
  - TSRz.RROAT = 0<sub>B</sub>

If the DMA channel is configured to execute a DMA transaction of 1 x DMA transfer of 2 x DMA moves:

- Block Mode: 2 x DMA Move per DMA transfer
  - DMA\_CHCFGRz.BLKM = 001<sub>B</sub>
- Transfer Reload Value: 1 x DMA transfer
  - DMA\_CHCFGRz.TREL = 1<sub>B</sub>

then additional DMA moves are executed unexpectedly.

#### **Explanation of Effect**

If the peripheral generates two interrupt service requests in relatively quick succession then the first DMA hardware request is serviced by the DMA and performs one DMA transfer comprising two DMA moves. The second DMA

hardware request arrives before the completion of the first DMA transfer (i.e. before the clearing of HTRE at the end of the DMA transaction). The second hardware request is serviced by the DMA and performs a second DMA transfer comprising two DMA moves.

### Recommendation

If the second DMA hardware request arrives before completion of the first DMA transfer then the DMA channel Block Mode must limit a DMA transfer to one DMA move:

- `DMA_CHCFGRz.BLKM = 000B; //1 x DMA move/DMA transfer`

The total number of DMA moves must be defined by the Transfer Reload Value `DMA_CHCFGRz.TREL`.

### DMA\_TC.H007 Selecting the Priority for DMA Channels

All used DMA channels should be configured with the **highest** priority on SPB in respect to other used SPB master agents (CPUs, HSSL, ETH) to enable a robust execution of the configured DMA transactions.

The DMA channels are configured per default with the lowest priority on SPB:

- `DMA_CHCFGRz.DMAPRIO = 00B --> maps DMA channel z SPB requests to SPB priority DMAL`
- `SBCU_PRIOH.DMAL = 1111B --> configures DMAL with the lowest priority on SPB`

### Recommendation

There are several ways to configure used DMA channels with the highest priority on SPB with respect to other SPB master agents. Two examples follow:

#### Example1

Map the used DMA channels to SPB priority DMAH by setting `DMA_CHCFGRz.DMAPRIO = 11B` and keep the configuration of the DMAH priority (`SBCU_PRIOL.DMAH = 0000B`).

## Example2

Keep the mapping of the used DMA channels to DMAL (DMA\_CHCFGRz.DMAPRIO = 00<sub>B</sub>) and change the priority configuration of DMAL (e.g. set SBCU\_PRIOH.DMAL = 0001<sub>B</sub>).

## Background

The DMA can request for SPB access with three different requests (DMAH, DMAM, DMAL) that are configured with different SPB priorities with respect to the other SPB master agents (CPUx, HSCT, ETH). The priority of the DMA requests DMAH, DMAM and DMAL on the SPB in respect to the priority of other SPB master agents can be configured via the SBCU registers SBCU\_PRIOL / SBCU\_PRIOH.

Each DMA channel z can be configured via DMA\_CHCFGRz.DMAPRIO regarding which of three priorities (DMAH, DMAM or DMAL) it uses for SPB access.

The default configuration of DMA\_CHCFGRz.DMAPRIO = 00<sub>B</sub>. This means that the channels will request for SPB access with the DMAL priority.

The priority of a DMAL request on SPB is configured per default with the lowest priority (SBCU\_PRIOH.DMAL = 1111<sub>B</sub>).

## DMA\_TC.H008 Transaction Request State

The DMA Transaction Request State bit DMA\_TSRz.CH is cleared when the DMA transfer starts (RROAT = 0<sub>B</sub>) or at the end of a DMA transaction (RROAT = 1<sub>B</sub>).

Figure “Channel Request Control” and RROAT bit field description of register DMA\_MExCHCR in chapter “Register Description” of the User’s Manual are wrong.

## DMA\_TC.H009 Resetting Bits ICH and IPM in register CHCSRz

The Clear Interrupt from Channel bit (CICH) is accessible via the DMA channel CHCSR register.

The AURIX™ TC2xx User Manuals are incorrect with respect to the following statement:

- The DMA channel DMA\_CHCSRz ICH and IPM bit field description states: “is reset by software when writing a 1 to ADICRz.CICH”.

### Correction

- The text should read: “is reset by software when writing a 1 to **CHCSRz.CICH**”.

### **DMA\_TC.H010 Calculation of DMA Address Checksum for DMA read moves to Cacheable Addresses**

The DMA Move Engine (ME) stores the DMA read move data in eight 32-bit read registers. If a DMA read move is to a cached address (Segment 8 or 9), the ME shall translate the DMA read move access to the on chip bus into an SRI BTR4 access to a 32-byte aligned address. The DMA shall calculate the DMA address checksum from the on chip bus address i.e. the 32-byte aligned address. The DMA shall store the DMA address checksum in the SDCRCR.

### Recommendation

If an expected DMA address checksum is pre-calculated to test the DMA address generation, the user shall take note of the address translation to 32-byte aligned addresses when calculating the expected DMA address checksum from a cacheable DMA source address.

Alternatively, DMA read moves should be performed to non-cacheable source addresses (segments A and B).

### **DMA\_TC.H011 DMA\_ADICRz.SHCT - Reserved Values**

The DMA channel shadow control bit field DMA\_ADICRz.SHCT controls the function of the shadow address register. If software programs a reserved value in DMA\_ADICRz.SHCT, the DMA may deadlock the operation of the DMA.

Therefore, software shall not program DMA\_ADICRz.SHCT with the following reserved values:

- 0011<sub>B</sub> Reserved
- 0100<sub>B</sub> Reserved
- 0111<sub>B</sub> Reserved.

### **DMA\_TC.H012 TCS Update in Halt State**

If a DMA channel is in halt state,

- The DMA shall stop performing DMA moves to the destination location.
- Software may perform a background test on the destination location.
- Software may modify the DMA channel Transaction Control Set (TCS).

### **Recommendation**

If software modifies the DMA channel TCS, software shall only modify the DMA channel source address (DMA\_SADRz.SDAR) and the DMA channel destination address (DMA\_DADRz.DADR).

### **DMA\_TC.H013 MExSR.WS and MExSR.RS Status Bits**

As documented in the User's Manual, the Move Engine (ME) status bits RS/WS in register MExSR are set when the ME is performing a read move or DMA write move. This means:

- MExSR.RS = 1<sub>B</sub> when the ME is performing a DMA read move for the active DMA channel.
- MExSR.WS = 1<sub>B</sub> when the ME is performing a DMA write move for the active DMA channel.

It should be noted that the setting of these bits is not restricted to DMA read move and DMA write move. Additionally the status bits may be set when the ME is performing other operations:

- MExSR.RS = 1<sub>B</sub> when the ME is loading a new Transaction Control Set in a linked list.
- MExSR.WS = 1<sub>B</sub> when the ME is writing a DMA timestamp.

*Note: The additional setting of the ME status bits may be observed when debugging the operation of the DMA. There is no effect on the operation of the DMA.*

### **DMA\_TC.H016 DMARAM ECC Error Disable**

If software disables SPB bus errors caused by DMARAM ECC errors ( $\text{DMA\_MEMCON.ERRDIS} = 1_{\text{B}}$ ), the DMA will not correctly acknowledge a Read Modify Write (RMW) access on the SPB bus.

#### **Recommendation**

The application software must always enable the reporting of SPB errors ( $\text{DMA\_MEMCON.ERRDIS} = 0_{\text{B}}$ ; default after reset).

### **DMA\_TC.H017 DMA Channel Request Control - Documentation Update**

The following text (located below figure “Channel Request Control” in section “DMA Channel Request Control” of the DMA chapter in the User’s Manual):

“If  $\text{CHCFGRz.PRSEL} = 1$  in the current DMA channel z can bypass the ICU and trigger a DMA hardware request in the next lower DMA channel z-1. The latency to service a DMA channel z-1 request is reduced. DMA channel z interrupt service requests are disabled.”

#### **should read as:**

“If  $\text{DMA\_CHCFGRz.PRSEL} = 1$  **is selected** in the current DMA channel z, a **DMA channel trigger** can bypass the ICU and trigger a DMA hardware request in the next lower DMA channel z-1. The latency to service a DMA channel z-1 request is reduced. DMA channel z interrupt service requests are disabled.”

### **DTS\_TC.H001 Update of Bit DTSSTAT.BUSY**

The following statement in the description of bit BUSY in register DTSSTAT in the SCU chapter “Die Temperature Measurement” is incorrect:

*Note: This bit is updated 2 cycles after bit  $\text{DTSCON.START}$  is set.*

## Correction

The correct description is as follows:

*Note: This bit is updated 7 cycles after bit DTSCON.START is set.*

### **ENDINIT\_TC.H001 Endinit Protection for Registers KRST0, KRST1, KRSTCLR**

The access protection symbol ‘E’ to indicate Endinit-protection is missing in column “Access Mode - Write” in table “Register Overview” in the User’s Manual for the following registers:

- KRST0, KRST1, KRSTCLR

of the following modules (if implemented):

- E-Ray, ETH, PS15.

### **FLASH\_TC.H007 Advice for using Suspend and Resume**

As documented in the User’s Manual section “Operation Suspend and Resume”, an operation is suspended by writing ‘1’ to MARD.SPND. The Flash operation stops when it reaches an interruptible state. After that the flag FSR.SPND is set and BUSY is cleared.

The 1-to-0 transition of MARD.SPND alone is not indicating if the suspend request has been executed and the Flash can accept a new command. The BUSY flags have to be checked to determine if the Flash is still busy with the current operation. Only after the 1-to-0 transition of the BUSY flags the flag FSR.SPND indicates if the operation has finished or if it is in suspended state.

The following recipe describes the best practice for using suspend and resume.

#### **Suspending an Erase Operation**

In case of a request for suspending an ongoing erase operation:

As documented in the User’s Manual: Please ensure that between start or resume of an erase process and the suspend request normally at least ~1 ms erase time can pass.

- Check if the corresponding BUSY flag has already cleared. If yes, no suspend is necessary.
- Request the suspend with control flag MARD.SPND = 1<sub>B</sub>.
- Wait until the BUSY flag clears.
- After that check FSR.SPND. If this is 1<sub>B</sub> then the operation was suspended and needs to be resumed later. If this is 0<sub>B</sub> the operation has already finished, therefore no resume is necessary.
- Now new Flash operations are allowed with the restrictions documented in User's Manual section "Operation Suspend and Resume".

Note for PFlash erase operations in bank x that PxBUSY and D0BUSY are set at the beginning. The D0BUSY is cleared early after updating the Erase Counters, and PxBUSY is cleared when the erase operation has finished. Therefore, for PFlash the PxBUSY flag has to be used. (Polling for PxBUSY and DxBUSY can be a generic solution for suspend sequences before checking the SPND state.) Interrupt driven software receives two interrupts!

### Resuming a Suspended Erase Operation

The resume of the suspended erase operation is done in these steps:

- Resume the operation with the command sequence "Resume Prog/Erase".
- Wait until FSR.SPND is 0<sub>B</sub>.
- After that wait for the end of the operation signalled by BUSY going to 0<sub>B</sub>.

### Suspending a Program Operation

In case of a request for suspending an ongoing programming operation:

- Request the suspend with control flag MARD.SPND = 1<sub>B</sub>.
- Wait until the BUSY flag clears.
- After that check FSR.SPND. If this is 1<sub>B</sub> then the operation was suspended and needs to be resumed later. If this is 0<sub>B</sub> the operation has already finished, therefore no resume is necessary.
- Now new Flash operations are allowed with the restrictions documented in User's Manual section "Operation Suspend and Resume".

### Resuming a Suspended Program Operation

The resume of the suspended programming operation is done in these steps:

- Resume the operation with the command sequence “Resume Prog/Erase”.
- Wait until FSR.SPND is 0<sub>B</sub>.
- After that wait for the end of the operation signalled by BUSY going to 0<sub>B</sub>.

## **FLASH\_TC.H008 Understanding Flash Retention/Endurance Figures in the Data Sheet**

Flash retention/endurance is documented in the Data Sheet by the following parameters

- Program Flash Retention Time  $t_{RET}$  for PFlash,
- UCB Retention Time  $t_{RTU}$  for the UCBs,
- Data Flash Endurance per EEPROMx sector  $N_{E\_EEP10}$  for DFlash0,
- Data Flash Endurance per HSMx sector  $N_{E\_HSM}$  for DFlash1 (if available).

### **Retention**

To emphasize the importance of retention, the PFlash and UCB parameters are described as retention time under the condition of a maximum number of cycles.

The value “Min. x years” has to be interpreted as: the data retention is at least x years, i.e. x years or longer after the last programming data stays readable.

The condition “Max. y erase/program cycles” means: this data retention figure is valid if there were not more than y erase/program cycles.

### **Endurance**

For the DFlash the endurance is most important, therefore as parameter the number of cycles under the condition of the retention is given.

The value “Min. x cycles” has to be interpreted as: at least x cycles can be applied.

The condition “Max. data retention time y years” means: this endurance figure is valid if the expected data retention after the last programming is maximum y years.

*Note: As general remark, these figures are only valid if the parameters given in the Data Sheet are adhered to in their entirety.*

## **FLASH\_TC.H022 Flash Wait State configuration**

Configuring flash wait states in your application is critical for correct operation. Refer to these parts of the documentation of the respective TC2\*x design step for guidance on avoiding data read errors over the lifetime of the device:

- Data Sheet, chapter “Flash Parameters”:
  - minimum access times  $t_{PF}$  /  $t_{PFEC}$  for PFLASH,
  - and  $t_{DF}$  /  $t_{DFEC}$  for DFLASH
- AURIX™ TC2\*x User’s Manual, PMU chapter “Configuring Flash Wait Cycles”

When **increasing** the SRI and FSI clock frequencies: first set the wait state bitfields (WSECPF, WSPFLASH, WSECDF, and WSDFLASH) in register FCON to the correct values, and then change the clock configuration.

When **decreasing** the SRI and FSI clock frequencies: first change the clock configuration, and then set the wait state bitfields (WSECPF, WSPFLASH, WSECDF, and WSDFLASH) in register FCON to the correct values.

*Note: Applications that omit configuration of FCON may work in the development phase, but encounter data read errors in the field.*

## **FlexRay\_AI.H004 Only the first message can be received in External Loop Back mode**

If the loop back (TXD to RXD) will be performed via external physical transceiver, there will be a large delay between TXD and RXD.

A delay of two sample clock periods can be tolerated from TXD to RXD due to a majority voting filter operation on the sampled RXD.

Only the first message can be received, due to this delay.

To avoid that only the first message can be received, a start condition of another message (idle and sampling '0' -> low pulse) must be performed.

The following procedure can be applied at one or both channels:

- wait for no activity ( $TEST1.AOx=0$  -> bus idle)
- set Test Multiplexer Control to I/O Test Mode ( $TEST1.TMC=2$ ), simultaneously  $TXDx=TXENx=0$

- wait for activity (`TEST1.AOx=1` -> bus not idle)
- set Test Multiplexer Control back to Normal signal path (`TEST1.TMC=0`)
- wait for no activity (`TEST1.AOx=0` -> bus idle)

Now the next transmission can be requested.

### **FlexRay AI.H005 Initialization of internal RAMs requires one eray\_bclk cycle more**

The initialization of the E-Ray internal RAMs as started after hardware reset or by CHI command `CLEAR_RAM` (`SUCC1.CMD[3:0] = 1100B`) takes 2049 eray\_bclk cycles instead of 2048 eray\_bclk cycles as described in the E-Ray Specification.

Signalling of the end of the RAM initialization sequence by transition of `MHDS.CRAM` from `1B` to `0B` is correct.

### **FlexRay AI.H006 Transmission in ATM/Loopback mode**

When operating the E-Ray in ATM/Loopback mode there should be only one transmission active at the same time. Requesting two or more transmissions in parallel is not allowed.

To avoid problems, a new transmission request should only be issued when the previously requested transmission has finished. This can be done by checking registers `TXRQ1/2/3/4` for pending transmission requests.

### **FlexRay AI.H007 Reporting of coding errors via `TEST1.CERA/B`**

When the protocol engine receives a frame that contains a frame CRC error as well as an FES decoding error, it will report the FES decoding error instead of the CRC error, which should have precedence according to the non-clocked SDL description.

This behaviour does not violate the FlexRay protocol conformance. It has to be considered only when `TEST1.CERA/B` is evaluated by a bus analysis tool.

### **FlexRay\_AI.H009 Return from test mode operation**

The E-Ray FlexRay IP-module offers several test mode options

- Asynchronous Transmit Mode
- Loop Back Mode
- RAM Test Mode
- I/O Test Mode

To return from test mode operation to regular FlexRay operation we strongly recommend to apply a hardware reset via input `eray_reset` to reset all E-Ray internal state machines to their initial state.

*Note: The E-Ray test modes are mainly intended to support device testing or FlexRay bus analyzing. Switching between test modes and regular operation is not recommended.*

### **FlexRay\_AI.H011 Behavior of interrupt flags in FlexRay™ Protocol Controller (E-Ray)**

In the corner case described below, the actual behavior of the interrupt flags of the FlexRay™ Protocol Controller (E-RAY) differs from the expected behavior.

*Note: This behaviour only applies to E-RAY interrupts INT0 and INT1. All other E-RAY interrupts are not affected.*

#### **Expected Behavior**

When clearing an interrupt flag by software, the resulting value of the flag is expected to be zero.

A hardware event that occurs afterwards then leads to a zero to one transition of the flag, which in turn leads to an interrupt service request.

#### **Actual Behavior in Corner Case**

When the interrupt flag is being cleared by software in the same clock cycle as a new hardware event sets the flag again, then the hardware event wins and the flag remains set without being cleared.

As interrupt requests are generated only upon zero to one transitions of the flag, no interrupt request will be generated for this flag until the flag is successfully cleared by software later on.

### Workaround

After clearing the flag, the software shall read the flag and repeat clearing until the flag reads zero.

### **FlexRay\_TC.H002 Initialization of E-Ray RAMs**

After Power-on reset the ECC codes in the E-Ray RAMs may be set to an arbitrary state. Therefore the E-Ray RAM must be cleared and the ECC codes set to a defined state to avoid unintended traps.

To achieve this the following alternative methods are proposed:

#### **Method 1 using the MTU/MBIST:**

- Clear all E-Ray RAMs and the related ECC code storage by executing writes to all RAM locations using the AURIX MBIST engine. The MBIST engine supports filling the E-Ray RAM with ECC-correct patterns. For this purpose the AURIX MBIST auto-initialization algorithm can be used. See section “Filling a Memory with Defined Contents” in the corresponding User’s Manual/Target Specification. The following E-Ray RAM blocks have to be initialized with correct data:
  - Output Buffer
  - Input Buffer
  - Message BuffersThe MBIST function to be executed for each buffer is the same, only the function parameters have to be adapted.
- Execute one read from each E-Ray RAM block using the AURIX MBIST engine (reading from all E-Ray RAM locations is an alternative but not necessary solution). For this purpose the AURIX MBIST engine can also be used.
- Insert at the end of all MBIST function calls a status check, which makes sure that the launched MBIST tests are finished (check MSTATUS.DONE status flag).

- Clear all ECC error flags in the E-Ray module: these are flag EERR in register EIR, flags EIBF, EOBFB, EMR, ETBF1, ETBF2 in register MHDS. The flags are cleared by writing a '1' to the according bit position in the flag register.

After these steps the E-Ray RAM can be used for further operation, for example for initialization of the E-Ray buffer.

### Method 2 using "CLEAR RAMS" Command:

Step 1 to 4: Enable the clock of the module:

- 1. Remove EINIT protection for the writing of the CLC register.
- 2. Enable the clock in the CLC register.
- 3. Read the CLC register.
- 4. Enable the EINIT protection.

Enable the test mode, check if the state of the module is according to the expected settings and start clearing the RAMs.

- 5. Take care of the unlock sequence. See description of LCK.TMK and TEST1.WRTEN in User's Manual:
  - Test Mode Key: To set bit TEST1.WRTEN the write operation has to be directly preceded by two consecutive write accesses to the Test Mode Key.
  - If the write sequence is interrupted by other write accesses between the second write to the Test Mode Key and the write access to the TEST1 register, bit TEST1.WRTEN is not set to 1 and the sequence has to be repeated.  
First write:  $LCK.TMK = 75_H = 0111\ 0101_B$   
Second write:  $LCK.TMK = 8A_H = 1000\ 1010_B$   
Second write:  $TEST1.WRTEN = 1_B$
- 6. Check if CCSV.POCS is either 0x0 (DEFAULT\_CONFIG) or 0xF (CONFIG). If not in any of these states, perform the according command to get to CONFIG state.
- 7. Check if SUCC1.PBSY is equal 0x0. If 0x1 wait until 0x0.
- 8. Set SUCC1.CMD to 0xC meaning that the CLEAR\_RAMs command is entered.
- 9. Read SUCC1.CMD. If 0x0 the command has not been accepted. Repeat up from step 7. Otherwise continue.

- 10. Wait 1024 module cycles.
- 11. Enable RAM Test mode: TEST1.TMC = 01<sub>B</sub>. This mode enables access of all RAM blocks in E-Ray modules to the host.
- 12. CUST1.IBF1PAG := 1<sub>B</sub>
- 13. CUST1.IBF2PAG := 1<sub>B</sub>.
- 14. Repeat steps 7 to 10.
- 15. Read at least one address in all the RAM blocks within E-Ray module.
- 16. Switch off Test mode: TEST1.TMC = 00<sub>B</sub> and TEST1.WRTEN = 0<sub>B</sub>
- 17. Clear ECC error flags in MHDS and EIR registers
- 18. From here you can start the normal initialization process of the module.

*Note: In order to ensure proper FlexRay communication, RAM test mode must be explicitly disabled via TEST1.TMC = 00<sub>B</sub> in step 16 at the end of the initialization sequence.*

### **FlexRay\_TC.H004 Bit WRECC in register TEST2 has no function**

In the AURIX™ implementation of the E-Ray module, bit WRECC in register TEST2 has no function.

#### **Recommendation**

The value read from WRECC should not be evaluated by software, the value written (0<sub>B</sub> or 1<sub>B</sub>) to it is irrelevant.

For new software projects, keep bit WRECC at its reset value (0<sub>B</sub>) for easier migration to future AURIX™ generations.

### **FPI\_TC.H002 Write Access to Register ACCEN1**

The ACCEN1 (Access Enable Register 1) registers in the AURIX™ devices are reserved for future expansion. The bits in the ACCEN1 registers are described as “Reserved”, read-only. There is no need for software to configure (write to) the ACCEN1 registers.

*Note: For a write access to the ACCEN1 registers in the following modules, a bus error will be generated: MTU, SMU, ETH, I2C, FFT, CIF.*

**GPT12\_TC.H001 Timer T5 Run Bit T5R - Documentation Correction**

In the current version of the User's Manual, the lines for  $T5R=0_B$  and  $T5R=1_B$  in the register description of the Timer T5 Run Bit (T5R) erroneously have been swapped.

**Correction**

The correct behavior of bit T5R is as shown in **Table 23**:  $T5R=0_B$  (Timer T5 stops; default after reset),  $T5R=1_B$  (Timer T5 runs).

**Table 23 Timer T5 Control Register T5CON, Bit T5R - Correction**

Field	Bits	Type	Description
<b>T5R</b>	6	rw	<b>Timer T5 Run Bit</b> $0_B$ Timer T5 <b>stops</b> $1_B$ Timer T5 <b>runs</b> <i>Note: This bit only controls timer T5 if bit T5RC = 0.</i>

**GPT12\_TC.H002 Bits TxUD and TxUDE in incremental interface mode - Additional information**
**Description**

The present description of the incremental interface mode for timers T2, T3, T4 in the User's Manual, including figures and tables, implicitly refers to the following configuration of bits TxUD and TxUDE ( $x = 2, 3, 4$ ):

- TxUD =  $0_B$
- TxUDE =  $1_B$

This is the recommended and validated setting for these bits in incremental interface mode.

**Additional information**

When bit TxUD =  $1_B$ , the count direction of timer Tx is inverted compared to the setting with TxUD =  $0_B$  in incremental interface mode.

The setting of bit TxUDE is irrelevant in incremental interface mode, the behavior of Tx for TxUDE = 0<sub>B</sub> and TxUDE = 1<sub>B</sub> is identical. The figures related to incremental interface mode shall be interpreted as if TxUDE is permanently tied to 1<sub>B</sub>.

**GTM\_TC.H004 Correction to Bit Fields GTM\_TIMi\_IN\_SRC.VAL\_x**

In the description of bit field VAL\_0 in register GTM\_TIMi\_IN\_SRC in the User’s Manual, the encoding 01<sub>B</sub> was erroneously repeated while 10<sub>B</sub> and 11<sub>B</sub> were missing.

The correct description is included in the following **Table 24**. As the description of bit fields VAL\_x, x>0 refers to VAL\_0, this description is valid for all VAL\_x bit fields in register GTM\_TIM0\_IN\_SRC.

**Table 24 Corrected Description of Bit Field VAL\_0 in Register GTM\_TIM0\_IN\_SRC**

Field	Bits	Type	Description
VAL_0	[1:0]	rw	<b>Value to be fed to Channel 0</b> 00 <sub>B</sub> Input signal 0 (ignore write access) 01 <sub>B</sub> Input signal is set to 0 10 <sub>B</sub> Input signal is set to 1 11 <sub>B</sub> Input signal 1 (ignore write access) ...

**GTM\_TC.H005 External Capture in TIM Pulse Integration Mode (TPIM)**

In table “TIM integration Mode” in section “External Capture in TIM Pulse Integration Mode (TPIM)” of the GTM chapter in the User’s Manual, the information that CNT is cleared upon external capture is missing in column “Action description”.

The corrected **Table 25** is shown below:

**Table 25 TIM integration Mode**

Input signal F_OUTx	selected CMU clock	External capture	ISL	DSL	Action description
0	1	0	-	0	CNT++
1	1	0	-	0	no
1	1	0	-	1	CNT++
0	1	0	-	1	no
-	-	rising edge	-	-	do GPRx capture; issue NEWVAL_IRQ; <b>CNT = 0</b>
-	0	0	-	-	no

### **GTM\_TC.H007 GTM to CAN Timer Triggers**

The CAN transmit trigger inputs of the individual CAN nodes are connected to GTM trigger outputs as specified in table “CAN Transmit Trigger Inputs” in the MultiCAN+ chapter of the User’s Manual.

The corresponding GTM TOM/ATOM channel is selected in register GTM\_CANOUTSEL as specified in tables “CAN Timer Triggers” in the GTM chapter. Note that not all specified SELx bit fields in register CANOUTSEL are used for trigger selection.

The following GTM to CAN connections are implemented:

**Table 26 GTM to CAN Connections in TC23x**

CAN Node	GTM Trigger Selection via Bit Field
CAN Node 0	CANOUTSEL.SEL0
CAN Node 1	CANOUTSEL.SEL1
CAN Node 2	CANOUTSEL.SEL2
CAN1 Node 0	CANOUTSEL.SEL0
CAN1 Node 1	CANOUTSEL.SEL1
CAN1 Node 2	CANOUTSEL.SEL2

**GTM\_TC.H009 TIM0 Channel x Input Selection - Mapping for QFP-80 and QFP-100 Packages**

Basically, the mapping of TIM0 input channels to port pins follows a strict family concept: functions available in a lower pin-count package are located on the same port pin in the next higher pin-count package.

In tables “TIM 0 Mapping for QFP-80” and “TIM 0 Mapping for QFP-100” in chapter “Port to GTM Control Registers” of the GTM chapter in the User’s Manual, some rows are incorrect. The following **Table 27** and **Table 28** show the corresponding corrections.

*Note: Table “TIM 0 Mapping for QFP-144/BGA-292” in the User’s Manual is correct, as well as the tables in chapter “Port Connections” of the GTM chapter, and the GTM connections listed in the Data Sheet.*

**Recommendation**

For the correct port connections on QFP-80 and QFP-100 packages, use tables “GTM to Port Mapping for QFP-80” and “GTM to Port Mapping for QFP-100” in chapter “Port Connections” of the GTM chapter, or the Data Sheet.

**Corrections**

The following **Table 27** and **Table 28** show the **corrected** rows of tables “TIM 0 Mapping for QFP-80” and “TIM 0 Mapping for QFP-100”.

*Note: Connections for CHxSEL encodings not listed in **Table 27** or **Table 28** are correctly printed in the corresponding tables in the User’s Manual.*

**Table 27 Corrections to Table “TIM 0 Mapping for QFP-80”**

Field CHxSEL	QFP-80: Pad / Input	Name
<b>CH0SEL</b>		
0100 <sub>B</sub>	Reserved	-
0101 <sub>B</sub>	Reserved	-
0111 <sub>B</sub>	Reserved	TIN53
1000 <sub>B</sub>	Reserved	-

**Table 27 Corrections to Table “TIM 0 Mapping for QFP-80” (cont'd)**

<b>Field CHxSEL</b>	<b>QFP-80: Pad / Input</b>	<b>Name</b>
1011 <sub>B</sub>	P02.8	TIN8
1100 <sub>B</sub>	Reserved	-
<b>CH1SEL</b>		
0011 <sub>B</sub>	Reserved	-
0100 <sub>B</sub>	P14.6	TIN86
0101 <sub>B</sub>	Reserved	-
0110 <sub>B</sub>	Reserved	TIN54
1000 <sub>B</sub>	P33.5	TIN27
<b>CH2SEL</b>		
0001 <sub>B</sub>	Reserved	-
0011 <sub>B</sub>	Reserved	-
0100 <sub>B</sub>	P10.5	TIN107
0101 <sub>B</sub>	Reserved	-
0110 <sub>B</sub>	Reserved	-
1000 <sub>B</sub>	Reserved	-
1001 <sub>B</sub>	P33.6	TIN28
<b>CH3SEL</b>		
0001 <sub>B</sub>	Reserved	-
0011 <sub>B</sub>	Reserved	-
0100 <sub>B</sub>	P10.6	TIN108
0110 <sub>B</sub>	Reserved	-
1001 <sub>B</sub>	P33.7	TIN29
<b>CH4SEL</b>		
0010 <sub>B</sub>	Reserved	-
0100 <sub>B</sub>	Reserved	-

**Table 27 Corrections to Table “TIM 0 Mapping for QFP-80” (cont'd)**

Field CHxSEL	QFP-80: Pad / Input	Name
0101 <sub>B</sub>	Reserved	-
<b>CH5SEL</b>		
0011 <sub>B</sub>	Reserved	-
0100 <sub>B</sub>	Reserved	-
0110 <sub>B</sub>	Reserved	-
<b>CH6SEL</b>		
0100 <sub>B</sub>	P23.1	TIN42
0101 <sub>B</sub>	Reserved	-
0110 <sub>B</sub>	Reserved	-
<b>CH7SEL</b>		
0010 <sub>B</sub>	P14.4	TIN84
0011 <sub>B</sub>	P20.8	TIN64
0100 <sub>B</sub>	Reserved	-
0101 <sub>B</sub>	Reserved	-
0110 <sub>B</sub>	Reserved	-

**Table 28 Corrections to Table “TIM 0 Mapping for QFP-100”**

Field CHxSEL	QFP-100: Pad / Input	Name
<b>CH0SEL</b>		
0001 <sub>B</sub>	Reserved	-
1010 <sub>B</sub>	Reserved	-
1100 <sub>B</sub>	Reserved	-
<b>CH2SEL</b>		
1000 <sub>B</sub>	Reserved	-

**Table 28 Corrections to Table “TIM 0 Mapping for QFP-100” (cont’d)**

Field CHxSEL	QFP-100: Pad / Input	Name
<b>CH4SEL</b>		
1001 <sub>B</sub>	Reserved	-
1010 <sub>B</sub>	Reserved	-

**GTM\_TC.H011 First CM0 updates in case of SR0=1 and (A)TOM used as Triggered Channel**

In case the CM0 register should be updated from the shadow register with 1, the Force Update mechanism (FUPD(x) signal) has to be enabled on the (A)TOM channel. Otherwise the first edge triggered from CM0 will not be generated after 1 appears in CM0.

**GTM\_TC.H014 Synchronous Bridge Mode Restrictions**

The reset value for register GTM\_BRIDGE\_MODE is specified as 0400 1001<sub>H</sub>, and should never be changed according to the User’s Manual, i.e. the AEI bridge should always operate in async\_bridge mode.

**Exception**

In order to improve access latency, operation in synchronous bridge mode is possible if it is ensured that the SPB frequency is identical to the GTM frequency:

- $f_{SPB} == f_{GTM}$

Sequence to configure the bridge in synchronous mode (pseudocode):

```
/* ensure that no data are read or written in the GTM */
if(fSPB == fGTM)
{
GTM_BRIDGE_MODE = 0x04011000; /* switch to sync mode, reset
bridge*/
```

```

while (GTM_BRIDGE_MODE & 0x100) /* wait till mode change
completed */
;
}
else
;

```

### **GTM\_TC.H015 Register TIMi\_CHx\_CTRL - Correction to Register Image**

The register image of register TIMi\_CHx\_CTRL (i=0) erroneously shows bit 19 as “Reserved” with type “r” (read only).

#### **Correction**

Actually, bit 19 has type “rw” and is correctly described in the register table as copied from the User’s Manual in **Table 29** below:

**Table 29 Bit EXT\_CAP\_EN in Register TIMi\_CHx\_CTRL**

Field	Bits	Type	Description
EXT_CAP_EN	19	rw	<b>Enables external capture mode</b> The selected TIM mode is only sensitive to external capture pulses, the input event changes are ignored 0 <sub>B</sub> External capture disabled 1 <sub>B</sub> External capture enabled

### **GTM\_TC.H020 GTM can cause unintended bus errors after enabling when SPB or GTM frequency is very low**

When the SPB frequency is low compared to the CPU frequency, or the GTM frequency is low compared to the SPB frequency, the GTM can cause an FPI bus error when it is accessed too early after being enabled.

**Recommendation**

To avoid an FPI bus error, after enabling the GTM via the DISR bit in register CLC, a time delay of 10 SPB clock cycles and 10 GTM clock cycles must be inserted before accessing any GTM kernel register.

**GTM\_TC.H025 Field TOCTRL in register GTM\_TIM0\_CHx\_CTRL - Documentation correction**

In the GTM chapter of the TC21x/TC22x/TC23x User's Manual V1.1, the description of the edge selection in field TOCTRL of register GTM\_TIM0\_CHx\_CTRL is incorrect.

**Correction**

The correct encoding for field TOCTRL is shown below:

**Table 30 Encoding of field TOCTRL - Correction**

Field	Bits	Type	Description
TOCTRL	[31:30]	rw	<b>Timeout Control</b> 00 <sub>B</sub> Timeout feature disabled 01 <sub>B</sub> Timeout feature enabled for rising edge only 10 <sub>B</sub> Timeout feature enabled for falling edge only 11 <sub>B</sub> Timeout feature enabled for both edges

**INT\_TC.H004 Corrections to the Interrupt Router Documentation**

The following corrections apply to chapter "Interrupt Router (IR)" of the TC21x/TC22x/TX23x Family User's Manual:

Figure "Block Diagram of the TC21x/TC22x/TC23x Interrupt System" erroneously shows ICU3 related to DMA.

- **Correction:**
  - Only ICU0 and ICU1 are implemented, with **ICU1** related to the DMA.

Table “Registers Overview - System, OTGM and ICU Control Registers” erroneously shows ICU1 registers INT\_LWSR1, INT\_LASR1, INT\_ECR1 related to CPU1.

- **Correction:**

- Registers INT\_LWSR1, INT\_LASR1, INT\_ECR1 are related to the **DMA**.

### **IOM\_TC.H001 How to clear the IOM\_LAMEWCm register**

The Logic Analyzer Module Event Window Count Status register IOM\_LAMEWCm stores the window count value reached prior to being cleared in the LAM block once an event has been generated.

Writing to IOM\_LAMEWCm by software will result in a bus error.

The IOM\_LAMEWCm register can be reset (cleared) by software with a write to the IOM\_LAMCFGm or IOM\_LAMEWSm registers, e.g. by writing the same configuration data that have been read to either of these registers.

*Note: The clock divider should be set to IOM\_CLC.RMC = 1 when configuring the IOM (see issue IOM\_TC.004 “Write to IOM register space when IOM\_CLC.RMC > 1”).*

### **IOM\_TC.H002 IOM Clock Control**

Contrary to the named clocks given within the subsections of the IOM chapter, the entire IOM operates at the higher of the SPB or GTM clock frequencies. This may be further divided via the RMC bit field of the IOM\_CLC register, where the physical RMC value represents the divisor. For example, RMC = 00000001<sub>B</sub> divides clock by 1, RMC = 00000010<sub>B</sub> divides clock by 2, and so on. Note that RMC = 00000000<sub>B</sub> disables the clock.

See also the following revised description of the IOM\_CLC register.

#### **IOM Clock Control Register (IOM\_CLC)**

The Clock Control Register CLC allows the programmer to adapt the functionality and power consumption of the module to the requirements of the

application. The description below shows the clock control register functionality which is implemented in the BPI\_FPI for the module. Where a module kernel is connected to the CLC clock control interface, CLC controls the  $f_{IOM}$  module clock signal, sleep mode and disable mode for the module.

**Table 31 Description of Fields in IOM Clock Control Register (IOM\_CLC)**

Field	Bits	Type	Description
<b>DISR</b>	0	rw	<b>Module Disable Request Bit</b> Used for enable/disable control of the module. $0_B$ Module disable is not requested $1_B$ Module disable is requested
<b>DISS</b>	1	rh	<b>Module Disable Status Bit</b> Bit indicates the current status of the module. $0_B$ Module is enabled $1_B$ Module is disabled
<b>0</b>	2	rw	<b>Reserved</b> Read as 0; should be written with 0.
<b>EDIS</b>	3	rw	<b>Sleep Mode Enable Control</b> Used to control module's sleep mode. $0_B$ Sleep mode request is regarded. Module is enabled to go into Sleep Mode. $1_B$ Sleep mode request is disregarded. Sleep Mode cannot be entered upon a request.
<b>RMC</b>	[15:8]	rw	<b>Clock Divider Value in Run Mode</b> $00000000_B$ No clock signal $f_{IOM}$ generated (default after reset) $00000001_B$ Clock $f_{IOM} = \max(f_{SPB}, f_{GTM})$ selected $00000010_B$ Clock $f_{IOM} = \max(f_{SPB}, f_{GTM})/2$ selected $00000011_B$ Clock $f_{IOM} = \max(f_{SPB}, f_{GTM})/3$ selected ... $11111111_B$ Clock $f_{IOM} = \max(f_{SPB}, f_{GTM})/255$ selected
<b>0</b>	[31:16], [7:4]	r	<b>Reserved</b> Read as 0; should be written with 0.

### **IOM\_TC.H003 Configuration of LAMCFG.IVW and LAMEWS.THR**

As shown in figure “Logic Analyzer Module (LAM) block diagram” in the IOM chapter of the User’s Manual, an EVENT will be generated if the required edge is detected and the XOR between the Event Window value and the invert bit (LAMCFG.IVW) is 1.

When the edge to be detected arrives at LAMEWSn.THR value of the counter, the EVENT will be generated depending on LAMCFG.IVW value:

- If LAMCFG.IVW==0 event will be generated,
- if LAMCFG.IVW==1 event will not be generated.

Taking this behavior into account, the description of the LAMCFG.IVW and/or LAMEWS.THR configuration in examples 2, 4, 5 and 6 of section “Example Monitor/Safety Measures” is misleading.

#### **Correction**

The corrected description, including the case “equal to”, is as follows (only modified lines are printed):

#### **Example 2 - Pulse or duty cycle too long**

LAMCFG.IVW: 0x0 ; don’t invert window, capture events when the counter is **equal or** above the threshold.

LAMEWS.THR: select appropriate threshold (maximum duty cycle length required. If duty cycle is longer than this value then an event will be triggered).

#### **Example 4 - Period too long**

LAMCFG.IVW: 0x0 ; don’t invert window, capture events when the counter is **equal or** above the threshold.

LAMEWS.THR: select appropriate threshold (maximum period length required. If period is longer than this value then an event will be triggered).

#### **Example 5 - Diagnosis of Command and Feedback - acceptable propagation window and/or signal consistency check**

LAMCFG.IVW: 0x0 ; don’t invert window, capture events when the counter is **equal or** above the threshold.

LAMCFG.THR: set to max delay allowed (if the delay between corresponding edges of reference and monitor signals is longer than this value, the event will be triggered).

### **Example 6 - Diagnosis of Set-up and Hold times**

#### **- Example settings for LAM block registers for Set-up**

LAMCFG.IVW: 0x0 ; don't invert window, capture events when the counter is equal or above the threshold.

#### **- Example settings for LAM block registers for Hold**

LAMCFG.IVR: **0x1** ; invert reference signal (use for gating).

LAMCFG.THR: Acceptable Hold (ref Threshold 2 on waveforms shown, changes in monitor signal will generate an alarm if they occur inside the "THR" cycles after a falling edge in the reference signal).

### **IOM\_TC.H004 Behavior of LAMEWCn.CNT when LAMEWSn.THR is 0**

When LAMEWSn.THR is set to 0, no event will be sent from the Logic Analyzer Module (LAM) to the Event Combiner Module (ECM) and no ALARM towards the SMU will be generated.

The rest of the effects derived from the cause generating the event inside the LAM will be maintained, for instance copying the counter to LAMEWCn.CNT (this means LAMEWCn.CNT also may change when LAMEWSn.THR is 0).

### **IOM\_TC.H006 ACCEN\* Protection for Write Access to IOM Registers**

The access protection symbol 'P' to indicate protection by the ACCEN\* register mechanism is missing in column "Access Mode - Write" in table "Register Overview" in the User's Manual for IOM registers with an offset address  $\geq 30_H$ . Actually, these registers have write access attributes 'U,SV,P'.

### Exception

In this design step, a write access to register LAMEWCm will result in a bus error, as correctly reflected by symbol 'BE' in column "Access Mode - Write" in table "Register Overview" in the User's Manual.

### **IOM TC.H007 Write Access to FPCEsr**

The Filter and Prescaler Edge Status Register FPCEsr stores the state of detected rising and falling edges from each of the Filter and Prescaler Channels k (k = 0..15).

The flags in this register can be selectively cleared by writing a 0 in the respective bitfield.

However, writing to register FPCEsr with a sub-word granularity (e.g. byte or half-word) leads to undefined behavior.

### Recommendation

Individual bits for channel k in FPCEsr are cleared with a write to the control register (FPCCTRk) or timer register (FPCTIMk).

Writing to FPCEsr directly shall be done always to the whole register (32-bit writes), with bits that should not be modified set to 1<sub>B</sub>.

In particular, LDMST or SWAPMSK.W should be used only with bit mask enabled for all 'rwh' bits in register FPCEsr.

### **LBIST TC.H004 Update reset behavior of LBISTCTRL2 register - Additional information**

Even though the LBISTCTRL2.[31:0] register bits are cleared by a power-on reset they will automatically recover their values from stored contents of the central LBIST controller in the TCU (Test Control Unit) afterwards.

So on first software access the user will never see the initial reset values, but the updated LBIST done status and MISR result from the TCU LBIST controller.

The stored LBIST done status and MISR result in the central TCU LBIST controller will be cleared only through an externally applied warm power-on reset or during any cold power-on reset (triggered from EVR voltage monitors).

### **LMU\_TC.H002 On-the-fly BBB:SRI clock ratio switching**

*Note: This problem only occurs in an ADAS or Emulation Device (ED), but may already need to be considered during software development for the target device.*

When switching the clock ratio for  $f_{\text{BBB}}$  relative to  $f_{\text{SRI}}$ , make sure that no MMES (Memory Mapped Emulation System) access to EMEM is performed by an SRI master via the LMU. Otherwise, data read/written may be incorrect.

#### **Recommendation**

After a MMES read is complete, allow at least 12 SRI clock cycles before initiating a clock ratio change.

After a MMES write is complete, allow at least 20 SRI clock cycles before initiating a clock ratio change.

After a clock ratio change, allow the clock ratio change to become effective before performing any MMES transfer (e.g. read back control register that was written for the clock ratio change).

### **LMU\_TC.H003 Function of Bit MEMCON.PMIC (Protection Bit for Memory Integrity Control Bit)**

In the LMU chapter of the User's Manual, the following text (last paragraph in section "Local Memory (LMU SRAM)") is incorrect: Some bitfields of the LMU\_MEMCON register are protected by LMU\_MEMCON.PMIC bit. If the data written to the register has the bitfield set to  $0_{\text{B}}$ , no change will be made to bits  $15_{\text{D}}$  to  $9_{\text{D}}$  of the register regardless of the data written to these fields.

### Correct Description

For the correct description (only bit 9 (ERRDIS) is protected) see the description of bit PMIC in the LMU Memory Control Register in section “LMU Registers”, copied in **Table 32** below:

**Table 32 Bit PMIC in Register LMU\_MEMCON**

Field	Bit	Type	Description
<b>PMIC</b>	8	w	<b>Protection Bit for Memory Integrity Control Bit</b> Will always return 0 <sub>B</sub> when read 0 <sub>B</sub> Bit Protection: Bit 9 remains unchanged after LMU_MEMCON write. 1 <sub>B</sub> Bit 9 will be updated by the current write to LMU_MEMCON
<b>ERRDIS</b>	9	rw	<b>ECC Error Disable</b> When set SRI bus errors caused by ECC errors in data read from the SRAM will be disabled. ...

### MTU\_TC.H003 AURIX™ Memory Tests using the MTU

The use of destructive tests such as March-U and Checkerboard etc. in conjunction with FAILDMP mode to get detailed failure information (errors, fail addresses) will cause the SRAM redundancy information to be overwritten.

Therefore, the MTU/MBIST module effectively only supports the Non-Destructive Inversion Test (NDIT).

### Recommendation

To avoid overwriting the SRAM redundancy information, only use Non-Destructive Inversion Test. In this case, failure is detected by ECC and the detailed information can be obtained from ETRR and ECCD registers.

Refer to the latest version of Application Note AP32197 “AURIX™ Memory Tests using the MTU” for more details on MTU/MBIST usage and fault coverage.

## **MTU\_TC.H004 Handling the Error Tracking Registers ETRR**

CPU and on-chip peripheral SRAMs are capable of detecting errors and generating SMU alarms for correctable, uncorrectable, and address errors. The failing addresses are stored in Error Tracking Registers (ETRR), and the corresponding indicator (CERR/UERR/AERR and SERR) and valid bits (VAL) are set in the Memory ECC Detection Register (ECCD). Only new errors will be considered, i.e. errors at already stored addresses will be ignored. In case the maximum number of ETRR for a memory is used up and a new error occurs, the error overflow bit ECCD.EOV is set, and the corresponding “address buffer overflow” SMU alarm is generated. For peripheral SRAMs, the second error will cause a buffer overflow, and for CPU SRAMs, up to five errors can be registered before the buffer overflow alarm is triggered.

Bit ECCD.TRC (Tracking Clear) allows to clear the EOVS and VAL bits in register ECCD and the associated ETRR registers, e.g. in response to a tolerated corrected single bit error.

### **Corner Case**

If in an exceptional corner case software would set TRC at the same time an error overflow occurs, then the EOVS bit is not set, and the SMU alarm is not generated.

### **Recommendation**

- It is not necessary to clear the Error Tracking Registers ETRR by software as part of an SRAM error handling concept. For correctable errors, the application software should only react on the address buffer overflow alarm (e.g. with a reset). Single correctable error events may be ignored (within limits) to increase the fault tolerance of the system without impacting the safety.
- If a different concept is used requiring clearing of the ETRR registers by software via ECCD.TRC, make sure that the corresponding SRAM instance is not functionally accessed while the application software writes ECCD.TRC, so that an overflow error cannot be generated during the clear operation.

Information on using the MTU for memory diagnosis is given in Application Note AP32197 “AURIX™ Memory Tests using the MTU”.

## **MTU\_TC.H005 Handling SRAM Alarms**

Alarms are generated for CPU and on-chip peripheral SRAMs when correctable, uncorrectable, and address errors are detected.

The failing addresses are stored in Error Tracking Registers (ETRR), and information on the error type is stored in the Memory ECC Detection Register (ECCD). Only new errors will be considered, i.e. errors at already stored addresses will be ignored. In case the maximum number of ETRR for a memory is used up and a new error occurs, the error overflow bit ECCD.EOV is set, and the corresponding “address buffer overflow” SMU alarm is generated.

For peripheral SRAMs, the second error will cause a buffer overflow, and for CPU SRAMs, up to five errors can be registered before the buffer overflow alarm is triggered.

In addition, traps and bus errors are generated for uncorrectable errors, depending on the bus master and type of access.

### **Corner Case**

If in an exceptional corner case

- two errors at different locations are present in the same SRAM
- and accesses are made to both locations within a time window of ~ 10 CPU clock cycles,

then the first access to the location with an error will correctly trigger an SMU alarm, while the second access to the other location with an error will not trigger an SMU alarm. In the worst case, a correctable error may thus mask an uncorrectable or address error.

*Note: In case the second error would result in an address buffer overflow, the corresponding bit ECCD.EOV is set and the “address buffer overflow” SMU alarm is correctly generated.*

*Therefore, this problem is **not** relevant for peripheral SRAMs that only have one ETRR, as the second error will always cause an SMU alarm.*

## Recommendations

- As recommended in Application Hint MTU\_TC.H004 (Handling the Error Tracking Registers ETRR), for correctable errors, the application software should only react on the address buffer overflow alarm (e.g. with a reset). Single correctable error events may be ignored (within limits) to increase the fault tolerance of the system without impacting the safety.
- In case an uncorrectable error for a CPU SRAM would neither generate an “address buffer overflow” nor an “uncorrectable” or “address error” SMU alarm, the error handling (typically resulting in a reset) should be performed in the corresponding trap routine.
- In particular for EMEM or FFT SRAMs used in Emulation, ADAS or Extended SRAM devices of the AURIX™ family, a workaround is possible by triggering a correctable error before application startup. This would result in the ECCD.CERR bit of the corresponding MBIST to be set. Any future correctable alarms will not be forwarded<sup>1)</sup> and this issue can be avoided completely.

### **MTU\_TC.H006 Alarm Propagation to SMU via Error Flags in MCx\_ECCD**

Upon any correctable, un-correctable or address error alarm in an SRAM, the corresponding error flags (CERR, UERR or AERR bits) in the MCx\_ECCD register are set, and the corresponding alarm is forwarded to the SMU.

However, in case these bits are set to 1<sub>B</sub>, and a further error of the same type occurs, then the corresponding alarm is no longer forwarded to the SMU.

If in a corner case software writes to Mx\_ECCD in the same cycle where an error event would set one of the CERR, UERR or AERR bits from 0<sub>B</sub> to 1<sub>B</sub>, the software write has priority and the status flags remain at 0<sub>B</sub>. In this case, however, the alarm is correctly propagated to the SMU.

*Note: This behavior does not endanger the concept recommended in Application Hints MTU\_TC.H004 and MTU\_TC.H005 (ignore correctable errors, react on first uncorrectable/address error/buffer overflow alarm).*

---

1) see MTU\_TC.H006 (Alarm Propagation to SMU via Error Flags in MCx\_ECCD)

## Recommendation

Upon any alarm from an SRAM/MBIST, if a further alarm of the same type is required to be sent to the SMU and processed, then the software shall clear the error flag (CERR, UERR, AERR) in the ECCD register.

The flags can be cleared by writing MCx\_ECCD.CERR (or UERR or AERR, respectively) with 0<sub>B</sub>.

## **MTU\_TC.H008 Memory Controllers for DSPR**

Due to its implementation, the Data Scratch Pad RAM (DSPR) of CPU0 has two Memory Controller instances, described as MC14 (MC\_CPU0\_DSPR) and MC27 (MC\_CPU0\_DSPR2) in the MTU chapter of the User's Manual.

Each Memory Controller covers one half of the SRAM. In order to fully test the DSPR, the test<sup>1)</sup> has to be executed once on each Memory Controller (i.e. only one of the two Memory Controllers is enabled at a time).

As both Memory Controllers share the same ECC decoders, any error detected by a test executed on one of the Memory Controllers will be logged in the Error Tracking Registers of both Memory Controllers.

Note that once an error status bit is set, further alarms of the same type are not forwarded to the SMU until the flag is cleared (see MTU\_TC.H006 "Alarm Propagation to SMU via Error Flags in MCx\_ECCD").

## Recommendation

- Enable only one of the two Memory Controllers at a time.
- Before executing a test on CPUx\_DSPR (respectively CPUx\_DSPR2), clear the Error Tracking Registers and the error status bits of CPUx\_DSPR (respectively CPUx\_DSPR2), so that the test reflects the results of only the one memory which is being tested.
- It is also recommended to clear the Error Tracking Registers and the error status bits of both CPUx\_DSPR and CPUx\_DSPR2 after executing each test.

---

1) Test in this context means Non-Destructive Inversion Test (NDIT, see also MTU\_TC.H003).

- Alternatively, before executing a test on CPUx\_DSPR (respectively CPUx\_DSPR2), disable the error notifications in CPUx\_DSPR2 (respectively CPUx\_DSPR) and reenable them after the test. It is also recommended to clear the Error Tracking Registers and the error status bits of CPUx\_DSPR (respectively CPUx\_DSPR2) after executing a test on it.

Regarding configuration and use of the two Memory Controllers, see also the latest version of Application Note AP32197 “AURIX™ Memory Tests using the MTU”.

### **MTU\_TC.H009 Reset Value for Register ECCD**

The reset value of the ECC Detection Register ECCD is documented as 7800<sub>H</sub> in the User’s Manual. This is always the case for the SRAMs listed in [Table 33](#) below (if available in the corresponding product).

**Table 33 TC23x SRAMs with ECCD Reset Value = 7800<sub>H</sub>**

<b>Memory Controller No.</b>	<b>Associated SRAM</b>
17	CPU0 PTAG
38	ERAY0 OBF
39	ERAY0 IBF_TBF

For other SRAMs the ECCD reset value may either be 7C00<sub>H</sub> or 7800<sub>H</sub>.

Bit ECCD.10 is marked as ‘Reserved’ in the User’s Manual:

- When writing to ECCD, bit ECCD.10 should be written as 0<sub>B</sub>.
- When reading register ECCD, bit ECCD.10 should not be evaluated. Memory errors will be reported by the notification bits CERR, UERR, AERR and EOv in register ECCD.

**MTU\_TC.H010 Register MCONTROL - Bit Field Res4**

The position of the 3-bit field Res4 within register MCONTROL is incorrectly described as [14:10] in the register description of the User's Manual.

The correct position of the 3-bit field Res4 is MCONTROL.[14:12], as shown in the register image in the User's Manual, and in the following [Table 34](#):

**Table 34 Register MCONTROL - Position of Bit Field Res4**

Field	Bits	Type	Description
<b>Res</b>	15	r	<b>Reserved</b> Read returns 0 <sub>B</sub> , should be written with 0 <sub>B</sub>
<b>Res4</b>	14:12	rw	<b>Reserved</b> Read returns 0x4 Must always be written with 0x4
<b>Res</b>	11:10	r	<b>Reserved</b> Read returns 00 <sub>B</sub> , should be written with 00 <sub>B</sub>

**MTU\_TC.H011 Access Protection for Memory Control Registers**

The access protection symbol 'P' to indicate Access Enable Register protection is missing in column "Access Mode - Write" in table "Register Overview of each MTU Memory Control register block" of the MTU chapter in the User's Manual.

The MTU Memory Control register block actually has protection via the Access Enable registers (ACCEN0/1).

**MTU\_TC.H012 Kernel Reset triggers Reset of MBIST Registers**

When a kernel reset is executed (via bit RST in registers KRST0/1) for a module equipped with Memory Controllers (MC) for its internal RAMs, also the corresponding MTU Memory Control (MBIST) registers are reset.

## Recommendation

If required, analyze/save the contents of the MBIST registers before executing a kernel reset.

After a kernel reset, reconfigure the MBIST registers.

### **MTU\_TC.H014 Access to SRAM while MTU operations are underway**

When MTU operations on the SRAM are underway, the memories cannot be accessed. MTU operations in this context include:

1. Running an MBIST test (e.g. Non-destructive test).
2. Performing an SRAM initialization using the MTU.
3. When an Auto-data-initialization is underway.

During these operations, the SRAM shall not be accessed. If the SRAM is accessed during this time, unexpected behavior may occur (e.g. access timeout).

Cases 1. and 2. are easily identified, i.e. whenever the application has triggered an MBIST test or SRAM initialization.

Case 3. occurs whenever bit field PROCOND.RAMIN is not equal to 0x3. Whenever this is the case in specific MBIST controllers, the SRAM is fully or partially cleared under certain conditions:

- When MTU\_MEMTEST.\*EN bit is enabled or disabled.
- When MTU\_MEMMAP.\*MAP bit is set or cleared (applicable only to cache memories).

This means, when the above mentioned bits are set or cleared, it takes some time (~hundreds of clock cycles) for the associated SRAMs to be (fully or partially) initialized. During this time the SRAM is not accessible.

Affected SRAMs are:

- CPUx DMEM (DSPR+DCACHE)
- CPUx PMEM (PSPR + PCACHE)

## Recommendation

- For all memories, ensure that the SRAM is not accessed when any MTU operation is underway.

- For the specific memories listed above, ensure that the SRAM is not accessed:
  - When setting MTU\_MEMTEST.\*EN bit: as long as MEMSTAT.\*AIU bit is set or as long as the MEMTEST.\*EN bit is not yet set.
  - When clearing MTU\_MEMTEST.\*EN bit: as long as MEMSTAT.\*AIU bit is set or as long as the MEMTEST.\*EN bit is not yet cleared.
  - When setting or clearing MTU\_MEMMAP.\*MAP bit for DMEM/PMEM: as long as MEMSTAT.\*AIU bit is set.

### **MultiCAN AI.H005 TxD Pulse upon short disable request**

If a CAN disable request is set and then canceled in a very short time (one bit time or less) then a dominant transmit pulse may be generated by MultiCAN module, even if the CAN bus is in the idle state.

Example for setup of the CAN disable request:

`CAN_CLC.DISR = 1` and then `CAN_CLC.DISR = 0`

#### **Workaround**

Set all INIT bits to 1 before requesting module disable.

### **MultiCAN AI.H006 Time stamp influenced by resynchronization**

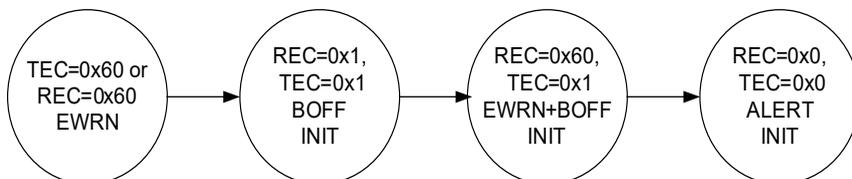
The time stamp measurement feature is not based on an absolute time measurement, but on actual CAN bit times which are subject to the CAN resynchronization during CAN bus operation. The time stamp value merely indicates the number of elapsed actual bit times. Those actual bit times can be shorter or longer than nominal bit time length due to the CAN resynchronization events.

#### **Workaround**

None.

### **MultiCAN\_AI.H007 Alert Interrupt Behavior in case of Bus-Off**

The MultiCAN module shows the following behavior in case of a bus-off status:



**Figure 8 Alert Interrupt Behavior in case of Bus-Off**

When the threshold for error warning (EWRN) is reached (default value of Error Warning Level EWRN = 0x60), then the EWRN interrupt is issued. The bus-off (BOFF) status is reached if  $TEC > 255$  according to CAN specification, changing the MultiCAN module with REC and TEC to the same value 0x1, setting the INIT bit to 1<sub>B</sub>, and issuing the BOFF interrupt. The bus-off recovery phase starts automatically. Every time an idle time is seen, REC is incremented. If REC = 0x60, a combined status EWRN+BOFF is reached. The corresponding interrupt can also be seen as a pre-warning interrupt, that the bus-off recovery phase will be finished soon. When the bus-off recovery phase has finished (128 times idle time have been seen on the bus), EWRN and BOFF are cleared, the ALERT interrupt bit is set and the INIT bit is still set.

### **MultiCAN\_TC.H003 Message may be discarded before transmission in STT mode**

If  $MOFCR_n.STT=1$  (Single Transmit Trial enabled), bit TXRQ is cleared (TXRQ=0) as soon as the message object has been selected for transmission and, in case of error, no retransmission takes places.

Therefore, if the error occurs between the selection for transmission and the real start of frame transmission, the message is actually never sent.

#### **Workaround**

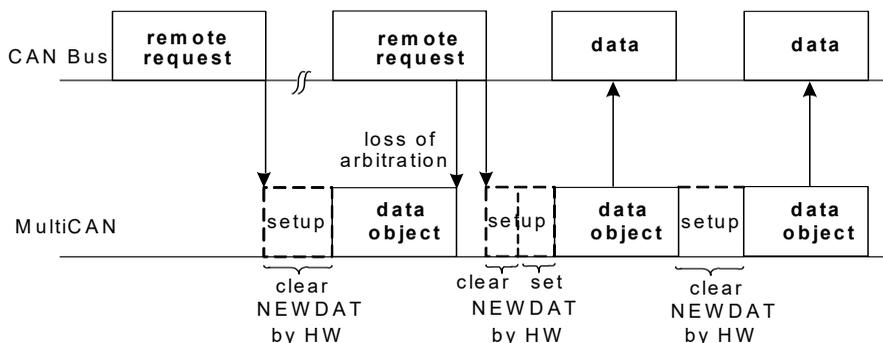
In case the transmission shall be guaranteed, it is not suitable to use the STT mode. In this case,  $MOFCR_n.STT$  shall be 0.

**MultiCAN\_TC.H004 Double remote request**

Assume the following scenario: A first remote frame (dedicated to a message object) has been received. It performs a transmit setup (TXRQ is set) with clearing NEWDAT. MultiCAN starts to send the receiver message object (data frame), but loses arbitration against a second remote request received by the same message object (NEWDAT will be set).

When the appropriate message object (data frame) triggered by the first remote frame wins the arbitration, it will be sent out and NEWDAT is not reset. This leads to an additional data frame, that will be sent by this message object (clearing NEWDAT).

There will, however, not be more data frames than there are corresponding remote requests.



**Figure 9 Loss of Arbitration**

**MultiCAN\_TC.H007 Oscillating CAN Bus may Disable the CAN Interface**

If the connected CAN network is in an unspecified oscillating state for more than 512 cycles this can result in disabling the CAN interface of the device. Enabling the CAN interface again requires then a Power-on Reset.

## Recommendation

Please refer to application note AP32264 “DXCPL DAP over CAN Physical Layer” for further information and how this situation can be prevented.

## MultiCAN TC.H008 Changes due to CAN FD protocol ISO 11898-1:2015

*Note: This Application Hint might affect the SFR C Header Definitions. In such cases, SFR usage in the software shall be analyzed within the applications for their correct handling.*

## Introduction

Specific variants of this device step support the CAN FD frame format according to standard version ISO 11898-1:2015. These variants are identified by the feature type code `N` as last letter in the device name, e.g.

- SAK-TC234LP-32F200N

*Note: In TC23x variants with feature type code `N`, nodes 0 and 1 in MultiCAN and MultiCAN1 support this feature, while nodes 2 don't; see [Table 38](#) at the end of this text module.*

For availability of the variants with this feature see the corresponding “AURIX™ TC2xx Variants / Data Sheet Addendum”.

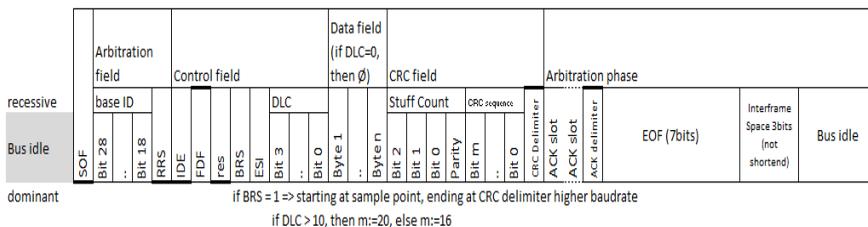
## Detailed Description

ISO 11898-1:2015 improves the failure detection capabilities of the ISO11898-1 DIS version 2014. Information about the number of stuff bits in the data field is added to the CRC field. These added bits are called ‘**Stuff Count**’.

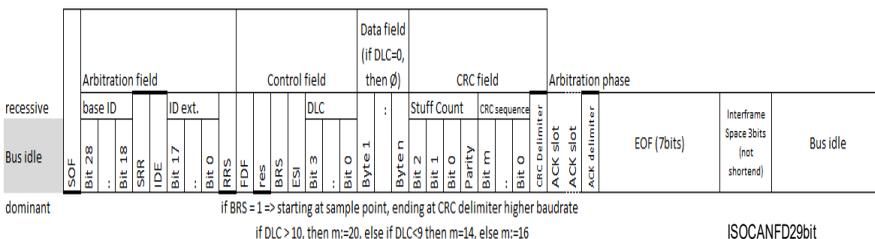
The Stuff Count contains 4 bits, including

- 3 bits gray code to represent the modulo-8 of number of stuff bits in the data field,
- and 1 bit for the parity.

Since the Stuff Count bits are part of the CRC field, fixed stuff bits will be added before and after the Stuff Count bits. [Figure 10](#) and [Figure 11](#) show the frame format of the ISO 11898-1:2015 CAN FD protocol. There is no change in the classical CAN frame format.



**Figure 10 ISO CAN FD 11-bit ID Data Frames**



**Figure 11 ISO CAN FD 29-bit ID Data Frames**

From here on,

- the ISO 11898-1:2015 frame format will be referred to as **ISO CAN FD** format,
- the previous frame format will be referred to as **Non-ISO CAN FD** format.

*Note: The ISO CAN FD frame format is incompatible with Non-ISO CAN FD frame format.*

AURIX™ devices (with feature type code `N`) support both ISO and Non-ISO CAN FD formats. The format can be selected by modified functionality of bits NBTR0.15 and NBTR1.15:

### Functionality of Bit NBTR0.15

NBTR0.15 is changed from NBTR0.DIV8 (Divide Prescaler Clock by 8) to **NBTR0.NISO**<sup>1)</sup> (Non-ISO operation) as shown in [Table 35](#):

**Table 35    Functionality of Bit NBTR0.15**

Field	Bit	Type	Description
NISO	15	rw	<b>Non-ISO Operation</b> If this bit is set, the MultiCAN+ uses the non-ISO CAN FD frame format. This bit is CCE protected. 0 <sub>B</sub> CAN FD frame format according to ISO 11898-1:2015 (default after reset) 1 <sub>B</sub> CAN FD frame format non-ISO.

**Functionality of Bit NBTR1.15**

NBTR1.15 is changed from NBTR1.DIV8 (Divide Prescaler Clock by 8) to **NBTR1.PED**<sup>1)</sup> (Protocol Exception Disable) as shown in [Table 36](#):

**Table 36    Functionality of Bit NBTR1.15**

Field	Bit	Type	Description
PED	15	rw	<b>Protocol Exception Disable</b> The protocol exception event is described in the ISO 11898-1:2015 as option. The error frame on the res bit can be controlled with this option. This bit is CCE protected. 0 <sub>B</sub> Protocol Exception Event is enabled (default after reset). 1 <sub>B</sub> Protocol Exception Event is disabled.

*Note: Both NBTR0.NISO and NBTR1.PED are global register bits. This means they affect all the ISO 11898-1:2015 compliant CAN FD nodes in the respective MultiCAN+ module.*

*The former DIV8 function of nodes 0 and 1 is hard-wired to 0<sub>B</sub> (i.e. a time quantum lasts (BRP+1) clock cycles).*

1) The symbolic names NISO and PED are only used for explanation in this context. If desired, the register definition file could be modified.

The DIV8 function (Divide Prescaler Clock by 8) for all other nodes  $x$  ( $x > 1$ ) remains the same, irrespective of the setting of NBTR0.NISO and NBTR1.PED.

**Table 37** describes the CAN FD behavior for different configurations of the NBTR0.NISO and NBTR1.PED bits. By default, the CAN FD behaves in compliance with ISO 11898-1:2015 if CAN FD is enabled (bit FDEN =  $1_B$  for corresponding node).

**Table 37 Configurations of PED and NISO**

PED	NISO	CAN FD Enabled
0	0	Default values - ISO 11898-1:2015 CAN FD compliant
0	1	Non-ISO CAN FD format - same behavior as previous AURIX™ devices
1	0	CAN FD with protocol exception event disabled - ISO 11898-1:2015 CAN FD compliant
1	1	Reserved

*Note: Nodes where  $FDEN = 0_B$  will operate using the classical CAN frame format.*

### Summary of Devices and Nodes supporting ISO CAN FD

The following table summarizes the nodes of devices with feature type code `N` which have the ISO 11898-1:2015 CAN FD functionality.

**Table 38 AURIX™ TC23x Devices/Nodes supporting ISO CAN FD**

Device / Step	ISO CAN FD supporting nodes	Non-ISO CAN FD supporting nodes
TC23x <sup>1)</sup> ≥ AC	MultiCAN - Nodes 0,1 MultiCAN1 - Nodes 0,1	MultiCAN - Node 2 MultiCAN1 - Node 2

1) Emulation (ED) and ADAS Devices only support Non-ISO CAN FD

## **MultiCAN\_TC.H009 Limitation on Secondary Sample Point (SSP) Position (ISO CAN FD nodes only)**

*Note: This Application Hint only applies to ISO CAN FD nodes. For devices and nodes supporting the ISO CAN FD format, see MultiCAN\_TC.H008.*

The MultiCAN+ of AURIX™ TC2xx has passed the ISO/DIS 16845-1(E), 2015 CAN Conformance test performed by an external test house C&S group GmbH and the test reports are available. The limitation on the range of SSP position is described in the Conformance test report.

In AURIX™ TC2xx devices, there are two limitations with the Secondary Sample Point (SSP) position for CAN FD with respect to ISO 11898-1, 2015 specification:

### **1. Granularity of the Transmitter loop delay measurement (only when CAN\_FNBTRx.FBRP = 1)**

#### **Limitation**

The Transmitter loop delay measurement is based on data-phase time quantum ( $t_{q(D)}$ ) and not by minimum time quanta (mtq) or CAN clock period as specified in ISO 11898-1 2015. Hence the granularity of the transmitter loop delay measurement is  $+1 t_{q(D)}$  in worst case scenario.

*Note: According to ISO 11898-1 – 2015, when Transmitter Delay Compensation is enabled (CAN\_NTDCR.TDC = 1), then the CAN\_FNBTRx.FBRP shall be either 0 or 1.*

#### **Effect**

In worst case scenario, the SSP could be delayed by  $+1 t_{q(D)}$ .

#### **Recommendation**

It has to be taken care that the SSP offset (CAN\_NTDCR.TDCO) is configured accordingly by including the granularity of the transmitter loop delay measurement of  $+1 t_{q(D)}$  in worst case scenario.

## 2. Range of SSP position (only when CAN\_FNBTRx.FBRP = 0)

### Limitation

The Secondary Sample Point Position is limited to  $31 t_q$  or  $31 mtq$  (bit field CAN\_NTDCRx.TDCV), when compared to  $63 mtq$  as required by ISO 11898-1, 2015.

*Note: When CAN\_FNBTRx.FBRP = 0, then*

*1 time-quantum ( $t_q$ ) = 1 minimum time-quantum ( $mtq$ ).*

CAN FD applications with fast data baud rate greater than 2 Mbit/s require Fast Baud Rate Prescaler setting CAN\_FNBTRx.FBRP = 0 and  $f_{CAN}$  at 80 MHz to ensure reliable CAN communication in long networks. In such a scenario, the max SSP position achievable by the TDC is limited to  $31 t_q$ , i.e. 388 ns ( $31 * 12.5$  ns).

### Effect

In scenarios where the sum of transmitter loop delay and SSP offset (CAN\_NTDCRx.TDCO) is more than 31 time quanta, the SSP value saturates at 31 time quanta, leading to SSP placed (at 31 time quanta) earlier than required.

### Recommendation

It has to be taken care to ensure that the sum of transmitter loop delay and SSP offset (CAN\_NTDCRx.TDCO) is within the limit of 31 time quanta.

### **MultiCAN\_TC.H010 Limitation on maximum SJW Range for CAN FD Data Phase (ISO CAN FD nodes only)**

*Note: This Application Hint only applies to ISO CAN FD nodes. For devices and nodes supporting the ISO CAN FD format, see MultiCAN\_TC.H008.*

The MultiCAN+ of AURIX™ TC2xx has passed the ISO/DIS 16845-1(E), 2015 CAN Conformance test performed by an external test house C&S group GmbH and the test reports are available.

ISO 11898-1, 2015 specifies the configuration range of the CAN FD Data phase (re-)synchronization jump width (SJW) as  $1-8 t_{q(D)}$ .

In AURIX™ TC23x devices, the CAN FD Data phase SJW is limited to  $1-4 t_{q(D)}$ , as bit fields CAN\_FNBTRx.FSJW and CAN1\_FNBTRy.FSJW are 2 bits wide.

### Effect

Configuring a MultiCAN+ node for CAN FD communication with CAN FD Data Phase SJW less than required, could result in wrong sampling of the received bit of CAN FD Data Phase, thus causing a Receive Error.

### Recommendation

Choose the CAN FD configuration in such a way that

- The period of time-quanta in Arbitration phase is equal to the period of time-quanta in data phase. This can be achieved by configuring  
CAN\_NBTEVRx.BRP = CAN\_FNBTRx.FBRP.
- CAN\_FNBTRx.FSJW =  $\min(\text{CAN\_FNBTRx.TSEG2}, 3)$

By this configuration the effect of limited Data SJW range offered by MultiCAN+ on maximum oscillator tolerance required (as given by conditions described in ISO 11898-1) is minimized.

### **MultiCAN\_TC.H011 Transmitter Delay Compensation Behaviour (CAN FD only)**

When using Transmitter Delay Compensation consider the following points:

1. The transmitter delay compensation does not take the Fractional Divider into account. This means that the values of CAN\_NTDCR.TDCO and CAN\_NTDCR.TDCV always correspond to CAN\_FDR.DM =  $01_B$  and CAN\_FDR.STEP = 1023, even though a different setting of the fractional divider is actually in place.

Therefore, it is recommended to use setting DM =  $01_B$  and STEP = 1023 in register CAN\_FDR so that the granularity of the transmitter loop delay measurement is depending only on the fast baud rate prescaler (CAN\_FNBTRx.FBRP).

2. If  $2 \cdot f_{\text{CAN}} < f_{\text{CLC}}$ , then the transmitter delay compensation measurement value of the previous measurement may be uploaded to bitfield CAN\_NTDCR.TDCV instead of the measured delay of the current message, i.e. the measured delay will appear in bitfield CAN\_NTDCR.TDCV with a delay of one CAN message.

### **MultiCAN\_TC.H012 Delayed time triggered transmission of frames**

The value written in the bit-field RELOAD of register NTATTRx(x=0-3), NTBTRx(x=0-3), NTCTTRx(x=0-3) represents the reload counter value for the timer used for triggered transmission of message objects (Classical CAN or CAN FD frames).

The timer source and the prescaler value is defined in the NTCCRx(x=0-3) register.

Once a value is written to bit-field RELOAD with bit STRT=1 the timer starts counting. This timer counts one value more than the written value in bit-field RELOAD, then it triggers the transmission of a message object.

#### **Effect**

The message object transmission is delayed by one counter cycle with respect to the desired count time written in bit-field RELOAD.

#### **Recommendation**

In order to transmit a message object at a specific time, when using one of these registers:

- NTATTRx(x=0-3), NTBTRx(x=0-3), NTCTTRx(x=0-3),  
set bit-field RELOAD one value less than the calculated counter value.

### **OCDS\_TC.H010 JTAG requires two initial clock cycles after PORST**

For a proper selection of the chip internal TCK clock path, two TCK clock cycles are needed after PORST release. They can be executed with TMS Low or High. A following TCK clock cycle with TMS High will always bring the JTAG TAP

state to Run-Test/Idle. This sequence is compliant to standard JTAG and can be used for all TriCore devices.

### **OCDS TC.H012 Minimum Hold Time for Inputs OCDS\_TGlx**

Inputs OCDS\_TGlx ( $x=0..7$ , depending on device/package type) may be used to trigger the On-Chip Debug System (OCDS) e.g. for break or interrupt from an external source.

To ensure the external trigger is sampled correctly and not missed, the trigger should be asserted for a minimum of two SPB clock cycles.

### **OCDS TC.H019 System or Application Reset while OCDS and lockstep monitoring are enabled**

After a System or Application Reset the Lockstep Alarm ALMx[0] gets activated if all of the following conditions are met ( $x$  = index of CPU with checker core):

1. Lockstep monitoring is enabled by BMI.LCLxLSEN =  $1_B$  for CPUx, AND
2. Debug System is enabled (CBS\_OSTATE.OEN =  $1_B$ ), AND
3. CPUx Performance Counters are enabled, AND
4. CPUx Clock Cycle Count register CCNT is read.

### **Recommendation**

To avoid the unintended ALMx[0] under the conditions described above, either:

- Keep the debug system disabled. OR
- Ensure CPUx Performance Counters are disabled for all CPUs that have lockstep monitoring enabled before executing a System or Application reset. OR
- Use PORST instead of a System or Application reset.

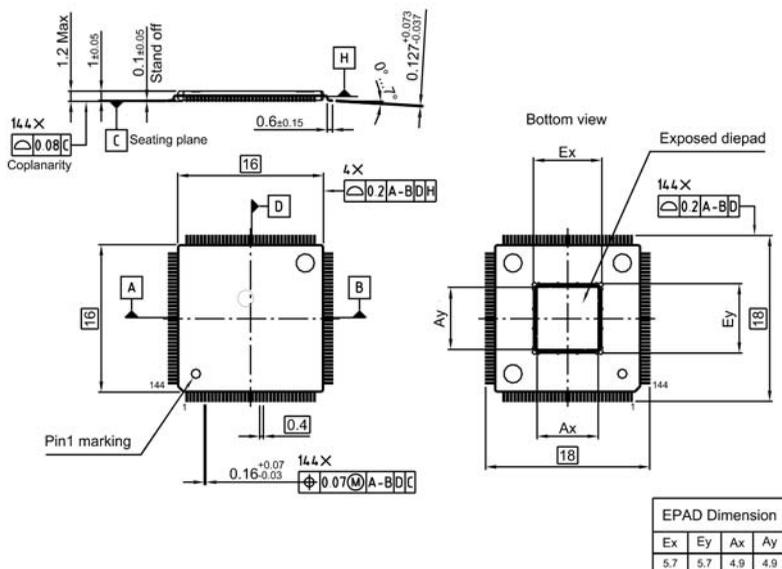
**PACKAGE\_TC.H008 Exposed pad dimensions and package outlines for QFP packages - Updates to TC23x Data Sheet**

In the scope of the harmonization of the package drawings, the drawings for the TQFP packages of the TC23x have been updated. No change of form, fit or function is implied.

The dimensions for the exposed pads are included in the respective figures. Furthermore, for the exposed pads, the maximum boundary of the structural corner protrusions to be considered during system design and integration has been added.

This information shall substitute the corresponding information in the TC23x AC-step Data Sheet V1.0 and in the TC23x A-step Data Sheet V1.1.

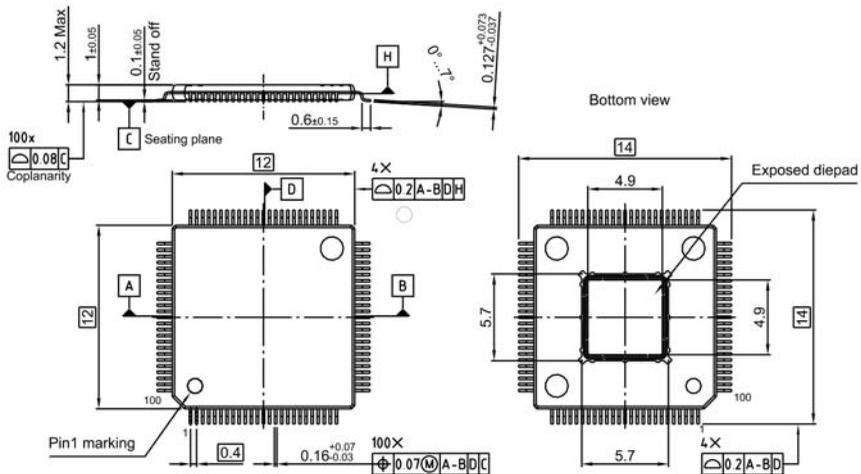
**Package Outlines TQFP-144 for TC23x**



**Figure 12 Package Outlines TQFP-144 for TC23x**

*Note: For the exposed pad of the TQFP-144 package of the TC23x, structural corner protrusions have to be considered for purposes of system design and integration with a maximum boundary of 6.2 mm.*

### Package Outlines TQFP-100 for TC23x



**Figure 13 Package Outlines TQFP-100 for TC23x**

*Note: For the exposed pad of the TQFP-100 package of the TC23x, structural corner protrusions have to be considered for purposes of system design and integration with a maximum boundary of 6.3 mm.*

### PLL\_ERAY\_TC.H002 Correction in Figure “PLL\_ERAY Block Diagram”

The signal originating from block “K2-Divider” in figure “PLL\_ERAY Block Diagram” in chapter “ERAY Phase-Locked Loop” of the User’s Manual is incorrectly labeled as PLLERAYSTAT.K1RDY.

## Correction

The correct name of the signal originating from block “K2-Divider” is PLLERAYSTAT.K2RDY.

## **PMC\_TC.H001** Check for permanent Overvoltage during Power-up

After an initial power-on with a permanent overvoltage condition on either  $V_{EXT}$ ,  $V_{DDP3}$  or  $V_{DD}$  supply rails, no overvoltage alarm may be generated by the SMU after configuration of the alarms, as the threshold transition condition has already happened.

However, in case an overvoltage condition was present, it will be indicated by flags OV13, OV33, and OVSWD, respectively, in register EVRSTAT.

## Recommendation

Check the OV13, OV33, and OVSWD flags in register EVRSTAT by software at start-up to identify an overvoltage condition.

## **PMC\_TC.H004** Selecting the WUT Clock Divider

Wake-up timer usage with  $PMSWCR3.WUTDIV = 1_B$  (10 ms count) for  $PMSWCR3.WUTREL$  values up to 20 ms is exposed to synchronization issues. The WUT counter  $PMSWUTCNT.WUTCNT$  is counted down to 0 every 10 ms, and reloading of  $WUTREL$  happens 10 ms later. If Standby request is sent before reloading  $PMSWCR3.WUTREL$ , regardless of  $PMSWSTATCLR.WUTWKPCLR$ , wake-up request is issued without counting down. This leads to immediate wake-up.

## Recommendation

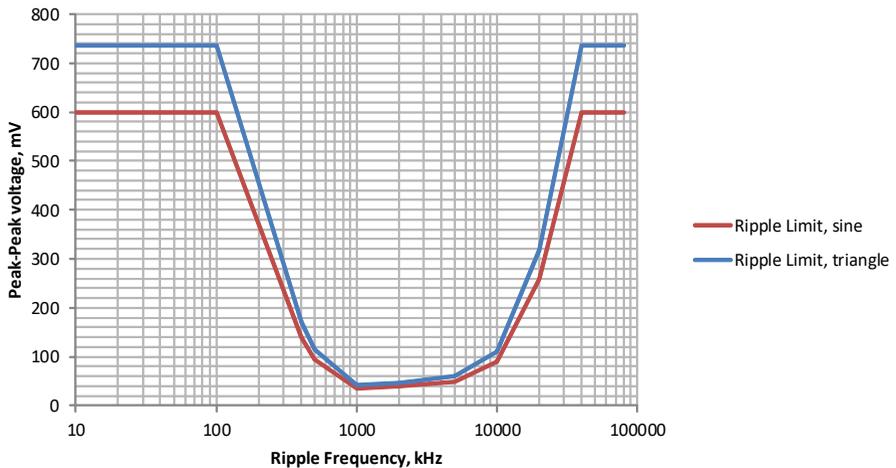
Use WUT with setting  $PMSWCR3.WUTDIV = 1_B$  only for longer time periods (more than 20 ms).

For shorter periods the 10  $\mu$ s clock should be used with setting  $PMSWCR3.WUTDIV = 0_B$  (default after reset).

**PMS\_TC.H002 Sensitivity to supply voltage ripple during start-up**

The internal back-up clock is sensitive to specific power supply voltage disturbance/ripple caused by a voltage ripple intrinsic to DC-DC converters. Specific conditions such as insufficient filtering of the ripple may lead to improper behavior of the start-up scheme of the back-up clock, and thus stuck-at state during the start-up of the microcontroller until this condition is removed.

The acceptable voltage vs. frequency characteristic is portrayed below on the chart:



**Figure 14 Ripple Voltage vs. Frequency Characteristic**

The diagram reflects acceptable ripple level during the cold start of the microcontroller at the respective VDDP3/VEXT/VEVRSB supply of the PMS subsystem, depending on the device and package type, as shown in the following table.

**Table 39 Pads/Pins sensitive to supply voltage ripple during start-up**

Device	Package	Pad/Pin	Symbol
TC29x	BGA-516	AA16	VEVRSB
TC29x	BGA-416	AD9	VEVRSB
TC29x, TC27x, TC26x	BGA-292	T11	VEVRSB
TC27x, TC26x	QFP-176	69	VEXT
TC26x	QFP-144	59	VEXT
TC23x	BGA-292	T11	VDDP3
TC23x, TC22x, TC21x	QFP-144	69	VDDP3
TC23x, TC22x, TC21x	QFP-100	47	VDDP3
TC22x, TC21x	QFP-80	37	VDDP3

### Recommendation 1

Apply an additional ceramic capacitor at the respective VDDP3/VEXT/VEVRSB supply input (at pins specified above) to attenuate the residual ripple of the buck converter. The resonant frequency of the additional filter capacitor shall be chosen in accordance with the amplitude-frequency characteristic given above and the switching frequency of the DC-DC converter in order to provide a proper attenuation in the range of interest.

The amount of ripple voltage can be approximated by  $V_{pk-pk} = I_{load} / (f \cdot C)$  and therefore the necessary nominal value of the blocking capacitance can be estimated as  $C = I_{load} / (f \cdot V_{pk-pk})$

It is recommended to take the  $I_{load}$  value as approximately 10 mA for the start-up load at the respective VDDP3/VEXT/VEVRSB domain before the internal regulator starts.

The frequency shall be taken same as the switching frequency of the external DC-DC voltage regulator. For example:

$$C = (0.010 \text{ A}) / (10^6 \text{ Hz} * 0.040 \text{ V}) = 0.25 * 10^{-6} \text{ F}$$

### Recommendation 2

Dimension the output LC filter of the external DC-DC converter to meet the limit of the ripple below the specified limit at the switching frequency. The effective value of ripple current flowing in and out of the buffer capacitor is calculated in accordance with standard formulas for the DC-DC buck converters. Selection of the low-ESR buffer capacitor is crucial in such applications, as the ESR value is directly proportional to the voltage drop caused by inductor current ripple.

### Recommendation 3

Supply the respective VDDP3/VEXT/VEVRSB rail by an external post LDO power stage.

### **PMS\_TC.H008 Interaction of interrupt and power management system - Additional information**

The description of steps to enter Idle, Sleep and Standby Mode in chapter “Power Management Overview” of the PMC chapters in the current TC2xx User’s Manuals is not comprehensive in explaining the dependency on pending interrupts as well as received interrupts. Hence, more explanation is provided here.

For a CPU to enter Idle Mode, it must have no interrupts pending. If it is in Idle Mode it will stay in Idle Mode until one of the specified wake-up events occurs – one of these is to have a pending interrupt.

Any SRN targeting a specific CPU (i.e. TOS set to that CPU), which is enabled, i.e. has SRE set, and has received a trigger event, i.e. has SRR set (whether by a received trigger from a peripheral or a master using the SETR control bit in the SRN) is a pending interrupt. Thus, even if a peripheral is shut down by having its clocks gated off, if it has presented a trigger event to the IR, and the SRE bit for that SRN is set, there will be a pending interrupt to the specified CPU.

It is not necessary for the priority of the pending interrupt to allow it to be taken, nor is it necessary for the CPU to have interrupt servicing enabled. It is possible and valid for Idle Mode to be entered with interrupts disabled, and to only re-enable interrupt acceptance subsequent to resuming execution. Equally, the CPU's priority may well dictate that the interrupt cannot be serviced immediately on re-enabling interrupts.

There may be some interrupts in a system that a CPU will be required to service and must exit Idle Mode (or Sleep Mode) or prevent entry to Idle Mode (or Sleep or Standby Mode) on their arrival. If one of these interrupts is raised prior to, or just as Idle Mode, Sleep Mode or Standby Mode is requested then that mode will not be entered.

The description for the REQSLP field states

- “In Idle Mode or Sleep Mode, these bits are cleared in response to an interrupt for the CPU, or when bit 15 of the corresponding CPU Watchdog Timer register (bit WDTCPUsSR.TIM[15]) changes from 0 to 1.”

For clarity, this also means, if a write to PMCSRx.REQSLP occurs while the IR has a pending interrupt for CPUx the write data will be ignored and the REQSLP value will remain as 00<sub>b</sub> “Run Mode”.

For the system to enter Sleep or Standby Mode by writing to PMCSRx.REQSLP (as opposed through an external low voltage condition), all CPUs must be in Idle Mode. Typically, first other CPUs will be brought into Idle Mode and then the master CPU will be the last to enter to Idle Mode as a transitional state of the request for the system mode Sleep or Standby. Consequently any pending interrupts for any CPU will prevent the entry into Sleep or Standby Mode.

## Recommendation

To ensure the transition to a power save mode, for a CPU intended to enter Idle Mode or for a system entering Sleep or Standby mode, all interrupts that are not intended to cause Run Mode to be re-entered or retained, should either have the SRE bit cleared in the respective SRN or be guaranteed to have the SRR bit clear.

If modifying the SRE bit of an SRN, to ensure the new state is reflected in IR arbitration information conveyed to the PMC and CPUs, sufficient time for an arbitration must have elapsed. Hence, a subset of the synchronisation

described in subsection “Changing the SRN configuration” of the IR chapter in the corresponding TC2xx User’s Manual is required.

After the last SRN (for CPUx) has been updated

- Read back the last SRN
- Read the LWSRx register

Clearing the SRR bit or disabling the source of the trigger can also be used if there are no timing hazards; i.e. no risk of a trigger being raised just before reconfiguring the peripheral (to not raise triggers), or no risk of an SRN that has had SRR cleared being set again while other SRNs are accessed. If the timing behaviour of these interrupt sources allows them to be disabled at source or in the SRN these are also valid methods. So long as the SRE bit and SRR bit are not both set, there will not be a pending interrupt. If the SRR bits are cleared, after the last SRN is modified there also needs to be a synchronisation step for the IR outputs to reflect the update before the PMCSRx is written.

Once there are no pending interrupts, request the power saving mode by writing to the respective PMCSRx.

*Note: There will still be several system clock cycles till the power saving mode is enabled by the PMC during which the CPU will continue to execute instructions.*

To ensure a deterministic boundary for execution to end after the power saving mode request, the write to PMCSRx should be followed by a DSYNC and a WAIT instruction.

### **PMU\_TC.H002 Impact of Application Reset on register FLASH0\_FCON**

Register FLASH0\_FCON is described in PMU chapter “Flash Configuration Control” as being reset by Application Reset with reset value 0091 XXXX<sub>H</sub> with a footnote adding the information

“<sup>1)</sup>The wait-cycles WSECDF, WSDFLASH, WSECPF and WSPFLASH are changed by the startup after system and power-on resets. **Attention: the configured value is only sufficient for the clock configuration used during startup.** The wait-cycles have to be configured after startup as described in <reference to the PMU section “Configuring Flash Wait Cycles”> before changing to higher clock frequencies.”

In this section the user is informed that after System Reset and Power-On Reset the wait cycles are configured to have a maximum allowed frequency of 100 MHz for  $f_{FSI}$  and  $f_{FSI2}$ .

In summary this results in the following reset behavior:

- Power-on reset and system reset: both change the wait-cycles to a value sufficient for  $f_{FSI}$  and  $f_{FSI2}$  at max 100 MHz.
- Application reset: changes the wait-cycles to a value not disclosed in the User's Manual. This value is WSPFLASH=10, WSECPF=2, WSDFLASH=45, WSECDF=2.

### Recommendation

Consequently after each reset the application software shall write values adapted to the clock configuration as described in the section "Configuring Flash Wait Cycles".

### PORTS\_TC.H006 Using P33.8 while SMU is disabled

Per default, the SMU is enabled ( $SMU\_CLC = 0x0$ ) and collects the alarms from the safety mechanisms defined by the safety concept. The SMU may optionally use P33.8 to output the Fault Signaling Protocol (FSP), selectable via register  $SMU\_PCTL$ . To satisfy safety requirements, it is ensured that the pad configuration of this pin is not affected by an application or system reset after the first 0-to-1 transition of bit  $SMU\_PCTL.PCS$ .

If the SMU is enabled, but is not using P33.8 for the FSP function, this pin may be used as general purpose input/output (GPIO) or alternate function input/output, controlled via the corresponding P33 registers.

However, if the SMU is disabled by software ( $SMU\_CLC.DISR = 1_B$ , i.e. not clocked), configuration of P33.8 (pull devices, driver settings, selection of alternate function, etc.) requires special considerations as described in the following, otherwise the configuration change may not become effective.

## Recommendations

- If P33.8 shall be used as GPIO or alternate function input/output, do not disable the SMU, i.e. keep SMU\_CLC = 0x0 (default after reset). In this case, the configuration of P33.8 may be changed by software at any time.
- Alternatively, configure P33.8 before the SMU is disabled by software (SMU\_CLC.DISR = 1<sub>B</sub>). After the SMU is disabled, the configuration of P33.8 can no longer be modified by software.
- Alternatively, if the SMU is disabled by software (SMU\_CLC.DISR = 1<sub>B</sub>, i.e. not clocked), clear bit position 8 at address 0xF003 D364 in the P33 address space once after any reset (Application, System Reset, PORST) before configuring P33.8. Controlling P33.8 as FSP by SMU is possible only once after a reset.

*Note: Write access to address 0xF003 D364 is Safety ENDINIT protected.*

## **PORTS\_TC.H016 Oscillating signal may enable DXCPL and reconfigure the functionality of the port pins P14.0 and P14.1**

The port pin P14.1 can be configured as input for different modules such as GTM input, CAN input, FlexRay input or General Purpose Input. In case oscillations are appearing on this input, DXCPL may get enabled unintentionally on P14.0 and P14.1 and disable the module previously assigned to the pins.

### **Recommendation**

Please refer to application note AP32264 “DXCPL DAP over CAN Physical Layer” for further information and how this situation can be prevented.

*Note: See also MultiCAN\_TC.H007 (Oscillating CAN Bus may Disable the CAN Interface).*

## **QSPI\_TC.H005 Stopping Transmission in Continuous Mode**

The QSPI module supports the following mechanisms to (temporarily) suspend its operation:

- Pause by setting bit GLOBALCON.EN = 0<sub>B</sub> via software
- Disable by setting bit CLC.DISR = 1<sub>B</sub> via software

- Sleep Mode (enabled with CLC.EDIS = 0) requested by hardware
- Suspend Mode requested by hardware (debugger)

These modes and their handling is described in detail in section “Operation Modes” of the QSPI chapter in the User’s Manual.

In **Continuous Mode**, the following specific behavior of QSPI module has to be considered:

- In case the QSPI module is put into **Pause** state by setting bit GLOBALCON.EN = 0<sub>B</sub> via software, it continues transmission until the end of the TRAIL phase of the frame with BACON.LAST = 1<sub>B</sub>.
- In case the QSPI module is put into **Disable**, **Sleep**, or **Suspend** mode, the frame is stopped after the next trailing delay (character n). In case BACON.LAST was not =1<sub>B</sub> at that time, transmission continues with character n+2 when operation from Disable/Sleep/Suspend state is resumed, i.e. data loss (character n+1) will occur.

### Recommendation

Ensure that software does not put the QSPI module into Pause or Disable state (via GLOBALCON.EN or CLC.DISR) while a transmission in Continuous Mode is ongoing.

If Sleep Mode is used in the system, disable acceptance of sleep requests (set CLC.EDIS = 1<sub>B</sub>) before starting data transmission in Continuous Mode.

During debugging, ensure that the QSPI is not suspended while it is transmitting in Continuous Mode.

### QSPI\_TC.H006 Corrections to Figures “QSPI - Frequency Domains” and “Phase Duration Control, Overview”

In the current version of the User’s Manual,

- Figure “QSPI - Frequency Domains” erroneously uses the term “f<sub>PER</sub>” instead of “f<sub>BAUD2</sub>”, and
- Figure “Phase Duration Control, Overview” erroneously uses the term “T<sub>PER</sub>” instead of T<sub>BAUD2</sub>”.

### Correction

- $f_{\text{SCLK}} = 1/f_{\text{BAUD2}}$  in Figure “QSPI - Frequency Domains”, and
- $T_{\text{BAUD2}} = 1/f_{\text{BAUD2}}$  in Figure “Phase Duration Control, Overview”.

### **QSPI TC.H007 RXFIFO Overflow Bit Behavior in Slave Mode**

In slave mode, if no data word has been written to TXFIFO during initialization before the master starts sending data, the error flag corresponding to an RXFIFO overflow (bit STATUS.5) is set to 1<sub>B</sub>.

### Recommendation

To avoid this RXFIFO overflow event, write (at least) one word to TXFIFO during initialization and after each reset in slave mode. For following transmissions, no data need to be written to TXFIFO to avoid this effect.

### **QSPI TC.H008 Details of the Baud Rate and Phase Duration Control - Documentation update**

To enhance readability, the last part of the second paragraph in the QSPI chapter “Details of the Baud Rate and Phase Duration Control”, starting with “Variations in the baud rates of the slaves ..”, shall be rephrased as shown below.

For further details see also the formulas in the chapter mentioned above and in the figures in chapter “Calculation of the Baud Rates and the Delays” in the User’s Manual.

### Documentation update

Variations in the baud rates of slaves of one module are supported by the ECONz.Q and the ECONz.A/B/C bitfield settings allowing for a flexible bit time variation between the channels in one module.

### **QSPI\_TC.H009 Dummy frame required after changing SCLK polarity and phase in three wire mode**

When three wire mode is used, and the SCLK polarity (bit ECONz.CPOL) or phase (bit ECONz.CPH) of the master is changed by software, the state of the clock and data signals is not defined before the first data is transmitted.

This may result in wrong data being received or transmitted by the slave.

#### **Recommendation**

After the SCLK polarity (bit ECONz.CPOL) or phase (bit ECONz.CPH) is changed by software, transmission of a dummy frame is required. The pad enable shall be after transmission of the dummy frame, such that the slave will not notice the dummy frame.

*Note: In four wire mode where the slave is controlled by a select signal from the master, this issue has no effect, because the output signals from the master are at the correct levels by the time the slave select signal gets active.*

### **RESET\_TC.H002 Unexpected SMU Reset Indication in SCU\_RSTSTAT**

Under certain conditions the Reset Status Register SCU\_RSTSTAT can show an SMU reset indication in addition to the real reset trigger (e.g. a SW reset).

The explanation of this behavior refers to section “Reset Generation” and following pages in chapter “RCU” of the User’s Manual.

Figure “Reset Overview” shows that all warm resets are executed in a defined sequence. This sequence ensures that first the active CPUs are ramped down, then at 80µs the Flash receives an idle request and at 180µs the reset is executed.

The idle request to the Flash makes it immediately busy, all read requests after this point fail with a bus error. All non-CPU masters (HSM, Ethernet, HSSL, DMA and DAM) however continue operation from 80µs to 180µs. When one of these masters reads the busy Flash, a bus error is signaled to the SMU as alarm ALM3[30] (SRI) and/or ALM3[31] (SPB).

If the SMU is configured to react on this by a reset request, this will be noted in the SCU\_RSTSTAT register in addition to the original warm reset.

This applies mainly to the master HSM which fetches its code from PFlash.

### Recommendations

- Generally a different alarm handling can be configured in the SMU for the mentioned alarms, e.g. trigger an NMI trap but not a reset.
- When the application detects after reset that SCU\_RSTSTAT has an additional SMU reset indication it might ignore it and proceed based on the other reset indication.
- In case of SW resets the application can prepare the system just before activating the reset:
  - The non-CPU masters can be disabled or in case of HSM it can be informed about the imminent SW reset and continue execution from RAM.
  - The mentioned alarms can be disabled or the alarm reaction can be changed to trigger an NMI trap.
  - The SMU module reset can be used to reconfigure the SMU into its initial state in which only watchdog timeout alarms are handled.

### **RESET\_TC.H003 Usage of the Prolongation Feature for ESR0 as Reset Indicator Output**

The ESR0 pin can be used as reset indicator output and in such a case its active low state can be prolonged upon user-configurable selection as described in section “ESRx as Reset Output” of chapter “Reset Control Unit (RCU) in the User’s Manual.

According to this description, an ESR0CNT value of 0 defines “as soon as possible after start of Boot Code execution”, where “as soon as possible” means:

- about 500  $\mu$ s after cold power-on,
- not less than 20  $\mu$ s after other types of reset.

## Warning

In case of  $ESR0CNT = 2$ , the ESR0 pin will never be released by the device and the user code will never start.

*Note: On the other hand - as explained before - configuring an ESR0CNT value of 1 or 2 would anyhow not be effective as a prolongation time below 20  $\mu$ s is conceptually unachievable.*

## Recommendation

Do not configure  $ESR0CNT = 2$ .

If prolongation of about 20  $\mu$ s or below is needed, configure  $ESR0CNT = 3$  or 0 instead.

## **RESET\_TC.H004 Effect of Power-on and System Reset on DSPR**

The following part of footnote <sup>2)</sup> on Table “Effect of Reset on Device Functions” in the RCU chapter “Module Reset Behavior” regarding the effect of startup firmware on Data Scratchpad RAM (DSPR): “DSPR is partially used as a scratchpad by the startup firmware. Previous data stored in the upper 32kB will be overwritten on start-up” is incorrect.

The correct effect is described in the Boot ROM chapter “RAM overwrite during start-up”:

Start-up procedure upon power-on and system reset can overwrite up to 8 Kbyte at the beginning of CPU0 DSPR.

## **SCU\_TC.H009 LBIST Influence on Pad Behavior**

The behavior of the GPIO and ESR0/1 pads during LBIST execution is as follows:

- ESR0 is switched to input direction during LBIST with weak pull-up and pull-down driver disabled (i.e. pad is tri-stated).
- ESR1 is switched to input direction during LBIST with weak pull-down driver enabled.

- Other GPIO pins are switched to input direction with weak pull-up devices either stable active or inactive (depending on LBIST user configuration).

### **SCU\_TC.H010 LBIST Signature Depends on Debug Interface Configuration**

The following three cases generate different sets of LBIST MISR signatures:

1. Pin  $\overline{\text{TRST}}$  is held high during  $\overline{\text{PORST}}$  rising edge (DAP operation): In this case the further values of  $\overline{\text{TRST}}$  will have no influence on the MISR signature
2. Pin  $\overline{\text{TRST}}$  is held low during  $\overline{\text{PORST}}$  rising edge (JTAG operation):  $\overline{\text{TRST}}$  is held continuously low also during LBIST operation
3. Pin  $\overline{\text{TRST}}$  is held low during  $\overline{\text{PORST}}$  rising edge (JTAG operation):  $\overline{\text{TRST}}$  is switched to high after  $\overline{\text{PORST}}$  has been released and at least one pulse occurred at TCK before LBIST starts.

If DXCM/DXCPL (Debug over CAN) is not needed it is recommended to keep pin  $\overline{\text{TRST}}$  always at a high level during  $\overline{\text{PORST}}$  rising-edge in the application environment (also in final application). This makes the MISR signature independent from further  $\overline{\text{TRST}}$  behavior and still allows debug access via DAP or to completely disable the debug IF via software (by setting OIFM.DAPMODE = 111<sub>B</sub>).

In the DXCM/DXCPL enabled case it is recommended to keep pin  $\overline{\text{TRST}}$  always at a low level (also after  $\overline{\text{PORST}}$  has been released). In this operation case a different set of MISR signatures will be received (case 2 in above list). Consequently the application software needs to be prepared to accept LBIST MISR signature results from case 1 or from case 2 as pass criteria.

### **SCU\_TC.H013 Correction to Register References in Chapter “Watchdog Timers”**

Some references to register names in chapter “Watchdog Timers” of the User’s Manual are incorrect.

The corrected references and their section headers are listed in **bold** below.

## Section Password Access to WDTxCON0

.. To ensure that a CPU fault could not allow a fault to be ignored an option is provided to prevent watchdog unlocking if the Safety Management Unit (SMU) is not in the RUN state. This option may be enabled by bit **WDTxCON1.UR**. If the password is valid and the SMU state meets the requirements of the **WDTxSR.US** bit then WDTxCON0 will be unlocked as soon as the Password Access is completed. ..

## Section Timer Operation

.. The parameter divider represents the user-programmable source clock division selected by **WDTxCON1.IRx**, which can be 64, 256 or 16384.

## Section Watchdog Timer Registers

- **WDTSCON1** - Safety WDT Control Register 1:
  - References to WDTxCON0 and WDTxSR should be consequently to **WDTSCON0** and **WDTSSR** in the context of WDTSCON1.
- **WDTCPUxCON1** - CPUx WDT Control Register 1:
  - References to WDTSCON0 and WDTSSR should be consequently to **WDTCPUxCON0** and **WDTCPUxSR** in the context of WDTCPUxCON1.

## SCU\_TC.H014 Reset Value of Bit Field IOCR.PC1 - Control for Pin ESR1

The reset value of register SCU\_IOCR is documented as 0000 20E0<sub>H</sub> in chapter “Reset Control Units” of the User’s Manual, i.e. the reset value of bit field PC1 = 2<sub>H</sub>.

This is not always correct under all circumstances:

The actual SCU\_IOCR reset value should be considered as 0000 X0E0<sub>H</sub> with the explanations given in the following [Documentation Update](#).

## Documentation Update

The reset value of bit field SCU\_IOCR.PC1 is influenced by pin HWCFG6 and bit PMSWCR0.TRISTREQ:

- When a cold reset is activated and HWCFG6=1 then PC1 is reset to 2<sub>H</sub> and pin ESR1 will have input pull-up mode.

- If HWCFG6=0 then PC1 is reset to 0<sub>H</sub> and  $\overline{\text{ESR1}}$  will have tri-state mode. PC1 and the  $\overline{\text{ESR1}}$  reset state can also be configured by software with the PMSWCR0.TRISTREQ bit. PMSWCR0.TRISTREQ is not affected by warm reset or wake-up from standby so the IOCR.PC1 reset value is configured as per the state of the TRISTREQ bit prior to the warm reset.

### **SENT\_TC.H003 First Write Access to Registers FDR and TPD after ENDINIT Status Change**

Due to an extra registering stage of the ENDINIT signal from the SCU inside the SENT kernel, the behavior of the first write access to SENT registers FDR and TPD protected by the Endinit write protection scheme after an ENDINIT status change is as follows:

- After unlocking protection (ENDINIT change from 1 to 0), if the first access to the SENT module is a write to FDR or TPD, it will still view ENDINIT as locked (value 1). The contents of FDR or TPD is not changed, but no BCU alarm will be generated, as the ENDINIT does not indicate a protected status in case of the access.
- By setting protection again (ENDINIT change from 0 to 1), if the first access to the SENT module is a write to FDR or TPD, it will still be effective, i.e., the value will be written. Nevertheless a SMU alarm through BCU will be generated as the protection status is ENDINIT.

*Note: After the first read of any SENT register, or first write to any SENT register, the ENDINIT change will be correctly considered for all following accesses. The CLC, KRST0/1 and KRSTCLR registers (that also have Endinit protection) are not affected at all. An initial value of 0 for ENDINIT is seen by SENT after reset before the first access.*

### **Recommendation**

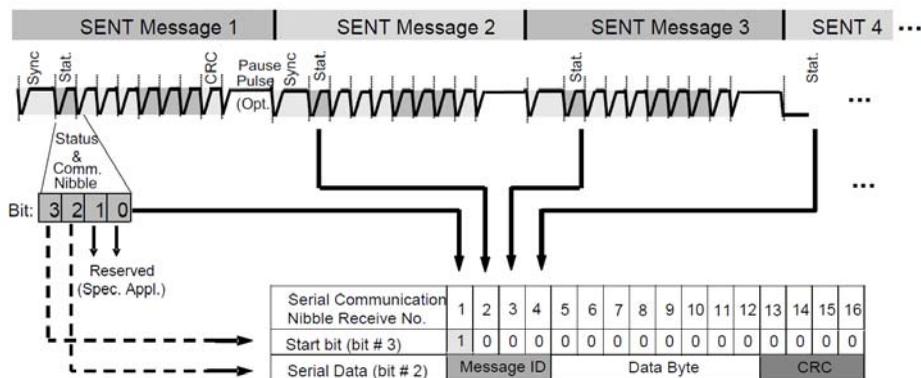
After a change of the ENDINIT protection status, first perform a read of any SENT register or a write to a non-Endinit-protected SENT register. The second access is then always equipped with correct information of ENDINIT.

**SENT\_TC.H004 Short Serial Message - Figure Correction**

In Figure “Short Serial Message, Serial Data Encoding over 16 messages” of the SENT chapter, the arrows originating from bits 2 and 3 of the Status & Comm Nibble are routed incorrectly and must be swapped.

**Correction**

**Figure 15** shows a corrected version of this figure.



One serial message is composed of 16 SENT consecutive error-free messages.

**Figure 15 Short Serial Message, Serial Data Encoding over 16 messages**

**SENT\_TC.H005 Interface Connections of the SENT Module - Documentation Correction**

The following corrections apply to chapter “Interface Connections of the SENT Module” in the SENT chapter of the User’s Manual:

### Figure “SENT Module Implementation and Interconnections”

- In TC23x/TC22x/TC21x, no TRIGOn signals are connected from the SENT module to the Interrupt Router (IR). All references to TRIGOn should be ignored in this figure.
- The range of index n for connected trigger inputs TRIGn in TC23x/TC22x/TC21x is  $n = 0..3$ .

### Interrupt and DMA Controller Service Requests

In TC23x/TC22x/TC21x, request lines SR0..3 of the SENT module are connected via the Interrupt Router and can be selected in register INPx accordingly. Values  $\geq 0100_B$  are reserved and should not be used for the bit fields in register INPx.

### SMU\_TC.H001 Write all bit fields of SMU\_PCTL with one write access

When configuring the FSP pin (e.g. P33.8), all bit fields (HWDIR, HWEN and PCS) of register SMU\_PCTL must be written with the same write access.

Otherwise, when first writing a  $1_B$  to HWEN before writing a  $1_B$  to PCS, the pad configuration will be modified to push/pull configuration before it is latched into field PCFG.

*Note: When  $PCS = 1_B$ , the bit fields PCFG and PCS are protected against any changes until the next power on reset. HWEN and HWDIR may still be modified by SW, unless locked via register SMU\_KEYS.*

### SMU\_TC.H005 Correction to Figure “SMU Register Map”

The start address “@SMU + 0x0E0” for the SMU System Registers shown in the lower part of figure “SMU Register Map” in the SMU chapter of the User’s Manual is incorrect.

The correct start address is “@SMU + 0x7E0”.

Addresses listed in table “Registers Overview” of the SMU chapter are correct.

**SMU\_TC.H006 Description of Bit EFRST in Register SMU\_AGC**

In the SMU chapter of the User's Manual, the description of the encoding of bit EFRST (Enable FAULT to RUN State Transition) in register SMU\_AGC (Alarm Global Configuration) is missing.

The complete description should be as shown in [Table 40](#):

**Table 40 Bit EFRST in Register SMU\_AGC**

Field	Bits	Type	Description
EFRST	29	rw	<b>Enable FAULT to RUN State Transition</b> 0 <sub>B</sub> FAULT to RUN State Transition disabled 1 <sub>B</sub> FAULT to RUN State Transition enabled See section " <b>FSP Fault State</b> " for the usage of this field.

**SMU\_TC.H007 SPB Bus Control Unit (SBCU) Alarm Signalling to SMU**

ALM3[31] is dedicated to System Peripheral Bus (SPB) alarms. As described in table "Alarm Mapping related to ALM3 group" in the SMU chapter of the User's Manual, an SPB bus error can result from multiple root causes, including protocol violation, incorrect address, register access protection violation.

More details on the SPB related error conditions can be found in the "On-Chip Bus System" chapter:

The SBCU signals an alarm to the SMU whenever it detects

- a SPB transaction that was finished with a Bus Error (Error Acknowledge)
- an un-implemented Address (no slave responds to a transaction request)
- a SPB transaction that was finished by a Time-out.

The alarm signaling to the SMU is independent of the BCU configuration (e.g. BCU interrupt configuration, BCU debug status).

## **SMU\_TC.H009 Alarm Table Corrections**

Some alarm tables were unintentionally changed between User's Manual (UM) version V1.0 and V1.1.

In the following, the issues are described and the correct table parts are included.

- **Table 10-2 HwAlarmOut[3:0]: CPU0 DCACHE/DSPR SRAM**

The PreAlarms 1, 3, 5, 7 of the DSPR2 part (TC23x only) were accidentally removed in UM V1.1.

The following **Table 41** copied from UM V1.0 is correct:

**Table 41 Table 10-2 HwAlarmOut[3:0]: CPU0 DCACHE/DSPR SRAM**

<b>Function</b>
HwAlarmOut[0]= PreAlarm[0=CPU0.DMI.DSPR.(ECC single bit correction)] or PreAlarm[1=CPU0.DMI.DSPR2.(ECC single bit correction)]
HwAlarmOut[1]= PreAlarm[2=CPU0.DMI.DSPR.(ECC uncorrectable error)] or PreAlarm[3=CPU0.DMI.DSPR2.(ECC uncorrectable error)]
HwAlarmOut[3]= PreAlarm[4=CPU0.DMI.DSPR.(Address error)] or PreAlarm[5=CPU0.DMI.DSPR2.(Address error)]
HwAlarmOut[4]= PreAlarm[6=CPU0.DMI.DSPR.(Address buffer overflow)] or PreAlarm[7=CPU0.DMI.DSPR2.(Address buffer overflow)]

- **Table 10-3 HwAlarmOut[7:4]: CPU1 DCACHE/DSPR SRAM**

HwAlarmOut[7:4] are reserved. The following **Table 42** copied from UM V1.0 is correct:

**Table 42 Table 10-3 HwAlarmOut[7:4]: CPU1 DCACHE/DSPR SRAM**

Function
HwAlarmOut[7:4] are reserved

- **Table 10-4 HwAlarmOut[15:8]: CPU2 SRAMs**

HwAlarmOut[15:8] are reserved. The following [Table 43](#) copied from UM V1.0 is correct:

**Table 43 Table 10-4 HwAlarmOut[15:8]: CPU2 SRAMs**

Function
HwAlarmOut[15:8] are reserved

- **Table 10-5 HwAlarmOut[19:16]: GTM SRAMs**

HwAlarmOut[19:16] are reserved. The following [Table 44](#) copied from UM V1.0 is correct:

**Table 44 Table 10-5 HwAlarmOut[19:16]: GTM SRAMs**

Function
HwAlarmOut[19:16] are reserved

- **Table 10-7 HwAlarmOut[27:24]: CAN SRAM**

The PreAlarms of the second CAN module RAMs (TC23x only) were accidentally removed in UM V1.1; these are: 81, 83, 85, 87.

The following [Table 45](#) copied from UM V1.0 is correct:

**Table 45 Table 10-7 HwAlarmOut[27:24]: CAN SRAM**

<b>Function</b>
HwAlarmOut[24]= PreAlarm[80=CAN.SRAM.MCAN0.(ECC single bit correction)] or PreAlarm[81=CAN.SRAM.MCAN1.(ECC single bit correction)]
HwAlarmOut[25]= PreAlarm[82=CAN.SRAM.MCAN0.(ECC uncorrectable error)] or PreAlarm[83=CAN.SRAM.MCAN1.(ECC uncorrectable error)]
HwAlarmOut[26]= PreAlarm[84=CAN.SRAM.MCAN0.(Address error)] or PreAlarm[85=CAN.SRAM.MCAN1.(Address error)]
HwAlarmOut[27]= PreAlarm[86=CAN.SRAM.MCAN0.(Address buffer overflow)] or PreAlarm[87=CAN.SRAM.MCAN1.(Address buffer overflow)]

- **Table 10-8 HwAlarmOut[31:28]: LMU sub-system SRAMs**

The PreAlarms of the FFT RAMs (TC23x ADAS only) were accidentally removed in UM V1.1; these are: 230, 231, 233, 234, 236, 237, 239, 240.

The following [Table 46](#) copied from UM V1.0 is correct:

**Table 46 Table 10-8 HwAlarmOut[31:28]: LMU sub-system SRAMs**

<b>Function</b>
HwAlarmOut[28]= PreAlarm[88=LMU.DAM.SRAM(ECC single bit correction)] or PreAlarm[230=LMU.FFT0.SRAM(ECC single bit correction)] or PreAlarm[231=LMU.FFT1.SRAM(ECC single bit correction)] or
HwAlarmOut[29]= PreAlarm[90=LMU.DAM.SRAM(ECC uncorrectable error)] or PreAlarm[233=LMU.FFT0.SRAM(ECC uncorrectable error)] or PreAlarm[234=LMU.FFT1.SRAM(ECC uncorrectable error)] or

**Table 46 Table 10-8 HwAlarmOut[31:28]: LMU sub-system SRAMs**

<b>Function</b>
HwAlarmOut[30]= PreAlarm[92=LMU.DAM.SRAM(Address error)] or PreAlarm[236=LMU.FFT0.SRAM(Address error)] or PreAlarm[237=LMU.FFT1.SRAM(Address error)] or
HwAlarmOut[31]= PreAlarm[94=LMU.DAM.SRAM(Address buffer overflow)] or PreAlarm[239=LMU.FFT0.SRAM(Address buffer overflow)] or PreAlarm[240=LMU.FFT1.SRAM(Address buffer overflow)] or

- **Table 10-9 HwAlarmOut[34:32]: SRI Agents**

PreAlarm[116] is accidentally listed as HSSL.SRI\_MASTER(SRI Read Data Phase Error) in UM V1.1.

Actually PreAlarm[116] is reserved.

- **Table 9-11 HwAlarmOut[44:41]: Misc. SRAMs**

The PreAlarms accidentally listed as assigned to PSI5 and CIF are actually reserved; these are 145, 147-149, 153, 155-157, 161, 163-165, 169, 171-173.

- **Table 10-12 HwAlarmOut[45]: Watchdogs Timeout**

The PreAlarms accidentally listed as assigned to WDTCPU1 and WDTCPU2 in UM V1.1 are actually reserved; these are 175, 176.

- **Table 10-13 HwAlarmOut[50:46]: PMU Alarms**

The PreAlarms accidentally listed as assigned to PFLASH1 in UM V1.1 are actually reserved; these are: 179, 187, 195, 203, 211.

- **Table 10-18 TC21x/TC22x/TC23x Alarm Mapping related to ALM1 Group**

This table incorrectly shows alarms for CPU1 in UM V1.1.

These alarms actually are reserved, as shown in the following [Table 47](#) copied from UM V1.0:

**Table 47 Table 10-18 Alarm Mapping related to ALM1 Group**

Alarm Index	Module	Description
ALM1[31:0]	Reserved	Reserved

- **Table 10-20 TC21x/TC22x/TC23x Alarm Mapping related to ALM3 Group**

The rows in the following [Table 48](#) replace the corresponding rows of the table in UM V1.1:

**Table 48 Part of Table 10-20 with corrected ALM3 Group Mapping**

Alarm Index	Module	Description
ALM3[9]	Reserved	Reserved
ALM3[13]	Reserved	Reserved
ALM3[14]	Reserved	Reserved
ALM3[19]	Reserved	Reserved
ALM3[20]	Reserved	Reserved
<b>ALM3[27]</b>	<b>Registers</b>	<b>Safety Mechanism: Register Monitor Alarm: register error detection</b>
ALM3[28]	SCU/LSCU	Safety Mechanism: Lockstep Dual Rail Monitor Alarm: dual rail error Note: monitors the dual-rail property (inverted signals) from the lockstep comparator unit (LSCU) alarms.

- **Table 10-21 TC21x/TC22x/TC23x Alarm Mapping related to ALM4 Group**

There are no GTM SRAMs in this device. The rows in the following [Table 49](#) replace the corresponding rows of the table in UM V1.1:

**Table 49 Part of Table 10-21 with corrected ALM4 Group Mapping**

Alarm Index	Module	Description
ALM4[3:0]	Reserved	Reserved
ALM4[7:4] connects to pre-alarms specified by table “HwAlarmOut[27:24]: CAN SRAM”		

- **Table 10-23 TC21x/TC22x/TC23x Alarm Mapping related to ALM6 Group**

This table incorrectly shows alarms for CPU2 in UM V1.1.

These alarms actually are reserved, as shown in the following [Table 50](#) copied from UM V1.0:

**Table 50 Table 10-23 Alarm Mapping related to ALM6 Group**

Alarm Index	Module	Description
ALM6[31:0]	Reserved	Reserved

### **SMU\_TC.H010 Clearing individual SMU flags: use only 32-bit writes**

The SMU registers shall only be written via 32-bit word accesses (i.e. ST.W instruction), as mentioned in table “Registers Overview” of the SMU chapter in the User’s Manual.

If any other instruction such as LDMST or SWAPMSK.W is used to modify only a few bits in the 32-bit register, then this may have the effect of modifying/clearing unintended bits.

### **Recommendation (Examples in C Language)**

- **Example 1:** To clear status flag SF2 in register AG0, use:
  - SMU\_AG0.U = 0x0000 0004;
- **Example 2:** To clear status flags EF2 in register RMEF and RMSTS, use:
  - SMU\_RMEF.U = 0xFFFF FFFB;
  - SMU\_RMSTS.U = 0xFFFF FFFB;

Here the <REGISTER>.U implies writing to the register as an unsigned integer, which normally results in a compiler translation into an ST.W instruction.

### Safety Considerations

As long as software uses only 32-bit writes to the SMU registers, there is no risk of malfunction.

In case the software does not use 32-bit writes (and for example uses bit-wise operations such as LDMST instructions instead) – then potentially unintended flags may be written and modified in the SMU registers. Depending on the application, this may potentially have an impact on safety and/or diagnostics.

*Note: The SMU reaction itself (e.g. alarm action triggering) is not affected even if the software unintentionally clears additional bits by not using a 32-bit write as recommended.*

### SMU\_TC.H013 Increased Fault Detection for SMU Bus Interface (SMU\_CLC Register)

Transient faults can possibly affect the SMU\_CLC register and lead to disabling the SMU\_core. This unintended switching off of SMU\_core cannot be detected if the FSP protocol is not used at all or used in FSP bi-stable mode.

### Recommendation

In order to increase the capability of the microcontroller to detect such faults it is recommended to:

- **Option 1:** Use FSP Dynamic dual-rail or Time-switching protocol only, don't use FSP bi-stable protocol.
- **Option 2:** In case FSP protocol is not used at all or Recommendation Option 1 is not possible, the [Application SW] shall read periodically, once per FTTI, the SMU\_CLC register to react on unintended disabled SMU.

### **SMU\_TC.H014 Unintended short pulse on FSP pins in Time switching or Dual-rail mode**

Due to an internal synchronization issue, an unintended short pulse of a duration of around 80 ns can be seen on the FSP pins if the FSP pins are configured for Time switching or Dual-rail mode, and one of the following scenarios happens in the SMU state machine:

- scenario a): transition from START to RUN state
- scenario b): transition from FAULT to RUN (Fault-Free) state

#### **Recommendation**

- Workaround for scenario a):
  - Enable FSP by writing SMU\_PCTL register 10 SPB clock cycles (or more) after sending SMU\_ReleaseFSP() command.
- Assessment for scenario b):
  - The pulse in scenario b), if it occurs, cannot be avoided but has no safety impact as the unintended pulse happens during the transition from fault state to fault-free state. This state transition is not considered as safety relevant.

### **SRI\_TC.H001 Using LDMST and SWAPMSK.W instructions on SRI mapped Peripheral Registers (range 0xF800 0000-0xFFFF FFFF)**

The LDMST and SWAPMSK.W instructions in the AURIX™ microcontrollers are intended to provide atomicity as well as bit-wise operations to a targeted memory location or peripheral register. They are also referred to as Read-Modify-Write (RMW) instructions.

The bit-manipulation functionality is intended to provide software a mechanism to write to individual bits in a register, without affecting other bits. The bits to be written can be selected via a mask in the instruction. Please refer to the TriCore Architecture Manual for further information about these instructions and their formats.

## Restrictions for SRI mapped Peripherals

The bit-manipulation functionality is supported only on registers accessed via the SPB bus, and is not supported on the SRI mapped peripheral range (i.e. address range 0xF800 0000 to 0xFFFF FFFF, including (if available) EBU, PMU0, SRI Crossbar, LMU, DAM, FFT, CPUx SFRs and CSFRs, MCDS, miniMCDS); see table “On Chip Bus Address Map of Segment 15” in chapter “Memory Map”).

On the SRI mapped peripherals, usage of these instructions always results in all the bits of a register being written, and not just specific individual bits.

*Note: The instructions are still executed atomically on the bus – i.e the SRI is locked between the READ and the WRITE transaction.*

## **STM\_TC.H001 Effect of kernel reset on interrupt outputs STMIR0/1**

The clock ratio  $f_{STM} : f_{SPB}$  is determined by the settings of bit fields STMDIV and SPBDIV in registers CCUCON1 and CCUCON0, respectively.

If  $f_{STM} \leq f_{SPB}$ , and a kernel reset of the STM module is performed in the same clock cycle where a compare match of the STM with the CMP0 or CMP1 registers occurs, a transition on the interrupt outputs STMIR0 or STMIR1 may occur. This may e.g. trigger the External Request Unit (ERU), or set the corresponding Service Request flags SRC\_STMmSR0.SRR or SRC\_STMmSR1.SRR in the Interrupt Router ( $m = 0, 1, 2$ , depending on number of CPUs).

*Note: For  $f_{STM} > f_{SPB}$ , this effect will not occur.*

## Recommendation

If  $f_{STM} \leq f_{SPB}$ , set bits ICR.CMP0EN = 0<sub>B</sub> and ICR.CMP1EN = 0<sub>B</sub> to disable the compare match interrupts before performing the STM kernel reset.

## **STM\_TC.H002 Access Protection for STM Control Registers**

The access protection symbol ‘P’ to indicate Access Enable Register protection is missing in table “Registers Overview - STM Control Registers” of the STM

chapter in the User's Manual for the STM registers CMP0, CMP1, CMCON, ICR, ISCR.

The STM registers CMP0, CMP1, CMCON, ICR, ISCR actually have protection via the Access Enable registers (ACCEN0/1), as shown in the following [Table 51](#).

**Table 51 Correction to Table Registers Overview - STM Control Registers**

Short Name	Description	Offset Addr.	Access Mode		Reset
			Read	Write	
CMP0	Compare Register 0	30 <sub>H</sub>	U, SV	U, SV, P	Application
CMP1	Compare Register 1	34 <sub>H</sub>	U, SV	U, SV, P	Application
CMCON	Compare Match Control Register	38 <sub>H</sub>	U, SV	U, SV, P	Application
ICR	Interrupt Control Register	3C <sub>H</sub>	U, SV	U, SV, P	Application
ISCR	Interrupt Set/Clear Register	40 <sub>H</sub>	U, SV	U, SV, P	Application

### **STM\_TC.H003 Suspend control for STMx - Documentation Update**

In contrast to the register description of bit OCS.SUS in the STM chapter of the current User's Manual, the suspend functionality of STMx is controlled by signal CPUxSUSOUT of the corresponding CPUx (and not by the signal coming from the OCDS Trigger Switch (OTGS)).

Therefore, the description for bit OCS.SUS in the STM chapter should read:

- “Controls the sensitivity to the suspend signal coming from the CPU (CPUxSUSOUT)”.

**STM\_TC.H004 Access to STM registers while STMDIV = 0**

If accesses to STM kernel registers are performed while field STMDIV = 0<sub>H</sub> in CCU Clock Control register CCUCON1 (i.e. clock  $f_{\text{STM}}$  is stopped),

- the SPB bus gets locked after the first access until a timeout (defined in BCU Control register field SBCU\_CON.TOUT) occurs;
- after the second access the STM slave will answer with RTY (retry) until the STM is clocked again with STMDIV > 0<sub>H</sub>.

**Recommendation**

Do not access any STM kernel register while CCUCON1.STMDIV = 0<sub>H</sub>.